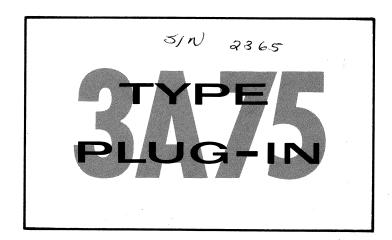
INSTRUCTION

Serial Number 2365

PRIOR TO SERIAL NUMBER 1120 THE TYPE 3A75 WAS REFERRED TO AS TYPE 75. The 3A75 Plug-in is interchangeable with the 75 and this manual also applies to the type 75 as far as operation and calibration are concerned.



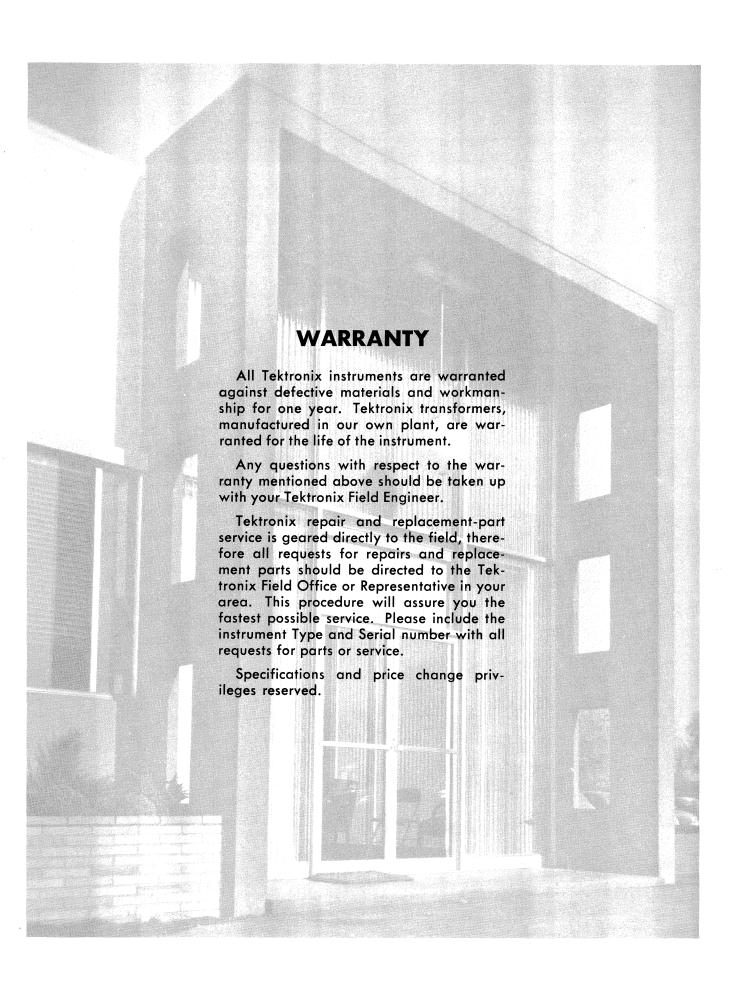
Tektronix, Inc.

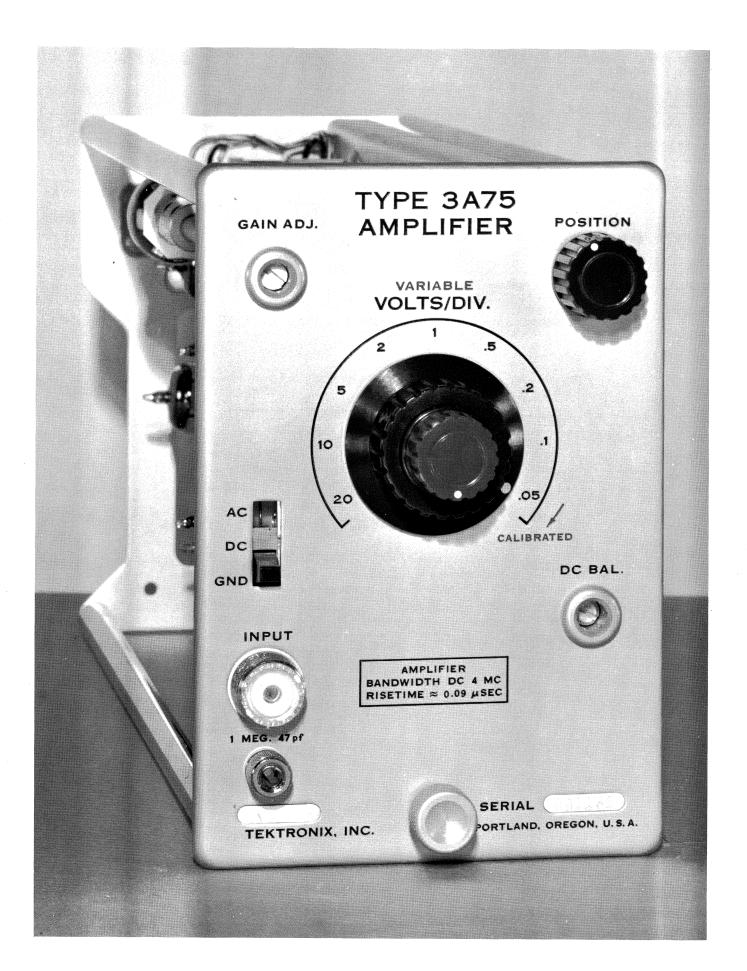
S.W. Millikan Way ● P. O. Box 500 ● Beaverton, Oregon ● Phone MI 4-0161 ● Cables: Tektronix

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TYPE 3A75 AMPLIFIER

INTRODUCTION

The Tektronix Type 3A75 Amplifier is a wide-band amplifier designed for use with the Tektronix Type 561 and Type 565 Oscilloscopes. It has a bandpass of dc to 4 megacycles (risetime of approximately 90 nanoseconds). Sensitivity is

variable in nine calibrated steps from 0.05 volt per division to 20 volts per division, and is continuously variable (uncalibrated) between steps and to about 50 volts per division. Accuracy is within 3% at each of the calibrated steps. The input impedance of the module is one megohm paralleled by 47 picofarads at all sensitivity settings.

Operating Instructions

Throughout the instructions that follow, it is assumed, unless otherwise noted, that the Type 3A75 is inserted in the left-hand (Y-axis) opening of a Type 561 Oscilloscope, or in either opening of a Type 565 Oscilloscope, thereby providing vertical deflection of the oscilloscope trace. If it is inserted in the right-hand (X-axis) opening of the Type 561 Oscilloscope, it will provide horizontal deflection of the trace and the instructions must be interpreted accordingly. It is further assumed that there is a time-base module in the X-axis opening of a Type 561 Oscilloscope.

Signal Connections

The signal to be displayed is applied to the INPUT connector on the front panel of the module. For best results, the signal should be applied through a shielded cable, with the shield connected to the chassis of both the oscilloscope and the signal source. When displaying high-frequency signals, it may also be necessary to terminate the shielded cable in its characteristic impedance to prevent resonance effects and "ringing" (high-frequency damped oscillations). All leads should be kept as short as possible, especially in high-frequency work.

High-impedance attenuator probes are available for use with the Type 3A75 Amplifier. These probes reduce the resistive and capacitive loading effect of the amplifier and, at the same time, attenuate the signal to allow display of larger signals than would otherwise be possible. These probes and other accessories are described in the Accessories section of the oscilloscope manual.

Input Coupling Selection

If it is desired to display both the ac and dc components of a signal, place the AC-DC-GND. switch in the DC position. In this case, the position of the trace at any instant is a function of the instantaneous signal voltage with respect to ground.

If you wish to display a relatively small ac signal which is riding on top of a relatively large dc signal, place the AC-DC-GND. switch in the AC position. Then the dc component is blocked by a capacitor in the input circuit, and only the ac component will be displayed. However, when the AC-DC-GND. switch is in the AC position, the lower frequency limit (3-db point) of the amplifier is about 2 cps (0.2 cps if you are using a 10X attenuator probe). Therefore, some low-frequency distortion and loss of amplitude will occur if you display signals with frequency components below this frequency. (This distortion will be noted if you

display the 60-cps calibrator signal without an attenuator probe when the AC-DC-GND. switch is in the AC position.)

Placing the AC-DC-GND. switch in the GND. position grounds the input of the amplifier in the module; it does not ground the applied signal.

Displaying a Signal

To display a signal with the Type 3A75 Amplifier, proceed as follows:

- 1. Apply the signal, preferably through a shielded cable or an attenuator probe, to the INPUT connector.
- 2. Establish a common ground between the chassis of the oscilloscope and the signal source. This can be done by connecting the ground clip of a Tektronix attenuator probe to signal ground.
- 3. Set the AC-DC-GND. switch as desired, in accordance with the previous discussion on Input Coupling Selection.
- 4. Adjust the time-base controls to obtain a stable display of the signal.
- 5. Set the VOLTS/DIV. switch and the POSITION control so that the display is placed as desired on the graticule.

Probe Compensation

If an attenuator probe is used, its capacitance must be compensated to that of the module for best signal response. To compensate a Tektronix attenuator probe, proceed as follows:

- 1. Display several cycles of the oscilloscope Calibrator signal on the screen, using the attenuator probe.
- 2. Adjust the variable capacitor in the body of the probe for best square-wave response as shown in Fig. 2, center.

Gain and DC Balance Adjustments

Any time you move the Type 3A75 Amplifier from one oscilloscope opening to another you must adjust the gain to compensate for differences in deflection plate sensitivities. This may also need to be done if you change the module in the other opening, since the differences in average deflection plate voltages between modules affects the overall deflection sensitivity of the crt. To properly set the gain of the Type 3A75 Amplifier, proceed as follows:

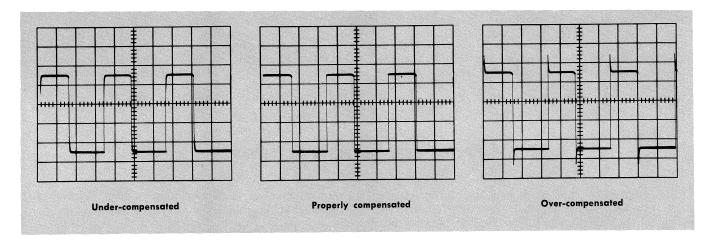


Fig. 2. Probe compensation waveforms.

- 1. Set the AC-DC-GND. switch to DC, the VOLTS/DIV. switch to .05, and the VARIABLE control fully clockwise (to the CALIBRATED position).
- 2. Apply a 0.2-volt Calibrator signal and adjust the front-panel GAIN ADJ. adjustment for a deflection of exactly 4 graticule divisions on the screen; or, display a 0.1-volt Calibrator signal and adjust the front-panel GAIN ADJ. adjustment for a deflection of exactly two graticule divisions. (The former is preferred, but the Type 565 Oscilloscope Calibrator does not produce a 0.2-volt signal.)
- If the dc balance of the module is not properly set, the position of the trace will shift as the VARIABLE control is rotated. To properly adjust the dc balance of the Type 3A75, proceed as follows:
 - 1. Set the AC-DC-GND. switch to GND.
- 2. Set the time-base controls to produce a free-running trace on the screen of the oscilloscope.
- 3. With the POSITION control of the Type 3A75, position the trace to the center of the graticule.
- 4. Set the DC BAL. adjustment to the point where there is no movement of the trace as the VARIABLE control is rotated back and forth between the ends of its range.

Voltage Measurements

NOTE

When making any voltage measurements, make sure there is a common ground between the oscilloscope and the signal source.

To measure the potential difference between two points on a signal (such as peak-to-peak ac volts), measure the vertical distance, in graticule divisions, between the two points and multiply by the setting of the VOLTS/DIV. switch and the attenuation factor, if any, of the probe. Make sure the VARIABLE control is set fully clockwise (to the CALIBRATED position).

To measure the dc level at a given point on a waveform, proceed as follows:

- 1. Set the AC-DC-GND. switch to GND. Make sure the VARIABLE control is set fully clockwise (to the CALIBRATED position).
- 2. Position a free-running trace so that it lies along one of the horizontal graticule lines. This line will be used as a ground reference line. If the voltage you plan to measure is positive, set the ground reference line toward the bottom of the graticule; if the voltage you plan to measure is negative, set it toward the top of the graticule. Do not adjust the POSITION control of the Type 3A75 after the reference line has been established.
- 3. Set the AC-DC-GND. switch to DC. (If the position of the trace shifts by more than 1 minor graticule division, see Troubleshooting, "DC Reference Level Shift.")
- 4. Apply the signal to the INPUT and adjust the timebase controls to obtain a stable display.
- 5. Measure the distance, in graticule divisions, from the ground reference line established in step 2 to the point you wish to measure.
- 6. Multiply this distance by the setting of the VOLTS/DIV. switch and the attenuation factor, if any, of the probe. This is the dc level of the point measured.

You can reestablish your ground reference line at any time merely by placing the AC-DC-GND. switch to GND. and setting the time-base controls for a free-running trace; you do not need to disconnect the probe from the signal source. If desired, you can establish a reference other than ground by setting the AC-DC-GND. switch to DC, connecting the signal probe to the desired reference voltage, and positioning the trace as described in step 2 of the foregoing procedure.

You can adjust the module for a set of deflection factors other than those selected by the VOLTS/DIV. switch, if desired, through the use of the VARIABLE control. This will be a convenience when you wish to compare some voltage or signal to a given reference. To establish a new set of deflection factors, apply the desired reference signal to the INPUT connector, and set the VOLTS/DIV. switch and the VARIABLE control for the desired number of divisions of deflection on the screen. Divide the amplitude of the reference

signal by the number of divisions of deflection, and divide this result by the setting of the VOLTS/DIV. switch. This results in a sensitivity correction factor. You can now obtain the true sensitivity in volts per division at any setting of the VOLTS/DIV. switch by multiplying the switch setting by the sensitivity conversion factor, as long as the VARIABLE control has not been moved from its original setting.

For example, suppose you have applied a reference signal of 32 volts, have set the VOLTS/DIV. switch to 5, and

the VARIABLE control for a deflection of 4 divisions. The true sensitivity at this setting is $32 \text{ volts} \div 4 \text{ divisions}$, or 8 volts per division. The sensitivity conversion factor is 8 volts per division $\div 5 \text{ volts}$ per division, or 1.6. The true sensitivity at any setting of the VOLTS/DIV. switch now is 1.6 times the setting of the switch. For example, when the VOLTS/DIV. switch is set to .1, the true sensitivity is 0.16 volt per division. A signal which produces 6 divisions of deflection at this setting is 0.96 volt in amplitude.

Circuit Description

The Type 3A75 Amplifier is basically a two-stage amplifier. The Input Amplifier, V434 and V534, is a cathode-coupled paraphase amplifier which converts the single-ended input to a push-pull output. The nominal gain of the Input Amplifier is about 24—that is, a single-ended input signal of one millivolt produces a push-pull output signal of 24 millivolts. The Output Amplifier, V454 and V554, is a straight push-pull amplifier with a nominal gain of about 16 to 18, depending upon the setting of the front-panel GAIN ADJ. adjustment. Cathode followers V423 and V443 isolate the stages from the INPUT connector and from each other.

Attenuators

When the gain of the Output Amplifier is properly set, a 0.05-volt signal at the grid of the Input Amplifier (pin 2 of V434) produces one division of deflection on the crt screen. A network of attenuators between the INPUT connector and the Input Amplifier attenuates the applied signal to produce the proper amount of signal at the Input Amplifier. The attenuators are selected by the VOLTS/DIV. switch. If, for example, the VOLTS/DIV. switch is set at 2, the signal is applied through the X10 and X4 attenuator to provide 40X attenuation. Then a 2-volt signal at the INPUT connector will produce 0.05 volt of signal at the Input Amplifier, which will produce 1 division of deflection on the screen.

The attenuators are both resistive and capacitive dividers which, when properly calibrated, provide equal attenuation at all frequencies from dc to over 4 megacycles. They also provide a constant input impedance of one megohm paralleled by 47 picofarads at all settings of the VOLTS/DIV. switch.

Input Amplifier

The Input Amplifier converts the single-ended input signal at the grid of V434 to a push-pull output signal. This is accomplished as follows:

A signal appearing at the grid of V434 tends to produce an in-phase reproduction of the signal at its cathode. Since the cathodes of V434 and V534 are commonly coupled, equal signals appear at both cathodes. The grid of V534, however, is held at a fixed voltage, which results in a 180° phase relationship between the grid-to-cathode signals of V434 and V534—that is, as the grid-to-cathode voltage of V434 moves in a negative direction, the grid-to-cathode

voltage of V534 moves in a positive direction, and vice versa. This, then, results in a push-pull output signal.

The VARIABLE control, R437, decreases the gain of the Input Amplifier by introducing degeneration into the cathode circuit (when the VARIABLE control is set fully clockwise, R437 has zero resistance). The DC BAL. adjustment provides the means of setting the dc level at the cathode of V534 exactly equal to that of V434 when R437 is set above zero resistance. This compensates for any difference in the values of R436 and R536 and in tube characteristics. When the DC BAL. adjustment is properly set, there will be no shift in the position of the trace as the VARIABLE control is rotated.

The input capacitance of V434 changes slightly with the setting of the VARIABLE control. The input cathode follower, V423A, isolates the INPUT connector from the Input Amplifier so that this change in capacitance is not reflected back to the attenuators to upset the high-frequency attenuation.

The OUTPUT PLATE VOLTAGE LEVEL adjustment, R537, by controlling the current through the Input Amplifier tubes, sets the dc levels throughout the amplifier. Although this adjustment has some effect on the gain of the amplifier, this is not its primary purpose. Its primary purpose is to allow for differences in the operating characteristics of the Input Amplifier tubes while maintaining the dc operating level of the Output Amplifier at the center of its dynamic range.

The RC networks R439-C439 and R539-C539 compensate for dc shift, or "slump", in the Input Amplifier tubes. This condition, characteristic to some degree of most pentodes, causes the Gm of the tubes to be slightly less at extremely low frequencies (dc to a fraction of a cycle per second) than at middle and high frequencies. The RC networks reduce the plate loads by about 3% at frequencies over about 1 cps to compensate for the drop in stage gain below this frequency.

L435 and L535 are peaking coils which widen the bandpass of the stage. At the frequencies at which the peaking takes place, C435 and C535 present a very low impedance to the signals, so both L435 and L535 may be considered as single center-tapped coils, with the center taps connected to the plates of the tubes. Thus, at high frequencies, the plate circuit of each amplifier tube looks approximately like Fig. 3.

The LR networks in the plate circuit further improve the high-frequency response.

At low frequencies, the plate circuit of each Input Amplifier tube looks approximately like Fig. 4.

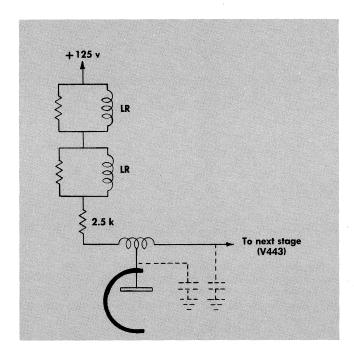


Fig. 3. Equivalent Input Amplifier plate circuit at high frequencies.

At low frequencies, the plate load of each tube is approximately 2.5 k. The drop in signal at the next stage due to the voltage-divider action of the 6.8-k and 390-k resistors is less than 2%. The purpose of the 6.8-k resistor is to allow greater positioning voltage swing (about 10 volts) at the grids of V443 than would otherwise be feasible if the positioning voltages had to be developed across the 2.5-k plate load only.

Output Amplifier

The gain of the Output Amplifier is adjustable by means of the GAIN ADJ., R557, in the common cathode circuit.

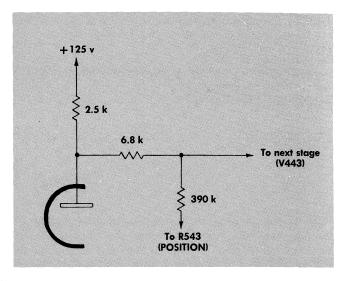


Fig. 4. Equivalent Input Amplifier plate circuit at low frequencies.

This adjustment allows for differences in deflection sensitivities of cathode-ray tubes.

D456 and D457 improve the linearity of the Output Amplifier near the ends of its dynamic range by shunting the cathode-coupling resistor, R457, on large signals. V463A also improves the linearity of the Output Amplifier by maintaining a constant dc level at the screens of V454 and V554 regardless of the screen currents in the tubes. Peaking coils L450, L451, L452, L550, L551, and L552 improve the high-frequency response of the Output Amplifier.

The trigger-pickoff cathode follower, V463B, couples a sample of the signal at the plate of V454 to the time-base circuitry for internal triggering. The signal amplitude at the cathode of V463B is approximately 4 volts for each division of deflection on the crt screen.

Troubleshooting

General maintenance and troubleshooting information is contained in the Type 560-Series Oscilloscope instruction manuals. In the following discussion, it is assumed that you have read that information and have isolated trouble to the Type 3A75 module by procedures described there.

Normally, trouble in the Type 3A75 will produce one of four effects in the oscilloscope display. These are: no display, insufficient deflection, signal distortion, and dc reference level shift. The following troubleshooting information is divided according to these symptoms and will, in most cases, enable you to isolate the trouble to a given stage or portion of the amplifier. When the trouble has been so isolated, you should first change the tube or tubes in that stage. If this does not eliminate the trouble, check the rest of the stage or circuit by voltage and resistance measurements. Typical voltages at various points in the module are shown on the schematic diagram at the rear of this manual. In most cases, resistance measurements will be

point-to-point checks whose values can be approximated from the schematic diagram. You will need to insert the Type 3A75 module in the right-hand oscilloscope opening and remove the right-hand side panel to make the voltage and resistance measurements. It is not necessary to have a module in the other opening to obtain the voltage readings given on the schematic diagram.

No Display

If you cannot position the display onto the screen, there is an imbalance in the module — that is, there is a significant dc voltage difference between the two sides of the push-pull circuits. First, set the POSITION control to midrange and check the setting of the DC BAL. adjustment according to the procedure described in the Operating Instructions. If this does not eliminate the trouble, set the

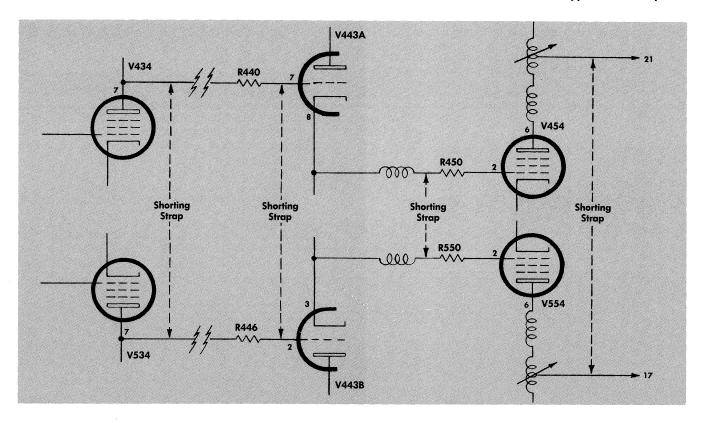


Fig. 5. Shorting points for isolating dc imbalance.

POSITION control to midrange and, with an insulated shorting strap, short between the pairs of points shown in Fig. 5, starting at the right and working toward the left. As you short between each pair of points, you should obtain a spot or trace on the screen. When you short between a pair of points and do not obtain a spot or trace, the trouble is between that pair of points and the next pair to the right.

If you still do not obtain a spot or trace when you short between the plates of V434 and V534, the trouble is between there and the input (or in the dc balance circuit). Due to a possible mismatch between V434 and V534, you may not obtain a spot or trace on the screen when you short their grids together even if the two tubes are good. Therefore, you must check the rest of the circuit by comparing voltages at corresponding points in the two sides of the circuit. If any of the voltages differ significantly from each other or from those shown on the schematic diagram, the trouble is in that portion of the circuit.

Insufficient Deflection

If insufficient deflection is noted in only certain positions of the VOLTS/DIV. switch, the trouble is a defective component in one of the attenuation networks or in the switch. If insufficient deflection is noted in all positions of the VOLTS/DIV. switch, the trouble is probably in one of the amplifier stages. In this case, rotate the POSITION control. If normal positioning range exists, the trouble is in the Input Amplifier or in the Input Cathode Followers.

Signal Distortion

Signal distortion — the amplification of some frequencies more than others — is almost always caused by faulty tubes (assuming the module is properly calibrated and the signal probe is properly compensated).

High-frequency distortion — where the high frequencies are attenuated or amplified more or less than the middle and low frequencies — will usually show up as undershoot or overshoot on a square wave, as shown in the "Undercompensated" and "Over-compensated" waveforms in Fig. Low-frequency distortion — where the low frequencies are attenuated or amplified more or less than the middle and high frequencies — will usually show up as a slope in the top and bottom of a square wave. A certain amount of high-frequency distortion is normal beyond the specified upper-frequency limit of the amplifier, and a certain amount of low-frequency distortion is normal when you are displaying waveforms with very-low-frequency components and have the AC-DC-GND. switch in the AC position. It is only when this distortion becomes excessive in the normal frequency range of the instrument that it constitutes a trouble.

Low-frequency distortion is usually caused by a change in the time constant of the input coupling circuit. This, in turn, is usually caused by a tube becoming gassy and drawing grid current. Therefore, when low-frequency distortion is encountered, check all the tubes in the module (V423, V434, and V534 first).

In all cases of high-frequency distortion, you should first check the compensation of the probe, if one is used, and

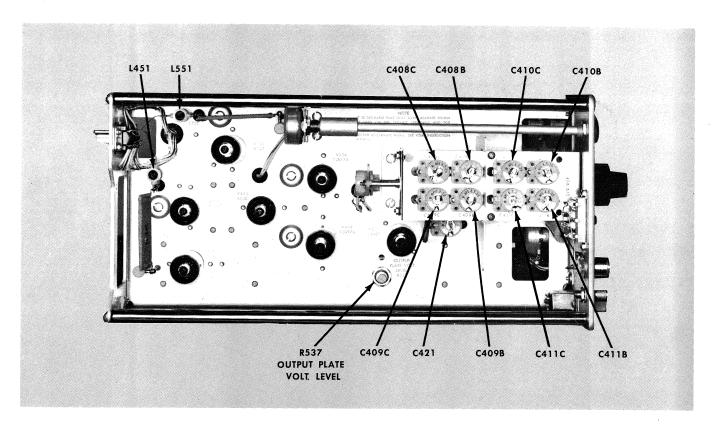


Fig. 6. Left side of Type 3A75 Amplifier showing location of internal adjustments.

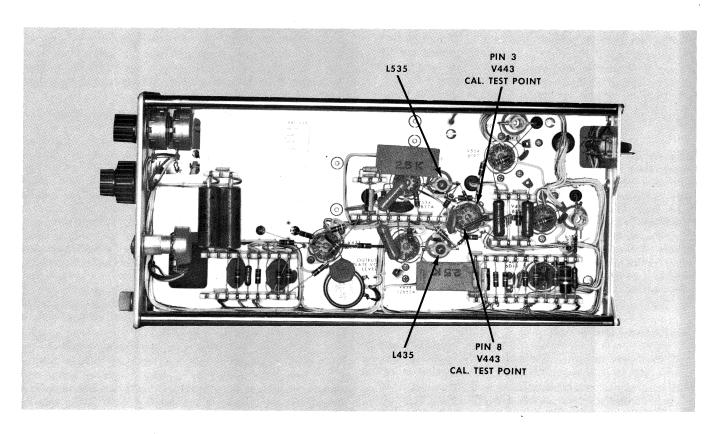


Fig. 7. Right side of Type 3A75 Amplifier showing location of internal adjustments.

then check the frequency compensation of the attenuators and the amplifier stages, as described in the Calibration section of this manual. If this does not cure the trouble, check all the tubes in the module.

DC Reference Level Shift

If, when the VOLTS/DIV. switch is set at .05, the position of a no-signal trace (or spot) changes by more than 1 minor graticule division as the AC-DC-GND. switch is moved

from GND. to DC, V423A is probably gassy and should be replaced.

If this condition occurs and cannot be immediately corrected, satisfactory measurement of dc levels can be accomplished by establishing the zero reference line with the AC-DC-GND. switch in the DC position and the signal probe connected to signal ground (see Operating Instructions, "Voltage Measurements"). In this case, you should establish the zero reference line each time you move the VOLTS/DIV. switch.

Calibration

To maintain the high degree of accuracy of the Type 3A75 Amplifier, we recommend that it be calibrated after each 500 hours of operation or about every six months, whichever is sooner. It should not require more frequent calibration. However, if tubes or other components are replaced, the calibration of the instrument should be checked and adjusted as necessary.

Calibration of the Type 3A75 is performed with the module inserted in the left-hand opening of a Type 561 or Type 565 Oscilloscope. If you are using a Type 561 Oscilloscope, you will need to have a time-base module in the right-hand opening. Complete calibration consists of the OUTPUT PLATE VOLTAGE LEVEL adjustment, standardization of the input capacity, frequency compensation of the attenuator circuits, and frequency compensation of the amplifier stages. Any of the adjustments may be performed individually or out of sequence as long as all of the steps in a given procedure are performed.

Figs. 6 and 7 show the locations of the internal adjustments referred to in the calibration procedures.

Equipment Required

The following equipment is required for a complete calibration of the Type 3A75 Amplifier.

- 1. DC voltmeter capable of measuring 180 volts.
- 2. Square-wave generator with a risetime of 45 nanoseconds or less and frequency outputs of about 1 kc and 400 kc; output amplitude should be variable up to at least 50 volts peak-to-peak (Tektronix Type 105 Square-Wave Generator recommended).
- 3. Coaxial cable with matching termination resistor (Tektronix Type P52 coaxial cable and 50-ohm terminating resistor recommended).
- 4. 47-picofarad capacitance standardizer (Tektronix Type CS 47 recommended), or a capacitance meter capable of measuring 47 picofarads (Tektronix Type 130 L-C Meter recommended) and an attenuator probe with adjustable capacitance.
- 5. Test oscilloscope with a sensitivity of at least 50 millivolts per division and a sweep rate of 1 microsecond or less per division (Tektronix Type 530-, 540-, 550-, or 580-Series Oscilloscope with suitable plug-in unit recommended.)

6. Low-capacitance calibration tools (Tektronix part no. 003-000 and 003-301 recommended).

Preliminary Setup

Insert the Type 3A75 into the left-hand opening of the oscilloscope and, after warm-up, set the GAIN ADJ. and DC BAL. adjustments as described in the Operating Instructions of this manual.

Output Plate Voltage Level Adjustment

Center a free-running trace or spot vertically on the screen and short the vertical deflection plates of the cathode-ray tube together.

CAUTION

Be careful not to short the deflection plates to the metal shield around the crt as it could damage components in the oscilloscope.

Connect the dc voltmeter between ground and the shorted deflection plate pins. Adjust the OUTPUT PLATE VOLTAGE LEVEL adjustment for a reading of +180 volts on the meter. Disconnect the meter and the shorting strap.

Amplifier Frequency Compensation

Proper frequency compensation of the amplifier is important when the module is to amplify frequencies near the upper limit of its bandpass (4 mc). Improper frequency compensation in the amplifier will normally be indicated by poor response at the high frequencies. Unless the need for amplifier compensation has been established, it is best not to attempt to make these adjustments. Set the OUTPUT PLATE VOLTAGE LEVEL adjustment before performing this procedure.

To frequency compensate the amplifier stages of the Type 3A75, proceed as follows:

- 1. Remove the Type 3A75 from the oscilloscope and preset L435, L535, L451, and L551 so that the tops of the slugs are flush with the tops of the plastic coil forms.
- 2. Reinsert the Type 3A75 and set the VOLTS/DIV. switch to .05 and the AC-DC-GND. switch to DC.

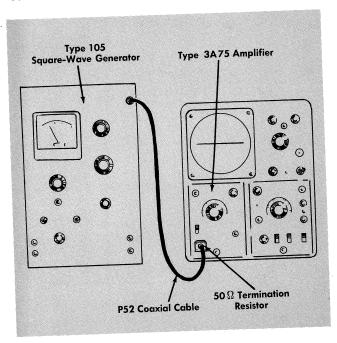


Fig. 8. Tektronix Type 105 Square-Wave Generator connected to the Type 3A75 for amplifier frequency compensation.

- 3. Connect the 400-kc output of the square-wave generator through the coaxial cable and matching termination resistor to the INPUT connector of the Type 3A75. (Fig. 8 shows the Tektronix Type 105 Square-Wave Generator properly connected to the Type 3A75 for this step.)
- 4. Set the time-base controls to display several cycles of the square wave, and adjust the output amplitude of the square-wave generator for four or five divisions of deflection on the screen. (You may have to adjust the VARIABLE control on the Type 3A75 module to obtain the proper deflection.) The display will be an overpeaked square wave.
- 5. Center the display on the screen with the POSITION controls.

- 6. Alternately adjust L451 and L551 in equal increments until the most evenly rounded leading corners on the square-wave are obtained (see Fig 9). L451 and L551 both will probably need to be turned about six complete clockwise turns before the proper display is obtained; both slugs should be at very nearly the same depth at the conclusion of this step.
- 7. Insert the Type 3A75 module into the right-hand oscilloscope opening. (Maintain the same signal connections between the Type 3A75 module and the square-wave generator.)
- 8. Connect the probe of the test oscilloscope to pin 3 of V443 and note the "ringing waveform" that is superimposed on the leading corners of the square wave as displayed on the test oscilloscope.
- 9. Adjust L535 for minimum ringing and optimum squarewave appearance on the positive half-cycles in the test oscilloscope display (see Fig. 10).
- 10. Connect the probe of the test oscilloscope to pin 8 of V443, and adjust L435 for minimum ringing and optimum square-wave appearance on the negative half-cycles in the test oscilloscope display (see Fig. 11). There is slight interaction between steps 9 and 10; they should be repeated to obtain the optimum settings.
- 11. Disconnect the test oscilloscope probe and move the Type 3A75 module to the left-hand oscilloscope opening. (Maintain the same signal connections between the Type 3A75 and the square-wave generator.)
- 12. Again adjust the time-base controls as necessary to obtain a stable display.
- 13. The displayed waveform on the oscilloscope screen should be a square wave with slightly rounded leading corners. If they are a bit underpeaked or overpeaked, adjust L451 and L551 in small and equal increments for the most evenly rounded leading corners; see Fig. 9 (a). Very little adjustment, if any, of L451 and L551 should be necessary.

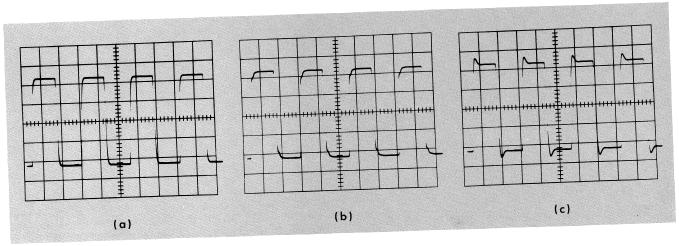


Fig. 9. Amplifier frequency compensation. (a) L451 and L551 properly set; (b) and (c), L451 and L551 improperly set.

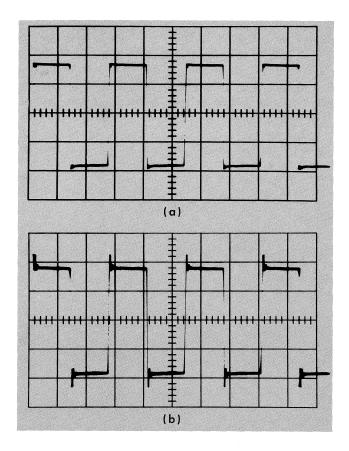


Fig. 10. Amplifier frequency compensation (waveform at pin 3, V443). (a) L535 properly set — minimum ringing and optimum square-wave appearance on positive half-cycles. (b) L535 improperly set — excessive ringing on waveform.

14. Disconnect the square-wave generator and the termination resistor from the Type 3A75 module.

Input Capacitance Standardization and Attenuator Frequency Compensation

The input capacitance standardization and attenuator frequency compensation interact so both must be performed in the same procedure. The need for either or both of these calibrations is normally indicated by distortion of fast-rising waveforms on one or more sensitivity ranges of the module.

Standardization of the input capacitance of the module requires the use of a 47-picofarad capacitance standardizer. If you do not have such a standardizer, you can make an attenuator probe into one by performing these five steps. (If you have a 47-picofarad standardizer, you may skip these five steps and go immediately to the procedure following them.)

- 1. Set the VOLTS/DIV. switch to .05 and the AC-DC-GND. switch to DC.
- 2. With the oscilloscope turned on, measure the input capacitance at the INPUT connector. With a low-capacitance screwdriver, adjust C421 for a reading of 47 picofarads on the capacitance meter.

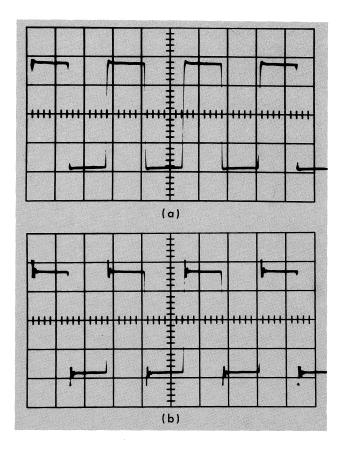


Fig. 11. Amplifier frequency compensation (waveform at pin 8, V443). (a) L435 properly set — minimum ringing and optimum square-wave appearance on negative half-cycles. (b) L435 improperly set — excessive ringing on waveform.

- 3. Connect the attenuator probe to the INPUT connector and touch the probe tip to the 1-kc output of the square-wave generator.
- 4. Adjust the oscilloscope controls to display several cycles of the square-wave signal.
- 5. Adjust the variable capacitor in the body of the probe to obtain the best square-wave response (no undershoot or overshoot).

Your probe is now standardized to perform as a 47-picofarad capacitance standardizer. You may use it as such in the following procedure to set the input capacitance of the module on all of the other sensitivity ranges. (Do not adjust the probe further during the procedure.)

To properly set the input capacitance and attenuator frequency compensation of the Type 3A75 Amplifier, proceed as follows:

1. Connect the 1-kc output of the square-wave generator through the coaxial cable and the <u>47-picofarad capacitance</u> standardizer (or the standardized probe) to the INPUT connector of the Type 3A75 module. (Fig. 12 shows the Tektronix Type 105 Square-Wave Generator properly connected to the Type 3A75 for this step.)

Type 3A75 Amplifier

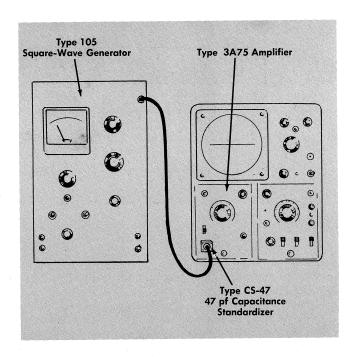


Fig. 12. Tektronix Type 105 Square-Wave Generator connected to the Type 3A75 Amplifier for input capacitance standardization.

2. Set the VOLTS/DIV. switch on the Type 3A75 to .05 and the AC-DC-GND. switch to AC.

- 3. Adjust the time-base controls on the oscilloscope to display several cycles of the square wave. You may use the VARIABLE control to get the entire signal on the screen, if necessary.
- 4. Adjust C421 for the best square-wave response. (If you used this module to standardize the probe, you will not need to adjust C421.)
- 5. Set the VOLTS/DIV. switch to each of the positions shown in the first column of Table 1, and adjust the corresponding capacitors for the best square-wave response at each setting. As you move the VOLTS/DIV. switch, adjust the output amplitude of the square-wave generator to maintain at least four divisions of deflection on the screen, if possible.

You may adjust the capacitors listed at any VOLTS/DIV. setting given in Table 1 without upsetting the adjustment of those listed at any other setting; there is no interaction among them. However, if you adjust C421, you will need to readjust all of the capacitors listed in the table.

TABLE 1

VOLTS/DIV. Setting	Adjust
.1	C408B, C408C
.2	C409B, C409C
.5	C410B, C410C
5	C411B, C411C

6. Disconnect the square-wave generator and the capacitance standardizer from the oscilloscope.

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PARTS LIST AND DIAGRAM

PARTS ORDERING INFORMATION

Replacement parts are available from or through your local Tektronix Field Office.

Changes to Tektronix instruments are sometimes made to accommodate improved components as they become available, and to give you the benefit of the latest circuit improvements developed in our engineering department. It is therefore important, when ordering parts, to include the following information in your order: Part number including any suffix, instrument type, serial number, and modification number if applicable.

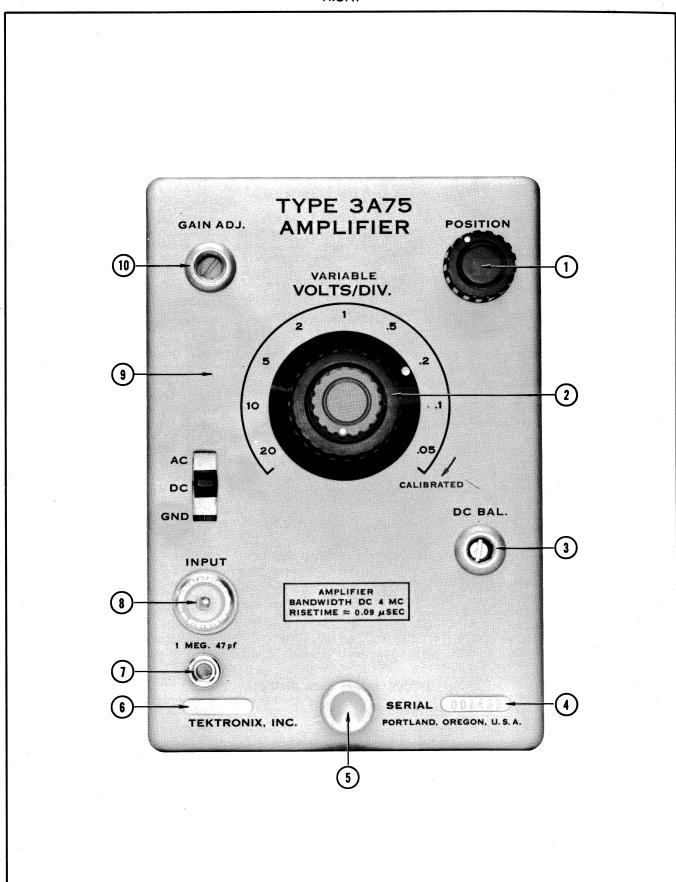
If a part you have ordered has been replaced with a new or improved part, your local Tektronix Field Office will contact you concerning any change in part number.

ABBREVIATIONS AND SYMBOLS

SPECIAL NOTES AND SYMBOLS

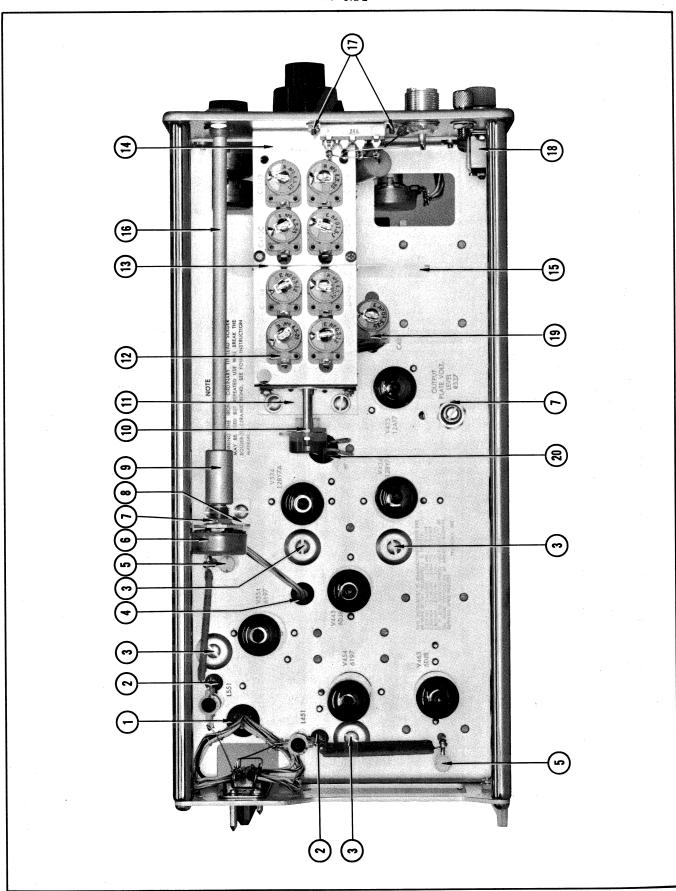
X000	Part first added at this serial number.
000X	Part removed after this serial number.
*000-000	Asterisk preceding Tektronix Part Number indicates manufactured by or for Tektronix, or reworked or checked components.
Use 000-000	Part number indicated is direct replacement.
	Internal screwdriver adjustment.
	Front-panel adjustment or connector.

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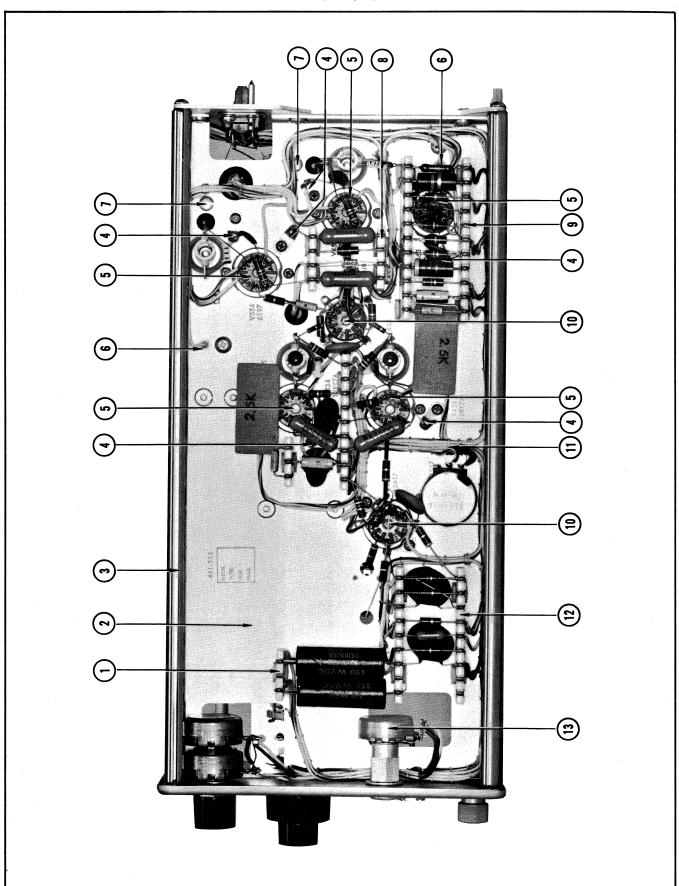
FRONT

REF.	PART	SERIAL	NO.	0	
NO.	NO.	EFF.	DISC.	Y.	DESCRIPTION
2 3 4 5 6 7	366-044 366-113 210-419 210-902 210-012 366-058 366-144 366-038 358-010 210-013 210-494 334-679 366-109 334-829 129-035	101 1120 101 1120	1119	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	KNOB, small black, POSITION KNOB, small charcoal, POSITION NUT, shoulder, ³ / ₈ -32 WASHER, flat, .470 ID x ²¹ / ₃₂ OD LOCKWASHER, int. ³ / ₈ x ¹ / ₂ KNOB, large black, VOLTS/DIV. KNOB, large charcoal, VOLTS/DIV. KNOB, small red, VARIABLE BUSHING, ³ / ₈ -32 x ⁹ / ₁₆ LOCKWASHER, int, ³ / ₈ x ¹¹ / ₁₆ NUT, hex, ³ / ₈ -32 x ¹ / ₂ TAG, metal serial no. insert KNOB, plug-in securing TAG, metal blank mod. insert POST, binding, assembly
8 9	355-507 200-103 210-455 210-011 131-081 210-241 333-621 333-729 387-577 358-010 210-413	101 1120	1119	1 1 1 1 1 1 1 1 1	Consisting Of: STEM, adapter CAP NUT, hex, ½-28 x ³½ LOCKWASHER, int, ½ CONNECTOR, chassis mt., coaxial, 1 contact, female LUG, ground, 15½ long PANEL, front (type 75) PANEL, front (type 3A75) PLATE, front subpanel BUSHING, ¾-32 x ½ NUT, hex, ¾-32 x ½



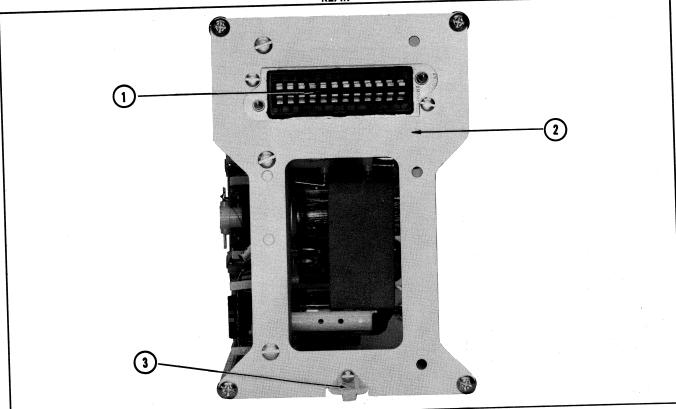
LEFT SIDE

REF.	PART	SERIA	L NO.	0	7		
NO.	NO.	EFF.	DISC.	Y.	DESCRIPTION		
1	348-005			1	GROMMET, rubber, ½		
2	348-002			2	GROMMET, rubber, 1/4		
3	211-507			4	SCREW, 6-32 x 5/16 BHS (coil mounting)		
	210-006			4	LOCKWASHER, int. #6		
4	348-003			1	GROMMET, rubber, 5/16		
5	385-040			2	ROD, nylon		
					Mounting Hardware For Each: (not included)		
	213-041			!	SCREW, thread cutting, 6-32 x 3/8 Truss HS phillips		
6	200-247			1	CAP, pot, poly.		
7	210-413 210-840			2 2	NUT, hex, $\frac{3}{8}$ -32 x $\frac{1}{2}$ (pot mounting)		
8	406-684		1	1	WASHER, .390 ID x ½ 6 OD BRACKET, gain adjust., pot		
U	400-004			'	Mounting Hardware: (not included)		
	211-504			2	SCREW, 6-32 x 1/4 BHS		
9	376-007			l ī	COUPLING		
					Includes:		
	213-005			2	SCREW, set, 8-32 x $\frac{1}{8}$ HSS		
10	262-482			1	SWITCH, VOLTS/DIV., wired		
					Includes:		
	260-158			1 1	SWITCH, VOLTS/DIV., unwired		
11	210-406			2	NUT, hex, $4-40 \times \frac{3}{16}$ (pot mounting)		
11	406-683		* .	1	BRACKET, switch support		
	211-504			2	Mounting Hardware: (not included) SCREW, 6-32 \times $\frac{1}{4}$ BHS		
	210-803			2	WASHER, 6L		
12	214-153			8	FASTENER, snap, double pronged		
13	337-423			1	SHIELD, attenuator bracket		
					Mounting Hardware:		
	213-035			2	SCREW, thread cutting, 4-40 x 1/4 PHS phillips		
14	406-610		*	1	BRACKET, attenuator mtg.		
	011 000		14		Mounting Hardware:		
	211-008			2	SCREW, 4-40 x 1/4 BHS		
	210-004 210-406			2 2	LOCKWASHER, int. #4		
15	337-424				NUT, hex, 4-40 x ³ / ₁₆ SHIELD, attenuator switch		
	007 -424		1	'	Mounting Hardware:		
	211-008			2	SCREW, $4-40 \times \frac{1}{4}$ BHS		
	210-004			2	LOCKWASHER, int. #4		
	210-406			2	NUT, hex, $4-40 \times \frac{3}{16}$		
					Mounting Hardware For Switch: (not included)		
	210-413]	NUT, hex, $\frac{3}{8}$ -32 x $\frac{1}{2}$		
16 17	384-215			1 1	ROD, extension		
18	210-406 214-052			2	NUT, hex, 4-40 x ³ / ₁₆ (slide switch mounting)		
10	214-032		1	'	FASTENER, pawl right, w/stop Mounting Hardware: (not included)		
	210-004	-		2	LOCKWASHER, int. #4		
	210-406			2	NUT, hex, 4-40 x ³ / ₁₆		
19	214-153			1	FASTENER, snap, double pronged		
20	348-004			1	GROMMET, rubber, 3/8		
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RIGHT SIDE

REF.	PART	SERIA	L NO.	RIGHT SIDE							
NO.	NO.	EFF.	DISC.	T Y.	DESCRIPTION						
1 2 3 4 5	124-087 361-007 441-334 211-504 211-538 384-566 210-201 213-044 136-072 213-044			2 2 1 3 2 4 6 6 5	STRIP, ceramic, $\frac{3}{4} \times 3$ notches SPACER, nylon, molded CHASSIS Mounting Hardware: (not included) SCREW, 6-32 × $\frac{1}{4}$ BHS SCREW, 6-32 × $\frac{5}{16}$ FHS 100° phillips slot ROD, frame, spacing LUG, solder, SE 4 SCREW, thread cutting, 5-32 × $\frac{3}{16}$ PHS phillips SOCKET, 9 pin UHF miniature Mounting Hardware For Each: (not included)						
6 7 8 9 10	348-031 211-504 124-088 361-007 124-091 361-007 136-015 213-044 179-509 179-739 124-090	101 1340	1339	2 2 2 4 2 4 2 2 1 1	SCREW, thread cutting, 5-32 x $^3/_{16}$ PHS phillips GROMMET, poly snap-in SCREW, 6-32 x $^1/_4$ BHS (coil mounting) STRIP, ceramic, $^3/_4$ x 4 notches SPACER, nylon, molded STRIP, ceramic, $^3/_4$ x 11 notches SPACER, nylon, molded SOCKET, STM9G Mounting Hardware For Each: (not included) SCREW, thread cutting, 5-32 x $^3/_{16}$ PHS phillips CABLE, harness CABLE, harness						
13	361-007 200-247			3 6 1	STRIP, ceramic, 3/ ₄ x 9 notches SPACER, nylon, molded CAP, pot, poly.						
·											



REF.	PART	SERIA	L NO.	Q	DESCRIPTION
NO.	NO.	EFF.	DISC.	Y.	
1	131-149			1	CONNECTOR, chassis mt., 24 contact, male
					Mounting Hardware: (not included) SCREW, 4-40 x 1/4 BHS
	211-008 210-004			2 2	LOCKWASHER, int. #4
	210-004			2	NUT, hex, 4-40 x ³ / ₁₆
2	387-581			1	PLATE, rear Mounting Hardware: (not included)
	222044		1	4	SCREW, 8-32 x 1/2 RHS
3	212-044 351-037			1	CUIDE plug-in
3	331-03/				Mounting Hardware: (not included)
	211-013			1 1	SCREW, 4-40 x 3/8 FHS LOCKWASHER, int. #4
	210-004				NUT, hex, 4-40 x 3 / ₁₆
	210-406				
		l .			
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PARTS LIST TYPE 3A75

Values are fixed unless marked variable.

Capacitors

Tolerance ±	±20% unless other	wise indicated.					
Ckt. No.	C/N. D						Tektronix
	S/N Range		Description				Part Number
C400 C407 C408B C408C		.1 μf 270 pf 3-12 pf 4.5-25 pf	PTM Cer.	Var. Var.	600 v 500 v	10%	285-587 281-543 281-007 281-010
C408D		4.7 pf	Cer.		500 v	±1 pf	281-501
C409B C409C C410B C410C C410E		4.5-25 pf 3-12 pf 4.5-25 pf 4.5-25 pf 150 pf	Mica	Var. Var. Var. Var.	500 v	10%	281-010 281-007 281-010 281-010 283-508
		•			300 V	10 /6	203-300
C411A C411B C411C C411E		12 pf 4.5-25 pf 1.5-7 pf	Cer.	Var. Var.	500 v	±1.2 pf	281-505 281-010 281-005
C420		250 pf .02 μf	Mica Discap		500 v 600 v	5%	283-543 283-006
C421 C426 C435 C439 C441		4.5-25 pf .01 μf 390 pf 20 μf .02 μf	Discap Cer. EMT Discap	Var.	150 v 500 v 150 v 600 v		281-010 283-003 281-551 290-008 283-006
C453 C460 C464 C468 C471		.01 μf .005 μf 1.5 pf .02 μf .1 μf	Discap Discap Cer. Discap Discap		150 v 500 v 500 v 150 v 500 v	±0.5 pf	283-003 283-001 281-526 283-004 283-008
C473 C475 C477 C479 C530		.01 μf .02 μf .1 μf .01 μf .01 μf	Discap Discap Discap Discap Discap		500 v 600 v 500 v 150 v 100 v		283-002 283-006 283-008 283-003 283-012
C535 C539 C552 C553		390 pf 20 μf 1.5 pf .01 μf	Cer. EMT Cer. Discap		500 v 150 v 500 v 150 v	±0.5 pf	281-551 290-008 281-526 283-003

Diodes

D 157	Germanium Germanium			152-007 152-007
D-137	Germanium DI	DR 746	152-007	

Inductors

			Tektronix Part Number
L435	A 12-21 μh B13-24 μh	Var.	*114-139
L450	4.5 μh		*108-231
L451	180-350 μ h	Var.	*114-138
L452	$13~\mu h$		*108-233
L535	A 12-21 μh B13-24 μh	Var.	*114-139
L550	4.5 μh		*108-231
L551	180-350 μh	Var.	*114-138
L552	13 μh		*108-233
LR432	9.25 μh		*108-232
LR433	3.25 μh		*108-230
LR532	9.25 μh		*108-232
LR533	$3.25\stackrel{\cdot}{\mu}$ h		*108-230

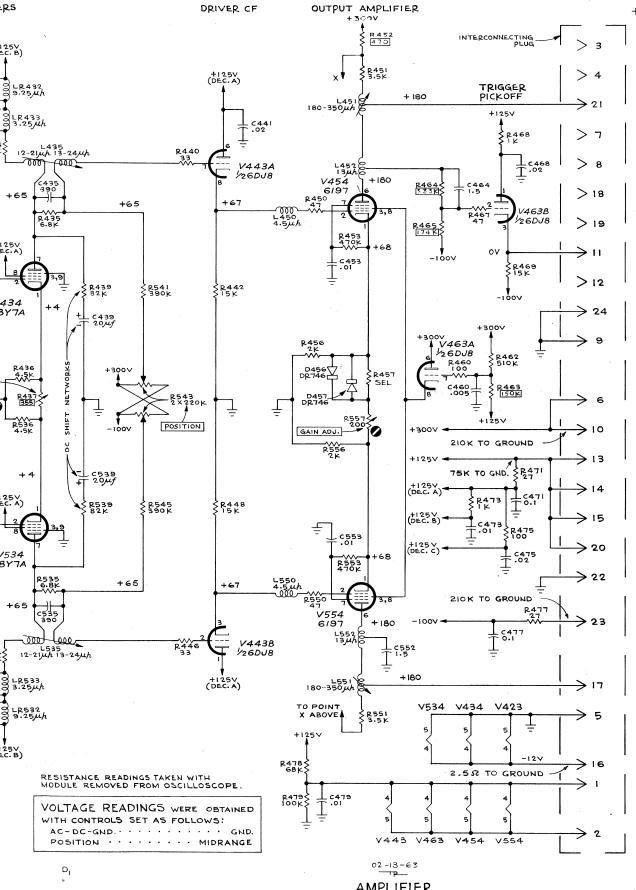
Resistors

Resistors are fixed, composition, $\pm 10\%$ unless otherwise indicated.

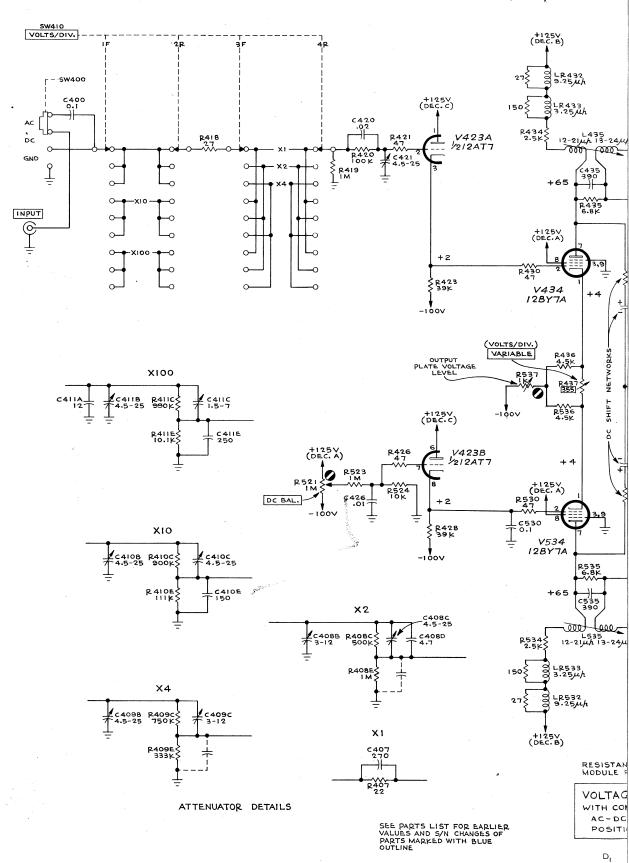
R407 R408C R408E R409C R409E		22 Ω 500 k 1 meg 750 k 333 k	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w		Prec. Prec. Prec. Prec.	1% 1% 1% 1%	302-220 309-140 309-148 309-141 309-139
R410C R410E R411C R411E R418		900 k 111 k 990 k 10.1 k 27 Ω	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w		Prec. Prec. Prec. Prec.	1% 1% 1% 1%	309-142 309-138 309-145 309-135 302-270
R419 R420 R421 R423 R426		1 meg 100 k 47 Ω 39 k 47 Ω	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w		Prec.	1%	309-148 302-104 302-470 302-393 302-470
R428 R430 R434 R435 R436		39 k 47 Ω 2.5 k 6.8 k 4.5 k	1/2 w 1/2 w 5 w 1/4 w 5 w		Mica Pl. WW	1% 5% 5%	302-393 302-470 *310-522 315-682 308-092
R437 R439 R440 R442 R446		355 Ω 82 k 33 Ω 15 k 33 Ω	½ w ¼ w ½ w ¼ w	Var.	ww	VARIABLE	Use *311-290 302-823 316-330 302-153 316-330
R448 R450 R451 R452 R453 R456	Х1340-ир	15 k 47 Ω 3.5 k 470 Ω 470 k 2 k	1/2 w 1/2 w 10 w 2 w 1/2 w 5 w	Mica Pl.	ww	1% 5% 5%	302-153 302-470 *310-584 305-471 302-474 308-091

Resistors (continued)

							Tektronix Part Number
R457 R457 R460 R462 R463 R463	101-1339 1340-up 101-1339 1340-up	33 Ω 27 Ω 100 Ω 510 k 240 k 150 k		ted nominal ted nominal		5% 5% 5%	301-330 301-270 302-101 301-514 301-244 301-154
R464 R464 R465 R465 R467 R468	101-1339 1340-up 101-1339 1340-up	400 k 333 k 180 k 174 k 47 Ω 1 k	1/4 w 1/4 w 1/2 w 1/2 w 1/2 w 1/2 w		Prec. Prec. Prec. Prec.	1% 1% 1% 1%	319-004 319-072 309-279 309-151 302-470 302-102
R469 R471 R473 R475 R477 R478		15 k 27 Ω 1 k 100 Ω 27 Ω 68 k	2 w 1/2 w 3 w 1/2 w 1/2 w 1/2 w		ww		306-153 302-270 308-077 302-101 302-270 302-683
R479 R521 R523 R524 R530 R534		100 k 1 meg 1 meg 10 k 47 Ω 2.5 k	1/2 w 1/2 w 1/2 w 1/2 w 5 w	Var.	Mica Pl.	DC BAL.	302-104 311-184 302-105 302-103 302-470 *310-522
R535 R536 R537 R539 R541		6.8 k 4.5 k 1 k 82 k 390 k	1/4 w 5 w 1/2 w 1/2 w	Var.	WW .	5% 5% Output Pl. V.	315-682 308-092 Level 311-006 302-823 301-394
R543 R545 R550 R551 R553		2 × 220 k 390 k 47 Ω 3.5 k 470 k	1/ ₂ w 1/ ₂ w 10 w 1/ ₂ w	Var. Mica Pl.		POSITION 5% 1%	311-031 301-394 302-470 *310-584 302-474
R556 R557		2 k 200 Ω	5 w	Var.	ww	5% GAIN ADJ.	308-091 311-178
			Switch	es			
SW400 SW400 SW410	101-1119 1120-up	Slide Slide Rotary	AC/DC AC/DC VOLTS/DIV				Jnwired Wired 260-251 260-450 260-158 *262-482
		,	Vacuum T	ubes			
V423 V434 V443 V454 V463		12AT7 12BY7 6DJ8 6197 6DJ8					154-039 154-047 154-187 154-146 154-187
V534 V554		12BY7 6197					154-047 154-146



AMPLIFIER



MANUAL CHANGE INFORMATION

At Tektronix, we continually strive to keep up with latest electronic developments by adding circuit and component improvements to our instruments as soon as they are developed and tested.

Sometimes, due to printing and shipping requirements, we can't get these changes immediately into printed manuals. Hence, your manual may contain new change information on following pages. If it does not, your manual is correct as printed.