[54]	PUSH-PULL AMPLIFIER CIRCUIT WITH		
	FIELD-EFFECT TRANSISTORS		

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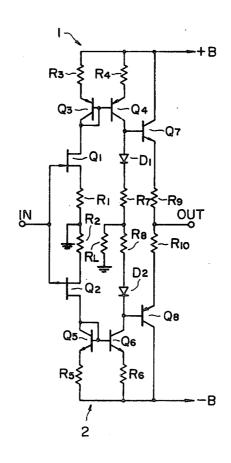
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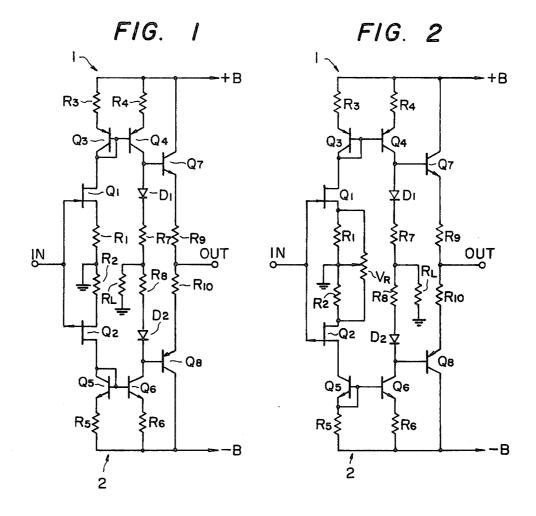
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[57] ABSTRACT

Disclosed herein is a push-pull amplifier circuit employing field-effect transistors in which the field-effect transistors opposite in electrical conductivity to each other are activated by a common input signal to perform a push-pull operation. Secondary distortion components due to the square-law characteristics of the field-effect transistors are cancelled out by impedance elements each being provided between the source of each field-effect transistor and a common reference potential point.

5 Claims, 2 Drawing Figures





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PUSH-PULL AMPLIFIER CIRCUIT WITH FIELD-EFFECT TRANSISTORS

BACKGROUND OF THE INVENTION

The present invention relates to amplifier circuits, and more particularly to a push-pull amplifier circuit using field-effect transistors.

In an amplifier circuit employing field-effect transistors as its amplifying elements, the input-output characteristics of the field-effect transistors are of the squarelaw characteristics, and therefore the amplified output unavoidably includes non-linear distortion components. The distortion components may be eliminated by employing a negative feedback technique. However, the 15 and Q2 are represented by VGS1 and VGS2, and currents employment of such a negative feedback technique causes not only to decrease the amplification degree but also to harm the circuit stability, which may result in the oscillation of the circuit.

SUMMARY OF THE INVENTION

Accordingly, an object of the invention is to provide an amplifier circuit in which the secondary distortion components due to the square-law characteristics of field-effect transistors have been eliminated by a simple 25 The currents I1 and I2 are outputted after being subcircuit arrangement, without utilizing a negative feedback technique.

The specific feature of the amplifier circuit according to the invention resides in that field-effect transistors opposite in electrical conductivity to each other are 30 activated by a common input signal to perform a pushpull operation, so that the secondary distortion components produced by the transistors are cancelled out by each other. In the invention, the secondary distortion components are more effectively eliminated by adjust- 35 rent mirror circuits 1 and 2 are equal to each other. ing impedance elements adapted to set circuit currents, so that the currents in the transistors are equal to each other.

BRIEF DESCRIPTION OF THE DRAWINGS

FIGS. 1 and 2 are circuit diagrams showing first and second embodiments of a push-pull amplifier circuit using field-effect transistors according to the invention.

DESCRIPTION OF THE PREFERRED **EMBODIMENTS**

The invention will be described with reference to the accompanying drawings.

FIG. 1 is a circuit diagram showing a first embodiment of a push-pull amplifier according to the inven- 50 tion. In FIG. 1, reference characters Q1 and Q2 designate an N-channel field-effect transistor and a P-channel field-effect transistor which are subjected to push-pull drive by a common input signal, respectively. A source sistor Q₁ and a grounding point at a reference potential. Similarly, a source resistor R₂ is connected between the source of the transistor Q2 and the grounding point. Current mirror circuits 1 and 2 are provided as the drain loads of the transistors Q1 and Q2, respectively. The 60 current mirror circuit 1 is constituted by a diode-connected input transistor Q₃, an output transistor Q₄ and emitter resistors R₃ and R₄. The current transfer ratio is determined to a desired value by suitably setting the Similarly, the current mirror circuit 2 is constituted by a diode-connected input transistor Q5, an output transistor Q₆, and emitter resistors R₅ and R₆. The current

transfer ratio is determined from the ratio of resistances of the resistors R5 and R6.

The output currents of the output transistors Q4 and Q₆ in the two current mirror circuits 1 and 2 are applied to a common load resistor R_L respectively through a series circuit of a diode D1 and a resistor R7 and a series circuit of a diode D₂ and a resistor R₈. Voltage drops across these series circuits are employed as base bias sources for output power transistors Q7 and Q8 in a SEPP (single ended push-pull) arrangement. The emitter follower outputs of the output transistors Q7 and Q8 are supplied to a load such as a loudspeaker (not shown) through resistors Q9 and Q10, respectively.

If the gate-source voltages of the input transistors Q₁ in the transistors Q_1 and Q_2 are represented by I_1 and I_2 , respectively, then the following equations are established:

$$I_1 = (V_i - V_{GS1})/R_1 \tag{1}$$

$$I_2 = (V_i - V_{GS2})/R_2$$
 (2)

where V_i is the input signal voltage.

jected to current conversion in the predetermined ratios by the current mirror circuits 1 and 2, respectively. Therefore, if the current transfer ratios are one (1), then the currents represented by equations (1) and (2) are outputted, as they are. The currents are supplied to the common load resistors R_L respectively through the bias circuits comprising the diodes D₁ and D₂. If the resistors R₁ and R₂ are equal in resistance to each other, then the DC components of the output currents of the cur-Therefore, in this case, the DC components are not allowed to flow to the resistor R_L ; that is, only the AC components are allowed to flow to the resistor R_L .

The amplifying transistors Q_1 and Q_2 implement a 40 push-pull operation by receiving the common input signal, and therefore the AC current components are opposite in phase to each other. Accordingly, the level of an AC current carried by the resistor R_L is twice that of the output AC current of each of the current mirror 45 circuits 1 and 2. As is apparent from equations (1) and (2), the output AC currents include secondary distortion components attributing to the square-law characteristic of a field-effect transistor. Under the conditions that R₁=R₂ and the current transfer ratios of the current mirror circuits 1 and 2 are equal to each other, the secondary distortion components are equal in level to each other and are in phase with each other. Therefore, currents due to the secondary distortion components are cancelled out and a distortionless signal is supplied resistor R₁ is connected between the source of the tran- 55 to the common load resistor R_L. In the above-described example, the emitter follower output transistors Q7 and Q₈ are driven in a push-pull mode by the collector outputs of the transistors Q4 and Q6 to apply power to a common load such as a loudspeaker (not shown). Therefore, according to a principle similar to the abovedescribed one, a distortionless signal is applied to the common load such as a loudspeaker.

FIG. 2 is a circuit diagram showing a second embodiment of the push-pull amplifier according to the invenratio of resistances of the emitter resistors R₃ and R₄. 65 tion. In FIG. 2, like parts or like components are designated nated by like reference characters or like reference numerals as used in FIG. 1. The second embodiment in FIG. 2 is different from the first embodiment in FIG. 1 only in that a variable resistor VR is connected between the sources of the transistors Q_1 and Q_2 and its movable terminal (or the wiper terminal) is grounded, so that equivalently the resistors R_1 and R_2 are connected in parallel to the halves of the variable resistor VR.

In the first embodiment shown in FIG. 1, the secondary distortion due to the push-pull operation may not be completely eliminated unless the source resistors R_1 and R_2 are accurately equal to each other. However, the difficulty can be eliminated by the provision of the variable resistor VR thus connected, because the variation in resistance of the source resistors R_1 and R_2 can be corrected by adjusting the variable resistor VR. In the second embodiment, it should be appreciated that the resistance of the variable resistor is much larger than that of each of the resistors R_1 and R_2 (VR>>R1 or R_2).

As is apparent from the above description, the pushpull circuit according to the invention is advantageous in that the secondary distortion due to the square-law characteristic of a field-effect transistor can be simply eliminated without utilizing a negative feedback technique.

In the above-described embodiments, the currents of the amplifying transistors Q₁ and Q₂ are applied through the current mirror circuits to the load. However, it is obvious that circuits equivalent in function to the current mirror circuits can be employed. Furthermore, if instead of the source resistors, impedance elements dependent on frequency are employed, then the push-pull amplifier circuit can be modified into an amplifier circuit whose gain characteristic is dependent on frequency. Therefore, for instance a distortionless equalizer amplifier can be obtained.

What is claimed is:

1. A push-pull amplifier circuit comprising:

first and second field-effect transistors opposite in electrical conductivity to each other, a common input signal being applied to the gates of said first and second field-effect transistors;

first and second impedance elements connected respectively between the source of said first fieldeffect transistor and a common reference potential point and between the source of said second fieldeffect transistor and said common reference potential point; and

current generating means, comprising first and second current mirror circuits receiving inputs from said first and second field-effect transistors, respectively, for outputting currents which have predetermined ratios with respect to currents in said first and second field-effect transistors, respectively, said currents thus outputted being applied to a common load.

The push-pull amplifier circuit as claimed in claim
 1, wherein said first and second impedance elements are provided with means for adjusting the impedances of said first and second impedance elements to be equal to each other.

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In the above-described embodiments, the currents of e amplifying transistors Q₁ and Q₂ are applied through e current mirror circuits to the load. However, it is ovious that circuits equivalent in function to the cur-

4. The push-pull amplifier circuit as claimed in claim 30 2 wherein said first and second impedance elements are resistors and said adjusting means is a variable resistor and wherein the resistance of said variable resistor is much larger than the resistance of each of said resistors.

5. The push-pull amplifier circuit as claimed in claim35 1 or 2 wherein said first and second impedance elements are of frequency dependence.

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