component news

Dec. 20, 1978

COMPANY CONFIDENTIAL

Issue 266

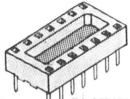
IC sockets: An asset or a liability?

Tek uses an extremely large volume of IC sockets each year. In the last few months, however, major reliability problems have caused us to seriously re-evaluate our need for these parts.

advantages in using IC sockets

IC sockets offer easier manufacturability and prevent large ICs from breaking due to board warpage. The possibility of multilayer board damage from careless desoldering of ICs is lessened. Field serviceability is improved using sockets, as well as providing the field and our customers the opportunity to upgrade their instruments.

Unfortunately, there are numerous disadvantages associated with IC sockets which tend to outweigh the benefits gained from their use.



disadvantages of IC sockets

T.I. C93-XX-02

Most of the problems encountered in using IC sockets has been in the area of manufacturability. For example, the JEDEC specification on IC leads is vague and ambiguous. Under the spec the lead thickness is allowed to vary by a factor of two.

The DIP lead frame has an unspecified base material, and plating is likewise not specified. As a result, IC leads have come into Tek with three platings – tin, silver and gold. The same device, if second-sourced, might come in with two different platings. The size of the burrs left on the IC leads after cutoff from the carrier strips is not specified. One manufacturer's tolerance on the burrs exceeds 50% of the lead thickness. Burrs and sharp edges on the leads remove significant amounts of plating, exposing the base material to corrosion and failure under adverse environmental conditions.

high insertion force required

High insertion force is needed to make good contact between IC and socket. The majority of ICs have tin leads and a high normal force is required to break through the surface oxide films and maintain contact integrity. This high insertion force leads to damaged and bent IC leads in manufacturing areas. Also, many large ICs are broken during removal, due to the high withdrawal force required.

Damage to the IC socket contact occurs when DIP adapters or probes are inserted in the sockets. These devices "pre-size" the contacts, lowering the insertion force and therefore lowering the overall reliability of the instrument. No device that exceeds the dimensions of the IC lead should ever go into an IC socket.

continued on page 2

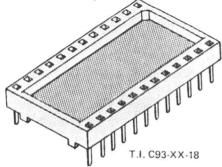
Also in this issue

variable:	Se	ele	ec	ct	ic	n	g	Ju	id	le	•	•	•	•	•	•) -1	0
CMOS test																				
Hex buffer		•	•								•	•	•	•					. 1	1
ROM, 64K																				
Standards,																				

other reliability considerations

IC sockets are susceptible to flow solder and wash procedures. Flux contamination from poor washing reduces the long-term reliability of the connection. The socket housing often hinders or prevents effective washing and hides the contaminants, making inspection for residues very difficult.

Sockets obstruct airflow in densely packed instruments. Also heat conduction to the circuit board is greatly reduced by connections to the IC socket. The resultant higher heat in the IC further lowers instrument reliability.



In serviceability, IC sockets encourage "shotgun" troubleshooting. 40-60% of the failed ICs returned to Beaverton still test good. This either attests to the poor reliability of IC sockets or to poor or incomplete service procedures.

The use of sockets for upgradeability of instruments in the field also brings about the ability for our customers to upgrade their instruments without purchasing the devices from Tek. Field Service has difficulty determining what has been purchased from Tek, and the result is lost revenue!

Finally, there is a higher manufacturing cost incurred because IC sockets are relatively expensive. There is no way to save labor dollars because IC sockets cannot be machine-inserted. At present, Incoming Inspection has no way to test IC sockets for quality or reliability other than a visual inspection.

a solution to the problem

Considerable cost savings and gain in reliability could be achieved by eliminating the use of IC sockets in the majority of applications. Enhanced manufacturability would result from the use of automated IC insertion directly into the board. The use of tri-state devices (to disable outputs during test) and techniques like "signature analysis" could replace the requirement for removing ICs in manufacturing board test.

I recommend that we eliminate the use of IC sockets wherever possible, and institute plans for soldering-in integrated circuits.

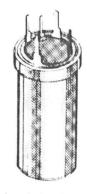
Peter Butler, ext. 5417 Electromechanical Component Engineering

Mallory drops aluminum electrolytic cap line

We currently part-number five 500 WVDC aluminum electrolytic capacitors. Mallory has just announced that they will no longer produce these large, can-type capacitors. This leaves Tek with only one approved supplier. Therefore, the following 500V aluminum electrolytics are not recommended for new design:



290-0028-00 290-0143-00 290-0150-00 290-0262-00 290-0668-00



twist-mount

printed circuit mount

We also part-number three 475 WVDC parts and they will probably suffer the same fate in the near future. If there are any questions or problems contact Don Anderson (58-299), ext. 5415, or Harry Tanielian (19-194), ext. 6405.

page 2

64K ROM performance compared

ne advent of more complex memory systems at Tek has generated interest in the 64K ROM to fill bry requirements. To meet the needs of various user groups, Memory Component Engineering has studying these devices as they become available from semiconductor manufacturers.

-devices under test -

ne first 64K ROM samples investigated were secured under a contract between Tektronix and AMI or number Am4264). These were the first large memory devices to use VMOS (V-groove MOS) techy and utilize a fully static design. The parts were later redesigned, and the results of our second sis are presented in this article.

osely following AMI was the introduction of a 64K ROM from Mostek (MK36000). This was also a e design for ROMs because it employed dynamic peripheral circuits around the cell array. Performwas greatly improved, but the device could not be operated in a mode inherently available in fully devices.

amples were also obtained from Motorola, National and Signetics. The Motorola (MCM68364) design first Motorola ROM to employ the HMOS (high-performance MOS) process. Devices from National 2164) and Signetics (S2664) appear to be based on standard design and process technology.

— results —

the table below shows some of the results obtained through testing. The first AMI design appeared to 'brute force'' approach which was intended to be refined at a later date. Their redesigned ROMs have aved performance over the original samples.

Contraction of the state of the	NAME AND ADDRESS OF TAXABLE AND ADDRESS	NAMES OF TAXABLE PARTY AND ADDRESS OF TAXABLE PARTY AND ADDRESS OF TAXABLE PARTY ADDRESS OF TAXABLE PARTY.			
	Mostek MK36000P-4 (3 samples)	Motorola MCM68364 (4 samples)	AMI Am4264 (3 samples)	Signetics S2664 (3 samples)	National MM52164 (1 sample)
(nS) ¹					
C C	132.0 146.3	197.4 233.1	207.1 223.8	313 350.1	378.5 556.1
standby ¹ N			2		
	3.93 3.30	4.7 4.1	123.1 109.0	114.5 99.1	64.3 56.3
dynamic ¹ ()					-
C C	17.5* 15.9*	47.0 42.4	137.3 123.0	121.3 105.5	79.0 72.5
ec = 5.0 V	* Clock	cycle = 500 nS	3		

continued on next page

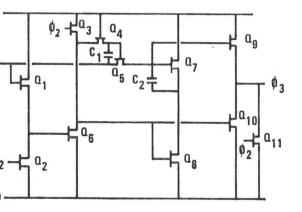
nued from previous page

The Mostek ROM exhibited the best perforce characteristics of all the devices tested. No r source, except perhaps Motorola, could ch Mostek's speed/power characteristics.

In the case of Motorola, HMOS is a new ress and we can expect a certain amount of trial error before it is fully understood. (For a ription of the HMOS process, refer to Coment News 265, page 5.) As mentioned earlier, onal and Signetics chose to rely on their dard fabrication processes, and the device ormance reflects this decision.

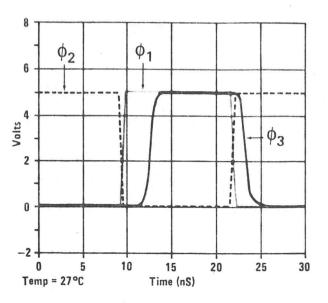
lusions -

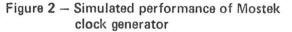
We feel that the Mostek 64K ROM is being as a vehicle to prove out circuit designs for 64K RAM (which is yet to be released). For reason, we decided to more fully analyze the ce by investigating chip circuitry and modeling a of the basic internal functions. Figure 1 vs a schematic diagram of the clock generator Figure 2 shows the performance obtained from puter simulation.



ire 1 - Clock generator; Mostek MK36000-P4

The results indicate a propagation delay of 3 to S. Considering that the MK4116 (16K RAM) k generator has a 5 nS delay, we may expect e form of this circuit to show up in the 4164 (64K RAM).





In a fully static ROM like the AMI 4264, it is possible to keep chip select low and perform the address sequencing to obtain valid data. The only concern is that a certain amount of delay equal to access time be established to ensure valid data is available on the data bus. Because the Mostek ROM is a clocked part, chip enable must be brought back high to ensure valid data for the next cycle. If this is not done the ROM will not operate properly. The trend toward clocked parts is expected to continue as ROMs become larger and more sophisticated.

The parts tested can be grouped into three general categories: The Mostek-type part, characterized by low power consumption and fast access time; the intermediates, illustrated by the AMI 4264 and Signetics 2664; and the slow performer, characterized by National's 52164.

Fortunately, many ROM manufacturers are designing parts with goals of much lower power consumption and less than 100 nS access times. We expect this type of part to dominate the marketplace very soon.

If you have any questions concerning our evaluation, please contact me at 58-299, ext. 6302.

·Bob Goetz

component news 266

Benefits of incoming inspection on CMOS

by Wilton Hart Digital Component Engineering

Editor's note: Wilton presented this paper at the "Testing in Electronics Manufacturing" symposium held in San Jose during April of this year. His findings were also featured in the September/October issue of Evaluation Engineering magazine.

The interest in CMOS testing has grown in the last few years as usage increases. The question that comes from product assembly areas is, "Why don't these parts work in our circuits?" To prevent this question from arising, some form of testing is needed. Testing can take two forms: the first, and probably the easiest to implement, is incoming testing; the second is characterization.

Tektronix set up an incoming inspection procedure on CMOS to sample or 100% test incoming lots. Over the last year 127,000 CMOS parts were tested with a 1.5% reject rate. This means that 1,851 parts did not meet the parameters specified on the data sheet and probably would not have worked in the production line. If the cost of labor to find and replace a bad part exceeded \$4.80 it would be cheaper to test each part first. This conclusion is based on testing cost of seven cents per part. See Figure 1.

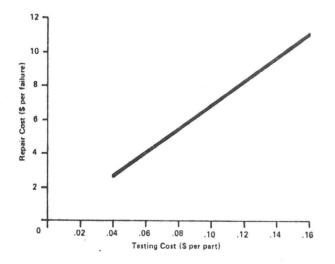


Figure 1 - Repair Cost vs. Testing Cost

What tests need to be performed?

A question which often comes up when discussing incoming inspection is: "What tests need to be performed?" Some think a simple functional test is enough while others also recommend parametric and AC tests. To help determine what tests are needed, CMOS reject information was saved from incoming inspection tests. The reject data was broken out into three categories: functional, input current, and other. Input current failures accounted for 58% of the total rejects. Functional rejects accounted for 27% and other rejects made up the remaining 15%.

page 5

The input current may go out-of-spec for two reasons: one of the protection diodes may become leaky, or the part may have a gate-oxide punch-through. The input current spec is $1\mu A$, but the typical values are several orders of magnitude less.

The input may receive a static discharge which degrades it but the input current may still be only 500 nA. This device will probably fail in the first few months of operation. The ability to measure input current much lower than 1μ A is needed.

To be effective, in catching both upper and lower protection diode failures, a tester should measure the input current near ground and at the supply.

The functional failures were truth table errors, so high speed testing was not needed to catch these rejects. (The 27% did not include AC or propagation delay failures.) A simple test fixture could be made to catch the first two types of failures. It would consist of some standard truth table pattern generator which is connected to the part through a set of large value resistors. (The large value resistors are used to detect input current problems.) The output would then be compared to a known good device. The pattern would be a slow but complete truth table test. If the input is not high impedance, the signal will be loaded and the part will not have the correct output. See Figure 2.

continued on page 6

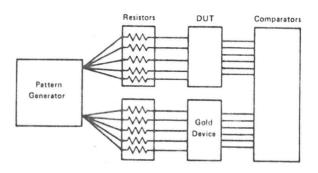


Figure 2 – Functional and Input Current Test Fixture

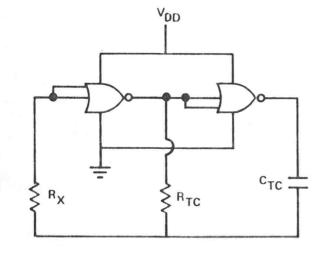
Even though this test fixture requires a known good device, it is simple and much less expensive than one capable of measuring exact values of input current as well as performing a functional test.

The category "other" contained two types of rejects. The largest portion of the parts failed to meet power supply current limits. These rejects were interesting because they probably reveal a contaminated process. The remaining parts were rejected for speed problems.

The test we use for quiescent current and for leakage paths caused by process contamination is slightly different from the way most vendors specify it. At Tektronix the part is powered up and all inputs are forced to zero. A current meter is inserted in the ground lead. The inputs are forced to the supply one at a time and a current measurement is taken. This type of test not only checks the quiescent current but also checks for leakage to ground from any input. The vendors leave the inputs at ground. This quiescent current test requires current measuring capability and programmable power supplies so the simple tester would not be adequate.

Characterization

The second area of interest is characterization. A benefit of characterization is the ability to determine why one part works in a particular circuit and another doesn't. For example, Tektronix used a two gate oscillator circuit made up of "A Series" type parts. See Figure 3. The circuit did not always start up properly and with some parts it oscillated at a different frequency.



To solve this problem we characterized both parts that worked and those that did not. The transfer curve seemed to be the key. Figure 4 shows the transfer curve of the parts which worked.

Figure 5 shows the transfer curve of those that did not work. The DIP or lower gain point caused the circuit not to work correctly.

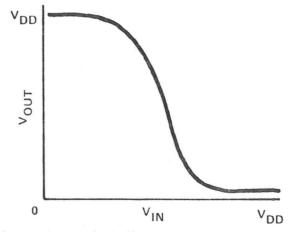


Figure 4 - "A Series" Transfer Curve

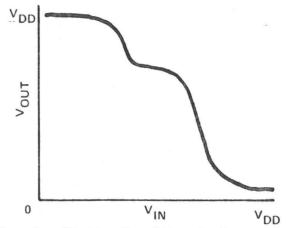
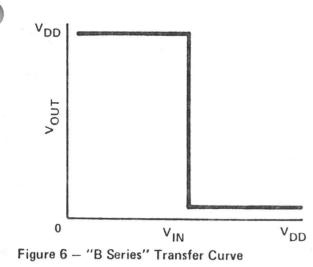


Figure 5 – "Problem Parts" Transfer Curve

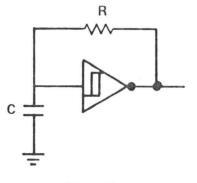
This circuit has caused other problems. "B Series" parts will not reliably work. The transfer curve of "B Series" parts is shown in Figure 6.

At some power supply voltages the circuit will not oscillate. At this point, the first gate is sitting with its input voltage equal to its output voltage because there is negative feedback provided by RTC. This point is either below or above the threshold point of the second gate so the circuit is stable. This could not happen with "A Series" parts because the threshold portion of the curve is much wider.

Figure 3

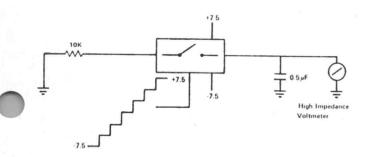


A better circuit is shown in Figure 7. This circuit uses a Schmidt Trigger which has specified threshold points so the percentage of frequency change due to part differences can be calculated.





Characterization can also be used effectively in choosing the right analog switch. The data sheets on these parts don't tell the whole story. For example, the data sheets on the CD4016 and CD4066 are quite similar yet the CD4066 will not work correctly in some sample-and-hold circuits. The test circuit is shown in Figure 8.



Using this circuit, it would seem that the voltmeter would read zero no matter what the control voltage was. This is not true. See Figure 9 and 10.

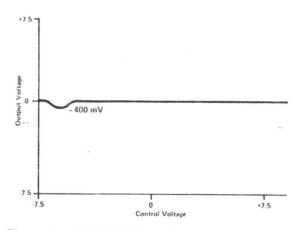


Figure 9 – CD4016; Sample and Hold

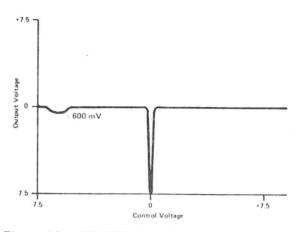


Figure 10 - CD4066; Sample and Hold

The CD4016 has a 400 mV DIP in the output near the -7.5 control voltage point. The CD4066 has the same problem but in addition the output goes to the negative rail at the point where the control equals zero.

In a sample-and-hold application this output glitch discharges the storage capacitor and ruins the stored voltage level. The cause of the glitch is a pair of FETs which were added to help with a latch-up problem. Using a fast risetime control signal reduces this problem.

Characterization can also be used to compare different vendors parts. This type of information can be used for qualifying new vendors or second sourcing.

Figure 8 - Sample and Hold Test Jig

An interesting comparison which I recently completed was a characterization of CD4011B (RCA) and MM74C00 (National). These two parts are quad two-input NAND gates. They have the same function but their pinout is different. Figures 11 and 12 show the comparison.

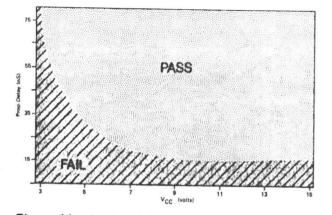


Figure 11- National 74C00; Prop Delay vs. V_{CC}

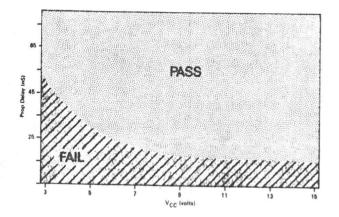


Figure 12 - RCA CD4011; Prop Delay vs. V_{CC}

The two parts were characterized for propagation delay versus power supply voltage. Both parts have a minimum propagation delay of 17 nanoseconds at the higher power supply voltages. The interesting area is below 5 volts. The National part has a 77 nanosecond prop delay at 3 volts where the RCA part has 47 nanosecond propagation delay. The data sheets for these parts do not list specs at 3 volts yet the RCA part is almost 1/3 faster at this power supply level. As illustrated, characterization yields actual boundaries of device operation.

Conclusion

In conclusion; a simple, inexpensive test fixture can catch 85% of the CMOS rejects at Incoming Inspection. The remainder require somewhat more complex equipment. Characterization is very helpful in revealing performance characteristics not evident from data sheets and can explain a device's failure to perform as expected.

New manager in Optoelectronic/ Passive Components group

Paul Curley is the new Optoelectronic and Passive Components manager within Component Engineering. Paul has a BSEE from the University of Cincinnati, and has been working for Tek since 1971. His most recent position was Applications Support Manager at Walker Road.

Paul can be reached at 58-299, ext. 6389.

Would whoever borrowed the black loose-leaf notebook on aluminum electrolytic capacitors from the Component Engineering area please return it immediately.

VONOBLE COPOCITO

Parameter	Q	Price	Size (approx.)
Ceramic (disc – single turn)	Moderate Q most >500 at 1 MHz some>3000 at 100 MHz	Medium 35 - 75¢	Small to medium 0.1×0.15'' to 0.4×0.3''
Ceramic (tubular – multi-turn)	Moderate Ω most>500 at 1 MHz	Low 25 · 45¢	Medium 0.2×0.7‴
Compression Mica	Generally the lowest Q of all variable capacitors	Low to medium 30-60¢	Large typically 0.4x0.8x0.5'
Air (plates - single turn)	Very High Q	Medium 50 - 80¢	Small - medium - large 0.25x0.38'' and up
Piston (air - glass - quartz) e.g., 281-0152-00	Very High Q typically 2000 - 5000 some>10,000 at 100 MHz	High \$1.50 - \$12.00 typically \$3,50	Small - medium - larg 0.075x0.2'' 0.15x0.5'' 0.30x 1.0''
Film	Varies greatly from 200 at 1 MHz to 5000 at 1 MHz	Low 15 · 40¢	Medium to large 0.2x0.4'' to 0.4x0.4x0.5''
Teflon e.g., 281-0064-00	High Q rarely less than 2000 at 1 MHz	Low 25 - 35¢	Medium 0.15×0.7''

component news 266

Selection guide

If you have an application for a variable capacitor and several different types are being considered, this chart might help you make a decision. Much of this information is based on manufacturer's data and on parts purchased in large quantities. For example, the price of an air variable capacitor could be \$15 each (18 - 1000 pF), but most of the air variable caps we use (in quantity) are approximately \$0.50 - 0.80 each. For questions concerning these parts contact Alan LaValle (58-299), ext. 5415.

Temperature Coefficient	Resolution	C-Range $\left(\frac{\text{Size}}{\text{Range}} \right)$	Voltage
Good to poor 50±50 to – 1500±800 PPM/°C	Single turn	High or low depending on K of dielectric	Most from 100 - 350 VDC
Good to medium/poor 0±100 to 100±400 PPM/°C	Multi-turn	Low	400 VDC
Very poor 0±1500 to 0±2500 PPM/°C	Multi-turn; but comparatively poor resolution for a multi-turn part.	Generally the highest	Most at 175 VDC Some at 500 VDC
Very good typically 45±45 PPM/°C some 0±15 PPM/°C	Single turn (multi-turn air variable caps have a gear reduction)	Very low (k≈1)	150 V and up
Excellent 0±15 PPM/°C	Multi-turn very good resolution	Low	500 VDC to 1250 VDC
Good to poor 0±150 to 100±500 PPM/°C	Single turn High C films have poor resolution	Very High	Most from 100 - 300 WVDC
Good 50±50 PPM/ ^o C	Multi-turn, grasshopper style good resolution	Very low	600 VDC

Invalid low from Signetics hex buffer

We have received many inquiries lately about Signetics' 8T97 hex buffer. It is possible to get an invalid low when the input stays high and the enable pin is toggled.

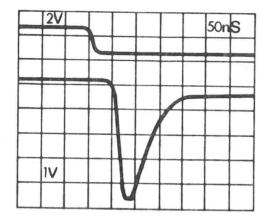


Figure 1 - Signetics hex buffer

In Figure 1, the top trace is the enable pin and the bottom trace is the output. When this test was performed there was a $10K\Omega$ resistor used to pull the output high while it was in the tri-state condition. The magnitude of the pulse is frequencydependent. Figure 1 shows the maximum voltage swing for this particular IC. The testing was conducted at 10 KHz. However, there was a 4.2 volt dip at 800 KHz which decreased abruptly as the frequency increased.

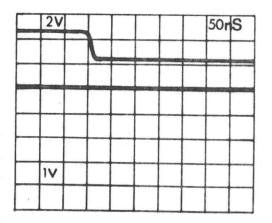


Figure 2 - Motorola hex buffer

The Motorola IC depicted in Figure 2 does not exhibit these characteristics. Therefore, we strongly recommend that the Motorola parts be used in all new designs. Also, be aware that many of the other devices in Signetics' 8T series exhibit the same characteristics.

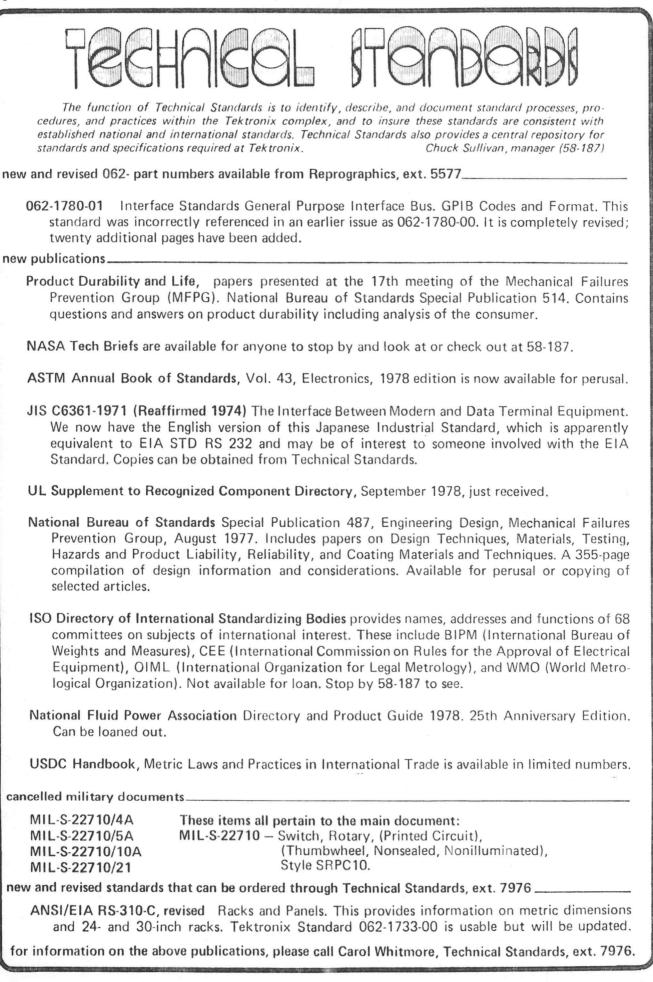
A more detailed evaluation of this device is underway. For details, contact Don VanBeek (58-125), ext. 5414.

EUROPEAN PRODUCT SAFETY POST FILLED

Bob Randall has been named the Tektronix European Product Safety Manager. He will be coordinating our efforts in meeting the various regulations in Europe, the Mid-East and Africa.

Bob has a great deal of product safety experience. He has represented Tek on numerous national and international product safety committees, and has a high degree of success with other projects at Tek. He will be moving to Amsterdam by mid-January.

If you have a question relating to international product safety, you can either work through Product Safety Engineering (58-123), or contact Bob directly.



page 12

PART-NUMBERED TEKTRONIX STANDARDS

Whether your job requires all issues of part-numbered Tektronix Standards, or only selected copies, please check the following list to ensure you have the latest available. Call Reprographics, ext. 5577, to obtain new or revised standards, unless otherwise noted.

Tek P/N	Standard	Latest Issue
062-1699-00	Introduction to the Directory	26Jun78
	(contained in Directory, obtain from Technical Standards)	
062-1700-00	Subject Index, Directory	12Aug77
062-1701-00	Trademarks, Copyrights, and Related Proprietary Matters	12Jun78
062-1702-00	Fabrication Standards, Welding, Soldering, and Brazing	15Mar77
062-1703-00	Finish Standards, Glossary of Terms	2Sep77
062-1704-00	Drafting Standards, Decimal Inch Dimensioning	22Jun78
062-1705-00	Metric Standard, Drawings (revision pending)	5Aug77
062-1706-00	Converting Fractions to Decimals	18Nov74
062-1707-00	Rounding off Decimal Numbers	18Nov74
062-1708-00	Drafting Standards, Drawing Format	1Jun78
062-1710-00	Test Procedures for Panels and Tags	8Nov78
062-1714-00	Tooling Standard, Wiedeman Dies and Punches	25Jul78
062-1716-00	Drafting Standard, Symbols for Tolerances of Position and Form	29Nov77
062-1718-00	Finish Standard, Cosmetic	8Nov77
062-1720-00	Finish Standard, Mold Surface Finishes for Plastics	12Jan77
062-1721-00	Circuit Board Standards, Drafting (absorbs and replaces G-102, now invalid)	10Nov78
062-1723-00	Circuit Board Standard, Manufacturing, One and Two Layers	2Jun78
062-1725-00	Circuit Board Standard, "EC 200B"	4Feb77
062-1727-00	Circuit Board Standard, Multi-Layer, Manufacturing	
062-1721-00	Hardware Standard, Weld Studs and Posts	12Sep78
062-1731-00	Component Mounting Standards	12Jan77
062-1732-00		10Jul78
	Component Mounting Standards, A Series Details	14Apr78
062-1732-02	Component Mounting Standards, B Series Details	14Apr78
062-1732-03	Component Mounting Standards, C Series Details	14Apr78
062-1732-04	Component Mounting Standards, D Series Details	14Apr78
062-1732-05	Component Mounting Standards, E Series Details	120ct78
062-1732-06	Component Mounting Standards, F Series Details	8Sep77
062-1732-07	Component Mounting Standards, H Series Details	120ct76
062-1732-08	Component Mounting Standards, M Series Details	120ct76
062-1733-00	Rackmount Standards	1Feb77
062-1736-00	Communications Standard, Glossary of Technical Terms	7Jan77
062-1737-00	Communications Standard, Abbreviations, Acronyms, and Symbols	20Jul78
062-1738-00	Available Part-Numbered Standards (Directory, Technical Standards)	26Jun78
062-1778-00	Circuit Board Standards, UL Recognition and Flammability Classification	24Jan77
062-1780-00	Interface Language Standard, High Level Format (replaced by 062-1780-01)	
062-1780-01	Interface Standard, General Purpose Interface Bus (GPIB) Codes and Formats	6Jul78
062-1860-00	Product Safety Standard for X-Radiation	19May75
062-1874-00	Drafting Standards, Line Conventions and Lettering	15May78
062-1875-00	Drafting Standards, Projections, Views and Sections/Details	13Jun77
062-1879-00	Cable Standard, Product Wiring, Internal	17Mar78
062-1880-00	Cable Standard, External Interconnecting and Power Cables, Drafting (revision pending)	4Apr77
062-2318-00	Circuit Board Standard, Marking Adherence and Solder Flux Cleaning	1Apr76
062-2319-00	Switch Standard, Cylindrical Cam Switch	30Jun76
062-2463-00	Index, Letter Series Standards, (Directory, Technical Standards)	26Jun78
062-2476-00	Drafting Standard, Symbols, Schematic Diagrams, Electronic Circuits	31Jul78
002 2470 00	prarting standard, symbols, schematic Didyrams, Electronic Circuits	3130178

- - - /- -

continued from preceding page

Tek P/N	Standard	Latest Issue
062-2801-00	Occupational Safety Standards, Distribution Procedures (from Corporate Safety and Health)	25Jun76
062-2801-01	Occupational Safety Standards, Warning Alarms	
062-2801-02	Occupational Safety Standards, Exhaust Hood Classification	28Jun77
062-2801-03	Occupational Safety Standards, Exhaust Hood Classification	20Sep76
062-2801-04	Occupational Safety Standards, Handling Flammable Liquids Under Pressure	22Feb77
062-2801-05	Occupational Safety Standards, Forming and Shaping of Asbestos Materials	16Sep76
062-2843-00	Occupational Safety Standards, Open Tank Classification Drafting Standard, Drawing Scale	8Aug77
062-2846-00	Drafting Standard, Drawing Scale	50ct76
062-2847-00	Drafting Standard, Glossary of Terms, Dimensioning and Tolerancing	140ct76
062-2848-00	Product Design Standard, Environmental Test, Atmospheric	15Jun77
062-2851-00	Fabricating Standard, Bend Allowance and Deduction	10Nov77
062-2853-00	Cable Standard, Glossary of Terms	2Nov76
062-2858-00	Product Design Standard, Environmental Test, Product Classification	13Jun77
002-2000-00	Floudet Design Standard, Environmental Test, Dynamics:	17Jun77
062-2862-00	Vibration, Shock and Transit	
062-2866-00	Product Design Standard, Environmental Test, Electrostatic Discharge	200ct77
062-2877-00	Froduct Design Standard, Environmental Test Electromagnetic Compatibility	31Mar77
062-3083-00	Cable Standard, wire and Cable Color Coding System	210ct76
062-3099-00	Documentation Standard, Assignment of Item Names	8Mar77
062-3108-00	Drafting Standards, Engineering Change Order (ECO)	28Apr78
062-3109-00	Circuit Board Standards, Electrodeposited Gold Plate, Contact Areas	200ct77
062-3134-00	Documentation Standards, Lechnical Standards, Procedures and Format	4Jan78
062-3159-00	Color Standard, Color Description, Selection, and Testing	lov77
062-3500-00	Drafting Standard, Marking, Heat and Roll Stamps, Die Cast and Molded	'n77
062-3539-00	rest method Standard, Lime Delay of 50μ RF Coaxial Cable and Cab'	77
062-3546-00	1051 Wellood Standard Cables Spark Test	7
062-3716-00	Drafting Standard, Draft Considerations, Molded Plastic Part	81, ⁸⁰ D
062-3744-00	Component I.D. Marking Standard, Transistors and Dior	a ⁰⁸ J/8 JMar78
062-3748-00	Occupational Safety Standard, Illumination, Minim	
062-3752-00		2Jun78
062-3797-00	Communication Standard, Product Marking	6Sep78
062-3901-00	Software Standard, BASIC Language Sta	7Aug78
002-3901-00	Drafting Standard, Temporary I.D. of Englished Standard, Temporary I.D	16Aug78
062-3923-00	Prior to Part Number Assignment 8 th Test Method Standard, Cables, Jacket Remov 69 ^V	8Nov78

To receive a copy of any of these part-numbered technical standards, contact Reprographics, ext. 5577. Direct any other questions concerning Tek standards to Technical Standards, ext. 7976.

Archiving ROMs and PROMs

In an effort to eliminate errors in archiving, we are requesting that you follow a standard format for all data tapes sent to Documentation Coordination (58-299).

By convention, a hole in the tape equals a ''1'' state (i.e., $V_{OH}).$ This is irrespective of whether the component programs to a high or low state.

The format is based on the majority of tapes received. *Tapes which do not adhere to the format will be returned to the initiator.*

For more information on the archiving system see Component News 256, page 6.

omponentNewsNewComponents

This column is designed to provide timely information regarding new components, vendors, availability and price. "New Components" can also be used as an informal update to the Common Design Parts Catalogs. Samples may or may not be available in Engineering Stock.

Harris HA4 T.I. TL33	59 42 44 45 105 195 26	analog d Pre-amp, low noise, wide bandwidth Pre-amp, Bi-FET using N-channel FETs, 1.8nV// Op-amp, Norton, dual, high speed Op-amp, low offset voltage BiFET, low suppl Op-amp, wider G.B.W. than AD542 Op-amp, precision, low drift, FET input (2 pA bit Op-amp, wide bandwidth (18 MHz), BiFET, low off Op-amp, wide band	now Hz noise 1st qtr.'79 now y current, low 1st qtr.'79 now as)	no P/N w bias current	\$ 2.15 0.50 2.50 3.00 5.95	John Hereford, 67(John Hereford, 67(John Hereford, 67(John Hereford, 67(John Hereford, 67(John Hereford, 67(
MatsushitaNationalLM3Analog DevicesAD5Analog DevicesAD5Analog DevicesAD5HarrisHA5HarrisHA5T.I.TL32EXARXR-0HarrisHA43T.I.TL33T.I.TL34	59 42 44 45 105 195 26	wide bandwidth Pre-amp, Bi-FET using N-channel FETs, 1.8nV// Op-amp, Norton, dual, high speed Op-amp, low offset voltage BiFET, low suppl Op-amp, wider G.B.W. than AD542 Op-amp, precision, low drift, FET input (2 pA bit Op-amp, wide bandwidth (18 MHz), BiFET, low off Op-amp, wide band	Hz noise 1st qtr.'79 now y current, low 1st qtr.'79 now as)	no P/N no P/N no P/N w bias current no P/N	0.50 2.50 3.00	John Hereford, 67(John Hereford, 67(John Hereford, 67(John Hereford, 67(
NationalLM3Analog DevicesAD5Analog DevicesAD5Analog DevicesAD5Analog DevicesAD5HarrisHA5HarrisHA5T.I.TL32EXARXR-0HarrisHA43T.I.TL33T.I.TL33T.I.TL34	59 42 44 45 105 195 26	Pre-amp, Bi-FET using N-channel FETs, 1.8nV// Op-amp, Norton, dual, high speed Op-amp, low offset voltage BiFET, low suppl Op-amp, wider G.B.W. than AD542 Op-amp, precision, low drift, FET input (2 pA bis Op-amp, wide bandwidth (18 MHz), BiFET, low off Op-amp, wide band	Hz noise 1st qtr.'79 now y current, low 1st qtr.'79 now as)	no P/N no P/N w bias current no P/N	0.50 2.50 3.00	John Hereford, 670 John Hereford, 670 John Hereford, 670
Analog DevicesAD5Analog DevicesAD5Analog DevicesAD5HarrisHA5HarrisHA5T.I.TL32EXARXR-0HarrisHA42T.I.TL33T.I.TL34	42 44 45 105 195 26	Op-amp, Norton, dual, high speed Op-amp, low offset voltage BiFET, low supply Op-amp, wider G.B.W. than AD542 Op-amp, precision, low drift, FET input (2 pA bis Op-amp, wide bandwidth (18 MHz), BiFET, low off Op-amp, wide band	1st qtr.'79 now y current, low 1st qtr.'79 now as)	no P/N w bias current no P/N	2.50 3.00	John Hereford, 670 John Hereford, 670
Analog DevicesAD5Analog DevicesAD5HarrisHA5HarrisHA5T.I.TL32EXARXR-0HarrisHA42T.I.TL33T.I.TL33	44 45 105 195 26	Op-amp, low offset voltage BiFET, low suppl Op-amp, wider G.B.W. than AD542 Op-amp, precision, low drift, FET input (2 pA bis Op-amp, wide bandwidth (18 MHz), BiFET, low of Op-amp, wide band	y current, lov 1st qtr.'79 now as)	w bias current no P/N	3.00	John Hereford, 67(
Analog Devices AD5 Harris HA5 Harris HA5 T.I. TL32 EXAR XR-0 Harris HA4 T.I. TL33 T.I. TL33	45 105 195 26	Op-amp, wider G.B.W. than AD542 Op-amp, precision, low drift, FET input (2 pA bis Op-amp, wide bandwidth (18 MHz), BiFET, low of Op-amp, wide band	1st qtr.'79 now as)	no P/N		
Harris HA5 Harris HA5 T.I. TL32 EXAR XR-0 Harris HA43 T.I. TL33 T.I. TL33	105 195 26	Op-amp, precision, low drift, FET input (2 pA bi Op-amp, wide bandwidth (18 MHz), BiFET, low of Op-amp, wide band	as)	no P/N	5.95	John Hereford, 670
Harris HA5 T.I. TL32 EXAR XR-C Harris HA42 T.I. TL33 T.I. TL49	195 26	Op-amp, wide bandwidth (18 MHz), BiFET, Iow of Op-amp, wide band				,
T.I. TL32 EXAR XR-0 Harris HA43 T.I. TL33 T.I. TL48	26	Op-amp, wide band	fset	no P/N	2.60	John Hereford, 670
EXAR XR-0 Harris HA43 T.1. TL33 T.1. TL48		(150 MHz G.B.W.), fast se	now	no P/N S)	6.50	John Hereford, 670
Harris HA43 T.I. TL33 T.I. TL48		Op-amp, HEX 741 characteristics	2nd qtr.'79			John Hereford, 670
T.I. TL33 T.I. TL49	94/095	Op-amp, quad, power programmable BiFET		no P/N		John Hereford, 670
T.I. TL48	295	Comparator, quad, 40 nS response, 2.0 mV offset	now	no P/N	4.75	John Hereford, 670
	36	Comparator, HEX, LM339 characteristics		no P/N	0.85	John Hereford, 670
T.I. TL48	90/TL491	Comparator, 10-step analog level detector, thresholds		no P/N	1.00/ 1.10	John Hereford, 670
	37	from 200 MV to 50 mV Comparator, 5-step level detector, 3 dB thresholds		no P/N	0.60	John Hereford, 670
		electromechan	ical devices			
		Wire, stranded 10 AWG, Black, UL style 1015	now	175-5090-00		Rod Christiansen, 595
	1	Wire, stranded 10 AWG, White, UL style 1015	now	175-5091-00		Rod Christiansen, 595
Zepher/3M	(Cable assembly, 34 conductor, 28 AWG, 2	1/79 '' long 3M#	175-2456-00	5.00	Rod Christiansen, 595
Zepher/3M	. (Cable assembly, 25 conductor, 26 AWG, 6	now	012-0882-00	11.00	Rod Christiansen, 595
Zepher/3M	(Cable assembly, 25 conductor, 26 AWG, 6	now	012-0883-00	12.50	Rod Christiansen, 595
Zepher/3M		Cable assembly, 50 conductor, 28 AWG w, Connectors: 131-1781-00,	now /grnd. plane,	012-0853-00 14' long.		Rod Christiansen, 595
		memory and I	/O devices			
Intel 2732		EPROM, 4K x 8	now	no P/N	91.65	Bob Goetz, 630

page 16

component news 266

	optoelectro	onic devices			
VPR	Capacitor, 5600µF,	now	290-0853-00	and all so the	Don Anderson, 5415
	6.3 WVDC, single ended	l, high ripple	e current		
VPR	Capacitor, 1200 μ F,	now	290-0877-00		Don Anderson, 5415
	6.3 WVDC, single ended	d, high ripple	e current		
35ELA3300	Capacitor, 3500µF,	now	290-0873-00		Don Anderson, 5415
	35 WVDC, axial lead				
PFP	Capacitor, 1100 μ F,	now	290-0878-00		Don Anderson, 5415
	200 WVDC, printed circ	cuit mount			
	Capacitor, 15µF,	now	290-0876-00		Don Anderson, 5415
	25 WVDC, molded tant	alum capaci	tor, P.C. mount		
PME 265	Capacitor, 2200 pF,	1/79	285-1192-00	-	Don Anderson, 5415
	250 VAC, plastic, UL, I	EC-65 appr	oved for across-the	line use	
	Capacitor, .012 μ F,	now	285-1191-00		Don Anderson, 5415
	1000 WVDC, plastic, hi	gh current			Sec.
MV57124	LED, rectangular, red		150-1070-00	0.50	Betty Anderson, 6389
	VPR 35ELA3300 PFP PME 265 	$\begin{array}{llllllllllllllllllllllllllllllllllll$	$\begin{array}{cccc} 6.3 \ \text{WVDC}, \ \text{single ended}, \ \text{high ripple}\\ \hline & & & & & & & & & & & & & & & & & & $	VPR Capacitor, 5600μ F, now 290-0853-00 6.3 WVDC, single ended, high ripple current VPR Capacitor, 1200 μ F, now 290-0877-00 6.3 WVDC, single ended, high ripple current 35ELA3300 Capacitor, 3500 μ F, now 290-0873-00 35ELA3300 Capacitor, 3500 μ F, now 290-0873-00 35 WVDC, axial lead PFP Capacitor, 1100 μ F, now 290-0878-00 200 WVDC, printed circuit mount Capacitor, 15 μ F, now 290-0876-00 25 WVDC, molded tantalum capacitor, P.C. mount PME 265 Capacitor, 2200 pF, 1/79 285-1192-00 250 VAC, plastic, UL, IEC-65 approved for across-the Capacitor, 012 μ F, now 285-1191-00 1000 WVDC, plastic, high current	VPR Capacitor, 5600μ F, now 290-0853-00 6.3 WVDC, single ended, high ripple current VPR Capacitor, 1200 μ F, now 290-0877-00 6.3 WVDC, single ended, high ripple current 35ELA3300 Capacitor, 3500 μ F, now 290-0873-00 35 ELA3300 Capacitor, 3500 μ F, now 290-0873-00 35 WVDC, axial lead PFP Capacitor, 1100 μ F, now 290-0878-00 200 WVDC, printed circuit mount Capacitor, 15 μ F, now 290-0876-00 25 WVDC, molded tantalum capacitor, P.C. mount PME 265 Capacitor, 2200 pF, 1/79 285-1192-00 250 VAC, plastic, UL, IEC-65 approved for across-the-line use 260 WVDC, plastic, UL, IEC-65 approved for across-the-line use Capacitor, .012 μ F, now 285-1191-00

component news_

Published by Technical Communications 58-299 ext. 6867

> Jacquie Calame, editor Birdie Dalrymple, illustrator Lola Janes, writer

To submit an article, call Jacquie on ext. 6867 or stop by 58-299.

For mailing list changes, contact Kelly Turner (19-123), ext. 5502.

company confidential