INSTRUCTION MANUAL

IMPORTANT

Before operating this instrument, be sure to remove the plastic shipping clamps from the shock mounts of the amplifier chassis. These clamps should be saved and reinstalled as shown in the sketch if the instrument is to be shipped. Be sure the tongue is inserted next to the chassis to prevent damage to the shock mount.

The clamp should be on the same side of the shock mount as the nut.





Tektronix, Inc.

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WARRANTY

All Tektronix instruments are warranted against defective materials and workmanship for one year. Tektronix transformers, manufactured in our own plant, are warranted for the life of the instrument.

Any questions with respect to the warranty mentioned above should be taken up with your Tektronix Field Engineer.

Tektronix repair and replacement-part service is geared directly to the field, therefore all requests for repairs and replacement parts should be directed to the Tektronix Field Office or Representative in your area. This procedure will assure you the fastest possible service. Please include the instrument Type and Serial number with all requests for parts or service.

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SECTION 1

CHARACTERISTICS

General Information

The Tektronix Type 519 Oscilloscope is a wide-band laboratory instrument designed expressly for the observation and measurement of high-frequency phenomena. Fast linear sweeps, high CRT accelerating potential, excellent triggering sensitivity, wide-band trigger system, and vertical bandwidth well beyond 1000 megocycles permit accurate repetitive and single-shot displays to be observed and photographed from fractional-nanosecond signals. An internal delay line in the vertical channel of the instrument permits display of the leading edge of the signal triggering the oscilloscope. Sweep delay control through 35 nanoseconds permits viewing signals before and after the main signal event.

The Type 519 incorporates two internal waveform generators. An adjustable repetition-rate fast-rise pulse generator and a fast-rise calibration-step generator supply waveforms which can be used to chack the calibration of the oscilloscope itself, or to drive external devices. These waveform generators meet most requirements to complete a test setup.

VERTICAL-DEFLECTION SYSTEM

Vertical Deflection Factor

With Type T519P-A CRT, less than 10 volts per cm. Exact CRT deflection factor indicated on CRT face mask of each instrument.

Passband

With Type T519P-A CRT, dc to 1000 megacycles minimum at 3-db down.

Risetime

With Type T519P-A CRT, less than 0.35 nanosecond *

Input Impedance

125 ohms.

*1 nanosecond=10-9 seconds.

۲

Maximum Allowable Input Power to Vertical Channel

1.8 watts, corresponding to ±15 volts dc or rms.

Maximum Allowable Peak Signal Amplitude

±100 volts. Repeated pulses of higher voltage may damage the 125-ohm signal termination resistor.

Internal Signal Delay

Approximately 45 nanoseconds, fixed.

Voltage Standing Wave Ratio

Nominally 1.25:1 to 1000 mc.

TRIGGER

Triggering Signal Sources

Internal from \pm applied signals, internal from the Rate Generator, internal from the Calibration-Step Generator, and external from \pm trigger inputs.

External Triggering Signal Requirements

Pulse amplitude: 20 millivolts. Duration: 1 nanosecond or longer. Maximum permissible external triggering signal: ± 10 volts peak, higher with external attenuators. Repetition rate: to 1000 mc.

Internal Triggering Signal Requirements

Pulse amplitude: sufficient signal to produce a 2 tracewidth deflection on the screen (approximately 200 mv). Duration: 1 nanosecond or longer. Repetition rate: to 1000 mc.

Countdown

A sweep is obtained for each trigger signal at trigger signal frequencies below the maximum sweep repetition rate. Triggering circuits countdown for trigger signal frequencies

Characteristics—Type 519

higher than the maximum sweep repetition rate. Sine-wave trigger requirements: 1 mc to 1000 mc, 20 mv peak-to-peak external trigger input, or 200 mv peak-to-peak signal to vertical input.

Pulse Amplitude or Sync

Single control adjusts pulse triggering level or high-frequency sync. An additional control provides vernier sync.

Delay

Sweep-start delay over a range of 35 nanoseconds. Permits waveform to be positioned horizontally within the sweep, to display a selected time interval.

TIME BASE

Sweep Rates

Nine Ranges: 2, 5, 10, 20, 50, 100, 200, 500, and 1000 nanoseconds per centimeter.

Accuracy

Typically within 2% of indicated rate on all ranges except the 2-nanosecond range, which is within 3%. These specifications apply to the entire sweep except for the first 2 nanoseconds or 2 mm (whichever is larger).

Single Sweeps

Normal or single sweeps selected by front-panel switch.

+ Trigger Output

Greater than 1-volt pulse into 50 ohms upon triggering.

Delayed + Gate

Greater than 1-volt gate into 50 ohms during sweep; delayed with respect to + Trigger Output depending on the setting of the sweep-delay control.

RATE GENERATOR

Risetime

Less than 0.8 nanosecond (0.5 nanosecond typical).

Pulse Repetition Rate

3 cps to 30 kc, continuously variable.

Pulse Duration

10 nanoseconds + or - 20% at the 50% amplitude points.

Output Impedance

50 ohms.

Amplitude

Approximately +15 volts.

CALIBRATION-STEP GENERATOR

Risetime

Approximately 0.1 nanosecond.

Repetition Rate

Adjustable from 400 to 850 step waveforms per second; normally operated near reed-switch resonant frequency at approximately 750 steps per second.

Output Impedance

125 ohms.

Amplitude

Into 125 ohms, 0 to 10 volts. Into 50 ohms through T50/T125 adapter, 0 to 1 volt. Voltages applied to both impedances are continuously variable and calibrated. Uncalibrated voltages up to 50 volts into 125 ohms.

Polarity

The ouput polarity can be selected by a front-panel switch.

CATHODE RAY TUBE

Type

T519P-A

Phosphors

Type P11 phosphor standard (recommended for single-shot photographic recording at fastest sweep rate). Other phosphors available only on special request.

Usable Viewing Area

Two centimeters vertical, six centimeters horizontal.

Accelerating Voltage

24 kv.

Spot Diameter

0.004 inch (approximately 0.1 mm) at normal intensity.

Deflection

Electrostatic. Vertical deflection system is 125-ohm distributed-constant delay line. Conventional horizontal deflection plates.

CONSTRUCTION AND POWER REQUIREMENTS

Construction

Single-unit construction with light-weight aluminum-alloy chassis and four-piece vinyl-finish cabinet. Side panels, top and bottom panels are separately removable.

Ventilation

Filtered forced air with protective thermal cutout insures safe operating temperatures.

Dimensions

Approximately 221/4" high, 143/4" wide, and 251/4" long.

Power Requirements

105 to 125 or 210 to 250 volts, 50 to 60 cycles, approximately 650 watts.

ACCESSORIES

Information on accessories for use with this instrument is included at the rear of the mechanical parts list.



Fig. 2-1. Type 519 Oscilloscope front panel.



SECTION 2

OPERATING INFORMATION

MULTIPLIER Multiplier for the CYCLES/SEC control. Should be set to OFF position when Rate Generator is not being used.

CALIBRATION-STEP GENERATOR

- RANGE Selects full-scale amplitude of the calibrated steps, a variable uncalibrated step amplitude, or a standby condition. Should be set to STANDBY when Calibration-Step Generator is not being used.
- VARIABLE Permits the step waveform to be preset to an arbitrary desired amplitude up to about 50 volts into a 125-ohm load.
- VOLTS Sets the output voltage of the calibrationstep waveform in the 10 V and 1 V positions of the RANGE switch.
- REED SWITCH:
 - DRIVE Adjusts the reed-switch magnetic excitation for proper closures or permits single reed operation.
 - FREQUENCY Controls frequency of the reed-switch magnetic excitation to help minimize contact bounce.

TRIGGER

PULSE AMPLI-Selects triggering signal amplitude re-TUDE OR SYNC * auired to operate triggering circuits, or adjusts synchronization. VERNIER SYNC Used in conjunction with PULSE AMPLI-TUDE OR SYNC control to synchronize the sweep. FUNCTION * Permits choice of triggered or synchronized displays. GAIN * Selects proper gain or attenuation for the triggering signal. TRIGGER Selects the triggering signal source and SOURCE polarity.

* Also see TRIGGER OPERATION section of front-panel, Fig. 2-1.

Introduction

The Type 519 Oscillascope is a high speed laboratory instrument designed for observing, measuring, and photographically recording phenomena in the nanosecond (milimicrosecond) domain. However, before the instrument can be used successfully, it is important for you to have an undestanding of the operation of each control. This section of the Instruction Manual is intended to help you acquire this understanding. Much of the familiarity with the controls will come only with actual use of the instrument. A brief description of each of the front-panel controls follows. Frontpanel markings are shown in Fig. 2-1.

FUNCTION OF CONTROLS

CRT

- FOCUS Used in conjuction with the ASTIGMATISM control to focus the oscilloscope trace. INTENSITY Adjusts the brightness of the trace. ASTIGMATISM Used in conjunction with the FOCUS control to obtain a round spot and a sharply focused trace. SCALE ILLUM. Adjusts the brightness of the araticule
- markings.
- Graticule A knurled knob located below the center Control of the graticule permits graticule to be moved down out of the viewing area.

TIME BASE

- NANOSEC/CM Selects the desired time base.
- DELAY Determines the delay of the start of the sweep with respect to the trigger signal input.
- NORMAL-SINGLE Selects either normal or single-sweep SWEEP operation.

RATE GENERATOR

CYCLES/SEC Used in conjunction with the MULTIPLIER control to set the Rate Generator output frequency.

Operating Information—Type 519

POSITIONING

VERTICAL Adjusts the vertical position of the trace. AXIS ROTA-TION Aligns the trace parallel to the horizontal graticule lines.

HORIZONTAL Adjusts the horizontal position of the trace.

POWER

- DIM ADJ. Used to adjust the brightness of pilot light after 45-second warm-up period.
- ON AC line switch.

PRELIMINARY INSTRUCTIONS

Cooling

A blower maintains safe operating temperature in the Type 519 Oscilloscope by drawing air through a filter and circulating it over the components. Therefore, the instrument must be placed so that the air intake and cabinet ventilating holes are not blocked. The air filter must be kept clean to permit adequate air flow.

A thermal cutout switch disconnects the dc power if the instrument becomes overheated. The pilot lamp will return to full brightness in the event that dc power is lost. If this happens, check immediately for proper airflow into the instrument. The blower continues to cool the interior and reduces the time the thermal switch remains open. DC power will be restored when the temperature drops to a safe value.

Power Requirements

The regulated power supplies in the Type 519 Oscilloscope will operate with line voltages from 105 to 125 volts (117 nominal) or from 210 to 250 volts (234 nominal). The line voltage for which your instrument is wired at the factory is indicated on a metal tag fastened to the rear panel near the power receptacle. Transformer connections may be changed for either 117- or 234-volt operation by using the information given in Fig. 2-2. The power transformer is wound with two 117-volt primaries which are connected in parallel for 117-volt operation and in series for 234-volt operation. Since the blower motor is connected across only one of the transformer primaries, no change in the motor lead connections is required. When the transformer connections are changed, the voltage indicated on the metal tag should be covered with another tag which conforms to the new operating voltage.

For maximum dependability and long life, the line voltage applied to the Type 519 Oscilloscope should be near the voltage indicated on the metal tag located near the power receptacle at the rear of the instrument. If the line voltage exceeds the operating limits, or has a poor waveform (distorted sine waves), unstable power-supply operation may



Fig. 2-2. Power transformer connections for operation of the Type 519 Oscilloscope at 117 or 234 volts.

result. Check for proper line voltages and waveform before checking for other causes of unstable operation.

Fuse Requirements

When the Type 519 Oscilloscope is connected for 117volt operation, use a 7-amp slow-blowing type fuse. When the instrument is connected for 234-volt operation, use a 4-amp slow-blowing type fuse.

Time Delay

A time delay relay used in the Type 519 delays operation of the instrument for approximately 45 seconds after the instrument is switched on. The relay allows a brief tubewarmup period before the dc-perating voltages are applied. When the ac-power pilot light dims, the instrument is ready for use.

If the ac-power is interrupted for only an instant, the normal 45-second delay will occur before the instrument returns to full operation.

Dim Adjustment

The DIM ADJ. control is a screwdriver adjustment which controls the brightness of the ac-power pilot light after the 45-second warm-up period. Normally, it is adjusted to a setting which will reduce glare from the pilot light when wareform observations are being made in a darkened room.

Camera Bezel

When one of the Tektronix cameras is used with the Type 519 the bezel supplied on the oscilloscope must be used. The bezel supplied with the camera will not take the graticule assembly properly.

FIRST-TIME OPERATION

To place the Type 519 in operation for the first time, the following procedure is suggested:

 Set the front-panel controls as follows (controls not mentioned may be placed in any position);

POWER		Off
DIM ADJ.		Centered
INTENSITY	Fully	counterclockwise
NORMAL-SINGLE SWEEP		NORMAL
NANOSEC/CM		5
DELAY		Centered
MULTIPLIER		X1000
CYCLES/SEC		10
RANGE		STANDBY
TRIGGER SOURCE		RATE GEN.
GAIN		NORMÁL
FUNCTION		PULSE
PULSE AMPLITUDE OR SYNC	Fully	counterclockwise
VERTICAL		Centered
HORIZONTAL		Fully clockwise

 (The line voltage for which the instrument is wired at the factory is indicated near the power cord receptacle.)
 Connect the power cord to the rear of the instrument and to the source of power.

3. Set the POWER switch to ON.

 Allow about 45 seconds for the pilot lamp to dim, indicating that dc operating voltages are applied and the instrument is ready for use.

CAUTION

Do not turn the intensity so high that a bright glow surrounds the spot. Excessive brightness of a stationary spot may damage the screen in a few seconds.

5. Advance the INTENSITY control until a visible spot appears near the left center of the screen.

6. Adjust the FOCUS and ASTIGMATISM controls to produce a small round spot.

 Advance the PULSE AMPLITUDE OR SYNC control fully clockwise to obtain a horizontal sweep across the screen. Readjust the INTENSITY control for suitable trace brightness.

8. Rotate the HORIZONTAL positioning control to position the start of the trace at the left marking of the graticule.

9. Adjust the AXIS ROTATION control until the trace is parallel to the horizontal markings of the graticule.

10. Connect a T50/N125 adapter to the +RATE 50 Ω connector making certain that the 50-ohm connectors are mated. If the wrong impedance connectors are mated, the signal path remains open and the connections will not seat fully.

11. Connect a 2-nsec 125-ohm cable from the 125-ohm end of the adapter to the SIGNAL 125 Ω connector.

 Rotate the PULSE AMPLITUDE OR SYNC control slowly counterclockwise and adjust the DELAY control until a stable display of the Rate Generator pulse is obtained. Locate the waveform vertically using the VERTICAL positioning control.

13. Adjust the FOCUS, INTENSITY, and ASTIGMATISM control until a sharp trace with adequate intensity is obtained. These controls until a sharp trace with adequate intensity is obtained. These controls are slightly interdependent. An external signal and/or trigger may now be applied to the input connectors. If external triggers are used, set the TRIGGER SOURCE switch to $+ \sigma - EXT$. If internal triggering from the input signal is used, set the TRIGGER SOURCE switch to $+ \sigma - INT$.

CRT CONTROLS AND GRATICULE

Intensity

The INTENSITY control is used to adjust the brightness of the oscillococycle display. Compensation can be made for changes in brightness resulting from changes in the triggering rate or time base. The INTENSITY control is rotated clockwise to increase brightness and counterclockwise to decrease brightness. Care must be taken when using the INTENSITY control that the brightness is not turned up to the point where the phosphor on the face of the cathode ray tube (CRT) becomes permanently damaged. The intensity of the beam should never be turned up to the point where a bright halo forms around a stationary spot.

The FOCUS and ASTIGMATISM controls permit a sharp, clearly defined spot or trace to be obtained. Perhaps the best way to adjust the FOCUS and ASTIGMATISM controls is to display a waveform on the oscilloscope and then adjust the FOCUS and ASTIGMATISM controls alternately for the best averall focus of the trace. It may be necessary to make a new adjustment of the controls if the intensity of the trace is changed.

The disappearing graticule used with the Type 519 Oscillacope is accurately marked with 6 harizontal and 2 vertical 1-centimeter divisions. The minor division markings on the horizontal centerline are 5 millimeters apart, these on the vertical centerline are 2 millimeters apart. The graticule markings allow accurate time and voltage measurements to be made from the oscillascope screen.

To move the graticule out of the viewing area of the screen, loosen the knurled knob located just below the graticule and slide it downward the full length of the slot. Tighten the knob. To return the graticule to functioning position, reverse the process.

The proticule cover and mask assembly is held securely in place by four slotted graticule nuts and is provided with hinge fittings for mounting the viewing hood. In addition, the hinge fittings allow quick removal of the viewing hood so that a Tektronix Model C-12 or C-19 camera may be mounted. The Model C-19 camera is especially designed to photograph the fast sweeps of the Type 519 Oscilloscope. When the camera is not being used, it can be unlatched and swung away from the CRT screen.

Graticule Illumination

The graticule is illuminated by two lamps located at the top edge of the graticule. The SCALE ILLUM. control, located below the oscilloscope screen, is rotated clockwise to brighten the graticule markings and counterclockwise to dim them.

Camera Jack

A camera jack, marked 6.3V CAMERA, provides a 6.3-volt source for use with a camera. When the camera plug is inserted in the jack, the SCALE ILLUM. control and oscilloscope graticule lights are automatically disconnected.

POSITIONING

Two controls, VERTICAL and HORIZONTAL, are used to position the trace to the desired point on the oscilloscope screen. A third positioning control, AXIS ROTATION, is used to align the trace with the horizontal markings of the graticule.

The VERTICAL position control has sufficient range to allow the trace to be positioned completely off the top or bottom of the screen, or to any intermediate point. The trace moves up when the control is rotated clockwise and down when the control is rotated counterclockwise.

The HORIZONTAL position control causes the trace to move to the right when it is rotated in the clockwise direction and to the left when it is rotated counterclockwise. The total horizontal positioning range of the control is about 2 centimeters.

The AXIS ROTATION control is a screwdriver adjustment located between the VERTICAL and HORIZONTAL controls. This adjustment permits the trace to be rotated about an axis through the center of the screen.

VERTICAL-DEFLECTION SYSTEM

Signal Input Connection

The electrical signal to be observed is applied externally through a 125-ohm coaxial cable to the SIGNAL 125 Ω connector. If the impedance of the signal source is other than 125 ohms, corresponding cables and a suitable adaptor should be used to prevent mismatches and resulting reflections. The signal passes internally first through a trigger takeoff, then through a 45-nsec delay cable to the distributed vertical deflection system of the CRT. The signal causes the spot to be deflected vertically. The spot traces out the signal waveform on the screen as the spot is deflected horizontally by the horizontal sweep circuits. The vertical size of the displayed waveform can be adjusted to a suitable amplitude by inserting external attenuators or an amplifier in series with the signal-carrying cable. Or, if the Calibration-Step Generator is being used as the signal source, the vertical amplitude of the waveform can be adjusted by means of the Calibration-Step Generator front-panel controls

The vertical sensitivity of the Type 519 Oscilloscope is dependent on the CRT mounted in the instrument and on the adjustment of the high voltage. The risetime and sensitivity of each Type 519 CRT is measured at the factory. These measurements are then recorded on the CRT face mask. The sensitivity measurement can be checked at any time by using the Calibration-Step Generator.

To check the measurement, connect a 125-bhm coble from the OUTPUT 125 Ω connector to the SIGNAL 125 Ω connector. Set the (CALIBRATION-STEP GENERATOR) RANGE switch to 10V TO 125 Ω and rotate the VOLTS control to 10.00. Adjust the oscilloscope front-panel controls for a stoble presentation of that sep waveform. Adjust the VOLTS control until the portion of the waveform located 2 nasc after the rise is exactly one centimeter high. The vertical sensitivity in volts per continuer controls read directly from the VOLTS dial.

For example, if the VOLTS dial shows a reading of 8.70, the vertical deflection factor is 8.7 volts per centimeter.

When connecting the oscilloscope to any signal source, the connections should be made directly through 125-ohm cables or through suitable impedance matching devices to the SIGNAL 125 Ω connector. However, when impedance matching devices are used, you must consider possible signal voltage changes produced by the devices. If the signal amplitude is too great, it will be necessary to attenuate the signal to a usable level before applying it to the SIGNAL 125 Ω connector. This can be done by inserting a 125 Ω attenuator (of known attenuation factor) between the signal source and the SIGNAL 125 Connector. Attenuators may be used individually or may be "stacked" (connected in series).

If the signal amplitude is too low to produce sufficient vertical deflection, an external amplifier can be inserted between the signal source and the SIGNAL 125 Ω connector. However, if the amplifier does not provide the correct input and output impedance, severe waveform distortion may result. In addition, if the amplifier stages have limited bandwidth, or do not operate linearly, the signal will not be reproduced faithfully on the CRT.

In general, to obtain an accurate waveform display and to prevent unwanted reflection of high-frequency waveforms or of fast-rise pulses, all cables should be terminated in their characteristic impedances. An exception is described under Accessories, part (4) Adaptor NSO(N125 (page 2-11).

Delaying the Signal

The Type 519 Oscilloscope contains a fixed 45-nanosecond signal-deloy line which allows sufficient time for the trigger circuits to process the trigger signal and start the sweep before the leading edge of the input signal arrives at the CRT. The internal delay dca weep start and before display of the triggering signal. At the slower sweep rotes the triggering signal vill appear very near the start of the trace. If you wish the signal to appear farther to the right, you may insert additional 125-ohne delay cable must be added after the trigger takeoff point to increase the signal delay. The most common point of insertion is at the CRT end of the fixed delay line.

The DELAY control provides a 35-nanosecond adjustment in sweep storting time with respect to the triggering signal. Within this range of adjustment, the DELAY control can be used to select the display time and thus apparently position the waveform horizontally on the screen with respect to the trace.

For triggered sweep operation with externally-derived trigger signals, the time relationship of the external trigger signal to the input signal must fall within the adjustment range of the DELAY control. If, for example, too much delay is introduced by using long cables to couple the trigger signal to the EXTERNAL TRIGGER US 0 connector, the input signal will arrive at the vertical deflection plates before the sweep is triggered. The signal input waveform, having arrived early, will not be displayed on the screen. To affste external delay of this type, shorten the external trigger cable, if possible. If this is not possible, cable can be added into the signal-carrying circuits, but only with loss of bandwidth due to high-frequency attenuation in the cable.

The delay provided by a typical 125-ohm cable such as the RG-G3/U cables shipped with the oscilloscope is approximately 1.2 nsec per foot. If any portion of the input waveform is displayed on the screen, the amount of delay which



Fig. 2-3. Typical waveforms resulting from incorrect signal or triggering signal delays. Waveform (a) results from either too much triggering delay or too little signal delay. Waveform (b) results either from too little triggering delay or too much signal delay.

must be added or subtracted to display the waveform properly on the screen can be determined by the sweep rate and the number of divisions that the display must be moved. If the display must be moved to the right, less delay in the trigger cable is required. Fig. 2.3 hows displays resulting from incorrect signal or trigger delays.

When 30-megacycle or higher repetition-rate signals of identical shape and amplitude are being displayed, the DE-LAY control will always permit display of the complete waveform. If all waveforms are uniform, it is not important which one is displayed. The internal delay line may be bypassed by direct connections to the CRT if desired, at the sarrifice of internal triagening from the signal.

Triggering (or Synchronizing) the Sweep

In most applications it is desirable for a repetitive waveform to appeer stationary on the oscilloscope screen so that the characteristics of the waveform can be examined in detail. As a necessary condition for this type of display, the start of each horizontal sweep must be time-related to a characteristic of the input waveform. In the Types 197 Oscilloscope this is accomplished either by triggering or synchronizing the sweep with the displayed waveform or with another waveform. More information about the horizontal sweep is given in the Time Base portion of this section of the manual.

The following paragraphs outline the operation of the various triggering controls in the TRIGGER section of the front panel, in the order normally encountered. Usually, the TRIGGER SOURCE switch would be set first, GAIN switch second, and FUNCTION switch third. Finally, the PUISE AMPLITUDE OR SYNC control is adjusted to obtain a stable display for "triggered" sweep operation. For

Operating Information—Type 519

"synchronized" sweep operation, both the PULSE AMPLI-TUDE OR SYNC and VERNIER SYNC controls may be adjusted to obtain a stable display.

Selecting the Trigger Source

The sweep can be either triggered or synchronized (depending upon the setting of the FUNCION switch) from the following sources: (1) displayed waveform, (2) externally-derived waveform, (3) Calibration-Step Generator, or (4) Rate Generator. The trigger source selection is by means of the TRIGGER SOURCE switch. Each trigger source has advantages for certain applications.

(1) Displayed Waveform. Triggering from the displayed waveform is the method most commonly used. Triggering is from the displayed waveform with the TRIGGER SOURCE switch set to either the +INT. or -INT. position. Internal friggering is convenient, since no external friggering signals or connections are required. A displayed waveform that produces at least two trace-widths of vertical deflection with a time duration of one nanosecond or more is sufficient for reliable triggering.

[2] Externally-Derived Signal. To trigger the sweep from an external signal, connect the triggering signal to the EX-TERNAL TRIGGER 125 Ω input connector. The external triggering signal must be at least 20 millivols in amplitude, with a time duration of 1 nanosecond or more. The maximum amplitude should not exceed ± 2 volts peak exceed the triggering signals more than the GAIN switch is set to X.2. In the X.2 position the external triggering signals more than the TRIGGER SOURCE switch is set to either $\pm EXT$. External triggering signals preferably should be disconceted from the EXTERNAL TRIGGER 125 Ω connector when some other mode of triggering is used in traduction the external triggering signals preferably should be disconceted from the EXTERNAL TRIGGER 125 Ω connector when some other mode of triggering is used to reduce the possibility of strue triggering.

External triggering provides definite advantages over other methods of triggering in certain applications. With external triggering, the triggering signal usually remains constant in amplitude and shape (depending upon the source). Also, time and phase relationships between waveforms at different points in a circuit can be seen. If, for example, the external triggering signal is derived from the waveform at the input to a device under test, it is possible to observe the shapring, jitter, amplification, or delay of the signal through the device without resetting the oscilloscope triggering controls for each observation.

(3) Calibration-Step Generator. In the +CAL or -CAL positions of the TRIGGER SUCKE switch, the triggering signal is advised through the use of a trigger takeoff circuit inserted near the termination of the generator. The signal is derived through the use of a trigger takeoff digut the output amplitude of the Calibration-Step Generator also affect the amplitude of the triggering signal available at the +CAL and -CAL positions of the TRIGGER SUDKE switch. These two positions of the TRIGGER SUDKE switch. These two positions of the TRIGGER SUDKE switch are used to beserve a waveform which is time-related to the output waveform of the Calibration-Step Generator. It is then position to delay at various points in the device under test. In

addition, the internal-trigger requirement for a minimum signal height and duration can be circumvented. When the step generator is not being used, it should be set on STAND-BY.

(4) Rate Generator. The RATE GEN, position of the TRIG-GER SOURCE switch provides triggering signals which can be varied to cover a continuous repetition-rate range from 3 cps to 30 kc. These signals can then be used to trigger the sweep at a known repetition rate within the above range. To select the Rate Generator trigger, set the RIGGER SOURCE switch to RATE GEN. Then set the MULTIVBRATOR switch and CYCLES/SEC control to the desired repetition rate.

Selecting the Trigger Polarity

The horizontal sweep can be triggered on either the rising (+slope) or falling (-slope) portion of the triggering waveform as determined by the position of the TRIGGER SOURCE switch.

In many applications the triggering polarity is important since triggering on the wrong slope will make it impossible to display the portion of the waveform which is of interest. In many other cases, however, such as high-frequency repetitive waveforms, the triggering signal polarity is usually not important.

Selecting the Trigger Gain

A four-patihon GAIN switch permits incoming trigger signols to be attenuated or amplified as necessary for proper triggering or synchronization. The four gain settings are: X2, NORMAL X5, and X20. To aid in determining which GAIN switch setting to use for reliable triggering, Table 2-1 is included. Additional information is given in the PULSE AMPLITUBE OR SYNC control description.

TRIGGER SOURCE	Approximate GAIN Switch Settings			
Switch Setting	X.2	NORMAL	X5†	X20†
RATE GEN.		Always set in this position.		
±CAL.*		8 v to 50 v	1.5 v to 50 v	0.5 v to 50 v
±EXT.	1 v to 10 v (peak)	0.2 v to 2 v (peak)	0.04 v to 2 v	0.01 v to 2 v
±INT. **	10 v (pulse) to 100 v (pulse)	2 v to 20 v (pulse)	0.4 v to 20 v	0.1 v to 20 v

TABLE 2-1

⁺Used for small amplitude triggers up to 200 mc.

* Calibration-Step Generator output step amplitudes are listed. Approximately 2.5% of the step amplitude is coupled to the + CAL. and - CAL, positions of the TRIGGER SOURCE switch.

** Voltage ranges of signals applied to the SIGNAL 125 Ω connector are given. Approximately 10% of the signal amplitude is picked off and coupled to the + INT. and - INT. positions of the TRIGGER SOURCE switch.

Selecting the Trigger Function

Three functions or modes of operation are provided in the Type 519 Oscilloscope to cover a wide range of triggering conditions: They are: PULSE, SYNC, and HF SYNC.

To determine the best trigger mode for a particular application, it is best to have some understanding of all three before making a selection.

Each of the triggering modes is designed to provide stable triggering from a certain type of waveform. For many applications, however, more than one mode will work well. For such applications, the triggering mode selected is simply a matter of choice.

The PULSE mode permits choice of a free-running sweep or a sweep triggered by signals at random or uniform repetition rates up to 50 mc. The upper repetition-rate limit varias, depending upon the regularity of the pulse period. The PULSE mode, when used in conjunction with the SINGLE SWEEP feature, permits photographs to be made of single events at any setting of the NANOSEC/CM switch.

The SYNC mode permits stable displays of waveforms occurring at a constant repetition rate-up to approximately 150 mc. To select this mode, place the FUNCTION switch in the SYNC position.

The HF SYNC mode permits the sweep to be synchronized from high-frequency signals in the range from opproximately 100 mc to more than 2 kmc. To use the high-frequency synchronization mode, place the FUNCTION switch in the HF SYNC position.

Triggering or Synchronizing the Sweep

The last controls normally operated in the TRIGGER section of the front panel are the PULSE AMPLITUDE OR SYNC and VERNIER SYNC controls. These controls are used for two functions: pulse amplitude selection and synchronization, depending upon the setting of the FUNCTION switch.

If the FUNCTION switch is set to the PULSE position, the PULSE AMPLITUDE OR SYNC control determines the level a signal must reach to initiate the sweep. All triggers below the set level are rejected. In order for the control to operate properly within its rotational range, sufficiently large triggers must be available as explained earlier under Selecting the Trigger Gain.

Triggering on small signals is best when the control is set just short of the point where the sweep free runs. The sweep normally free runs when the PULSE AMPLITUDE OR SYNC control is rotated clockwise past the RECURRENT arrow.



Fig. 2-4. The usual oscilloscope display is a graphical presentation of voltage versus time.

If the FUNCTION switch is set to either the SYNC or the HF SYNC position, the PULSE AMPLITUDE OR SYNC control is used for making the coarse synchronization adjustment. Final adjustment may be made with the VERNIER SYNC control. The sweep repetition rate will synchronize at the frequency of the triggering signal or at some submultiple frequency.

The VERNIER SYNC control is normally set at midrange until the coarse adjustment is made, then the VERNIER SYNC control is adjusted to obtain a stable display.

TIME BASE

Horizontal Sweep

The Type 519 Oscilloscope graphically presents instantaneous signal valtage verus time (see Fig. 2.4). The signal valtage produces vertical deflection of the trace; time is represented through horizontal deflection. The horizontal sweep is also known as the time base, since horizontal deflection of the spot bears a definite relationship to time and provides the means for making time measurements from the screen. The NANOSEC/CM switch selects the desired sweep rate from one of nine accurately calibrated trates available. Time base steps range from 2 nanoseconds per centimeter to 1000 nanoseconds [] usec) per centimeter. The sweep generator has been designed to provide long-term stability of sweep calibration and linearity.

Single Sweep Operation

The Type 519 Oscilloscope permits a single-sweep presentation to be obtained and eliminates all subsequent sweeps to the signal can be clearly recorded without contupion resulting from multiple traces. The single-sweep feature is selected by placing the NORMAL-SINGLE SWEEP switch in the SINGLE SWEEP position. The REST button must be actuated to "arm" the time base and permit a single-triager event.

When the FUNCTION switch is placed in the PULSE position and the PULSE AMPLITUDE OR SYNC control is set fully clockwise (past the RECURRENT arrow or line), a single sweep runs immediately each time the RESET button is depressed.

When the PULSE AMPLITUDE OR SYNC control is set for triggered sweep operation, the single sweep does not necessarily occur immediately after the RESET button is depressed. Instead, the READY lamp lights to indicate that the sweep is armed and ready to be triggered. When a trigger is received, the sweep runs ance and the READY light goes out. Each time the RESET button is depressed the procedure is repeated.

When the FUNCTION switch is placed either in the SYNC or HF SYNC position, a single sweep runs immediately each time the RESET button is depressed regardless of the settings of the PULSE AMPLITUDE OR SYNC control.

The time base may also be reset externally, using the 3-conductor plug supplied in the accessory kit and a normally open push-button switch. Connect the two center terminals of the plug through a cable to the two terminals on the pushbutton switch and insert the plug into the EXTERNAL RESET jack on the rear panel. With the NORMAL-SINGLE SWEEP switch in the SINGLE SWEEP position, the push-button switch may be used to arm the sweep.



Fig. 2-5. Using the Type 519 Oscilloscope to drive an external circuit. The output of the external circuit is then applied to the input of the oscilloscope for display.

Synchroscope Operation

In the usual oscilloscope application, the sweep is triggered or synchronized by the input waveform. However, in some applications it may be more desirable to reverse the process and drive an external circuit from the oscilloscope. In this "synchroscope" application, the sweep is caused to free run or to be triggered by the Calibration-Step Generator or the Rate Generator. The output signal from the +TRIGCER 50.0, the DELAYED +CATE, the + RATE 50.0, or the CUTPUT 125 Ω connector is used to initiate the input waveform tese Fig. 2-5).

The sweep can be made to free run in any position of the FUNCTION switch. If the PULSE position of the FUNCTION witch is used, the PULSE AMPLITUDE OR SYNC control must be rotated fully clockwise patt the RECURENT arrow. The sweep free runs at all times when the FUNCTION switch is neither the SYNC or HF SYNC positions. The number of free-running sweeps per second is determined by the settings of the NANOSE/C/dw switch (nefer to Table 22).

A free-running sweep also provides a convenient reference trace on the oscilloscope screen without requiring an input signal. The trace can then be positioned to a desired point on the oscilloscope screen or can be used to establish a zero-voltger efference line.

Delayed Trigger

A delayed triggering pulse is produced at the DELAYED +GATE 50 Ω connector of the oscilloscope at approximately

T A	DI	с.	2 2
1.4	DI	.с	2-2

NANOSEC/CM Switch Settings	SWEEP REPETITION RATES (Recurrent Rates)
2	Adjusted to 400 kc
5	200 kc nominal
10	100 kc nominal
20	50 kc nominal
50	20 kc nominal
100	10 kc nominal
200	5 kc nominal
500	2 kc nominal
1000	1 kc nominal

the time of sweep start. The delay of the delayed triggering pulse with respect to the time that the trigger is accepted and a pulse is produced at the +TRIGGER 50 Ω connector can be adjusted over a range of approximately 35 nanoseconds by means of the DELAY control.

RATE GENERATOR

The output pulse from a transistor operating in the avalanche mode is coupled to the RATE GEN, position of the TRIGGER SOURCE switch and to the +RATE 50 Ω connector. The pulse risetime is less than 0.8 nanosecond, amplitude is nominally +15 volts, and duration is approximately 10 nanoseconds. A typical Rate Generator waveform as displayed on the Type S1P is shown in Fig. 2-6. To use the



Fig. 2-6. Typical Rate Generator output waveform as displayed on the Type 519 Oscilloscope.

Rate Generator, set the TRIGGER SOURCE switch to RATE GEN. Then adjust the CYCLES/SEC and MULTIPLIER controls for the desired repetition rate. Any frequency between 3 cps and 30 kc can be selected within an accuracy of 10%.

Since the sweep can be triggered at the repetition rate set by the Rate Generator from 3 cps to 30 kc, this feature can be used for applications such as those described previously under the headings "Selecting the Trigger Source", and "Synchroscope Operation". When the Rate Generator is not used, if should be turned off by placing the MULTPLIER switch to the OFF position to reduce the possibility of stray triggering.

CALIBRATION-STEP GENERATOR

Step Waveform

The step waveform from the Calibration-Step Generator is generated by discharging a charged coaxial line into



Fig. 2-7. Locations of the Calibration-Step Generator line charging network and trigger takeoff.

an external load through a magnetically-operated dryreed switch. The physical length of the charged line determines the duration of the output step waveform. In the Type 519, with no external charge line added, the duration of the constant-amplitude portion of the step is equal to twice the transit time of the built-in 1.5-nsec charae line. (Transit time of the charge line is the time required for a signal to pass from one end of the line to the other.) For the 1.5-nsec charge line, then, the duration of the output pulse is 3 nsec. To obtain a longer duration step waveform. an additional length of charge line (cable) may be added to the Charge Line Connector located next to the Trigger Takeoff (see Fig. 2-7). When an additional charge line is added, the charging network and the charging voltage must be disconnected. They must then be connected to the open end of the added charge line. A typical display of the Calibration-Step Generator waveform as seen on the Type 519 Oscilloscope appears in Fig. 2-8.



Fig. 2-8. Typical Calibration-Step Generator waveform as displayed on the Type 519 Oscilloscope. It is not always possible to completely eliminate reed switch multiple contact bounce. Extraneous traces, therefore, sometimes will occur in the display.

Polarity

Step polarity is selected by the POLARITY switch. The polarity of the step at the output connector is the same as the polarity of the charge voltage. The setting of the TRIG-GER SOURCE switch should agree with the setting of the POLARITY switch for normal triggering.

Amplitude

The step amplitude is dependent upon the amount of charging voltage used. The charge voltage obtained from the charging source at the instant of read switch closure is 2 times the step voltage present at the OUTPUT 125 Ω connector when driving an external 125-bm load. The step voltage reading is accurately indicated by the settings of the RANGE and VOLTS controls.

When the VQLTS control is set to 10.00, the RANGE switch permits a choice of two full scale step anglitudes, 10 volto cr 1 volt. When the RANGE switch is placed in the 10V TO 125 oposition, 10 volto is produced across a 125-ohm load. When the RANGE switch is set in the 1V TO 50 position, a ToyIT25 acdopter must be properly connected to the OUTPUT 125 Ω connector to obtain a 1-volt step into a 50 ohm load.

The scale of the VOLTS control, when used with either of the two above RANGE switch positions, indicates the step amplitude. The VOLTS control is the 0 to 1 multiplier for the two ranges.

When the RANGE switch is set to the VARIABLE position, the step amplitude may be preset by the VARIABLE control to any uncalibrated amplitude from 0 to approximately 50 volts when driving a 125-ohm load. To determine the amplitude of the step for any setting of the VARIABLE control, apply the step waveform from the OUTPUT 125 Ω connector through a 125-ohm cable (and attenuator, if needed) to the SIGNAL 125 Ω connector. Measure the amplitude of the vertical deflection in centimeters and multiply the distance measured by the sensitivity of the oscilloscope (and attenuation if used).

Adjusting the Drive and Frequency

Two front-panel controls, DRIVE and FREQUENCY, control the movement of the dry-reed switch. These controls are adjusted to cause the reed to make-and-break contact with a minimum of contact bounce.

To adjust the two controls, they must first be preset fully counterclockwise. Then advance the DRIVE control until the reed vibrates (makes a buzzing sound). Advance the FRE-QUENCY control until the reed fails to operate and then rotate the control slightly counterclockwise to start the reed operating again. Slowly rotate the DRIVE control counterclockwise while rotating the FREQUENCY control back and forth to find the resonant frequency of the reed. The resonant frequency is found when the drive is decreased to a point where the reed will vibrate in only one small rotational area of the FREQUENCY control range. For optimum operation the DRIVE and FREQUENCY controls are then adjusted to obtain the most stable waveform near the resonant frequency of the reed. When adjusting the DRIVE control, use enough drive to get solid closures of the reed contacts. The resonant frequency of most reeds is usually within the range of 700 to 800 cps.

NOTE

The reed switches used in the Type 519 are chosen to produce the best possible waveform. The high requirements of these switches frequently result in a short lifetime. To extend the life of the reed switch set the RANGE switch to STANDBY when the Calibration-Step Generator is not being used

ACCESSORIES

The following information pertains to the accessories which are included with the Type 519 Oscilloscope. Other optional accessories which are available are accompanied by specific application notes. See also Section 7, Accessories

(1) 125 O Termination

The 125 Ω Termination (Fig. 2-9) is supplied as a spare for the T519P-A CRT termination or to terminate any 125ohm cable

[&]quot;The letter "T" in an adaptor type means "Terminated"; "N" means "Not Terminated".





Fig. 2-9. Construction of the 125Ω Termination.



Fig. 2-10. Construction of the T50/T125 Adaptor. Also shown are two typical applications for this adaptor.

(2) Adaptor T50/T125*

This adaptor (Fig. 2-10) is commonly known as a minimum loss matching pad. Designed to match between a 50-ohm line and 125-ohm line, the attenuator presents minimum loss and reflections. It contains a network composed of a shunt and a series resistor. Though the attenuator presents a correct impedance match "in either direction", the signal voltage transmission factor of 0.225 in going from 125 ohms to 50 ohms is less than the N50/N125 adaptor described later. In going from 50 ohms to 125 ohms, the signal voltage transmission factor is approximately 0.564. The primary advantage of the T50/T125 adaptor is that it receives signals into either end without producing reflections.

(3) Adaptor T50/N125

This adaptor is usually called a 50-ohm termination adaptor and the internal circuitry is shown in Fig. 2-11. In actual use the 83.3-ohm resistor is shunted by the 125-ohm input impedance of the load. The combined resistances present a total input impedance of 50-ohms to the signal source.



Fig. 2-11. Construction of the T50/N125 Adaptor. A typical application for this adaptor is also shown.

The adaptor is designed to handle pulse or continuouswave signals originating from a 50-ohm source. It is not generally used to handle a signal traveling in the 125-ohm to 50-ohm direction, since it does not provide a termination for the 125-ohm connector. When a T50/N125 is used to connect a 50-ohm signal to the 519 signal input connector, the signal voltage is unchanged and the signal cable is fully terminated. The vertical deflection system has 50 nexe of delay before it is terminated in 125 ohms. Any reflections from the CRT, its connections, or termination, would return through the delay line and reflect from the nonterminating end of the T50/N125 to reappear at the CRT 90 to 100 nonseconds ather the original signal.

(4) Adaptor N50/N125

Also colled an unterminated adaptor, this accessory (Fig. 2-12) is a straight-thur connector which connects a 50-ohm line directly to a 125-ohm line. This unit is used primarily for pulse applications. If a pulse from 50 ohms is applied to the 50-ohm end of the adaptor, the pulse amplitude increases 1.43 times at the 125-ohm end due to the reflection at the end of the 50-ohm system. It going from 125 to 50 ohms, approximately a 0.572 transmission factor results. This is more than for any of ther adaptor ad high VSWR when used with high-frequency sine waves.



Fig. 2-12. Construction of the N50/N125 Adaptor. Also shown is a typical application for the adaptor.

When the adaptor is used for pulses, the abrupt discontinuity inherent in the unit causes a reflection to occur which may interfere with the displayed waveform unless certain precautionary measures are taken. To prevent a reflection from occurring on the displayed waveform, make the electrical length of the cable supplying the adaptor equal to ar more than 172, where T is the length of the pulse to be observed. The reflection will then appear after the displayed waveform.

TABLE 2-	-3	
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SIGNAL VOLTAGE TRANSMISSION FACTORS			
SIGNAL DIRECTION 50 Ω to 125 Ω	ADAPTOR TYPE	SIGNAL DIRECTION 125 Ω to 50 Ω	
.564	T50/T125	.225	
1.000	T50/N125	Not Used	
Not Used	N50/T125	.400	
1.43	N50/N125	.572	
1.58	Theoretical Maximum Power Transfer	.633	

(5) 125 Ω Insertion Unit

This unit is a hollow tube with 125-ohm connectors on each end and access holes located on each side to permit small components to be mounted inside. A snap-on sleeve cover permits adequate shielding of components and provides minimum discontinuity in the line. The unit facilitates 125-ohm (or 50-ohm, if desired) connections for pulse testing components such as diodes or transistors. It can also be used for testing or design of networks such as filters, attenuators, impedance-matching circuits, etc., and measurements on amplifiers and many other devices. The device makes it unnecessary to use a chasis with long leads and poor impedance matching. Instead, the components or circuit can be mounted in the small insertion unit and used as part of the 125-ohm system.

For series tests, the effective impedance of the test circuit is 22₆. For shunt tests, the effective test circuit impedance is $Z_0/2$. Thus, for tests using 125-ohm coble, a series measurement is with 250 ohns equivalent series resistance while a test from center conductor to ground is with a d25-ohm equivalent source resistance. The above conditions from the test point.

(6) 125 Ω Coupling Capacitor

The 125 Ω Coupling Capacitor connector contains a silvered-ceramic, wafer-type capacitor connected in series with the inner conductor. A slight amount of compensating inductance is supplied by the conductors butt-soldered to the capacitor. (See Fig. 2-13).

This unit is normally used for ac-coupling high-frequency signals to the Type 519 Oscilloscope with minimum reflections. Low-frequency signals and dc are blocked. Its characteristics are:

Coupling Capacitance: 0.01 μ f \pm 20%, 0.0082 μ f GMV.



Fig. 2-13. Construction of the 125Ω Coupling Capacitor.

Voltage Rating:	400 volts.
Reflections:	Negligible.
Low-Frequency Cutoff	
in 125 Ω:	Approximately 65 kc

(7) 125 Ω 1 KMC Timing Standard

The 125 Ω 1 KMC Timing Standard is a Sweep Calibrator which can be used to make periodic sweep calibration checks of the 2- and 5-nanosecond/cm rates. Refer to the Calibration Procedure section of this manual.

(8) 125 Ω Delay Cables

Different length cables are supplied for use in coupling the signal and/or trigger to the appropriate input connectors on the front panel of the Type 510 Scillocope. The loss per foot of RG-63/U cable is 0.06 db at 1 kmc. The delay time marked on each cable is the time required for a signal to travel from one end of the cable to the other.

(9) 125 Ω Cable Connector Parts

- (a) Double Button Assembly. Used for replacing a damaged or worn insert in any of the 125-ohm front-panel connectors or cable connectors.
- (b) Panel Adaptor Assembly. Replacement for any of the front-panel 125-ohm connectors.
- (c) 125 Ω Cable Connector. Replacement for any of the cable connectors.

(10) Reed Switch

Two spare reed switches are included as replacements for the reed switch used in the Calibration-Step Generator. To replace the reed switch refer to the Maintenance section of this manual.



SECTION 3

APPLICATIONS

Introduction

Information presented in this section of the manual has been selected to show typical applications for the Type 519 Oscilloscope. Among these applications are the measurements of voltage, time, and frequency. In addition, other applications are described and illustrated to point out the various features designed into the instrument. Since the number of applications for the Type 519 is large, it is beyond the scope of this manual to cover more than a few of the most general applications.

Voltage Measurements

Vertical displacements on the CRT screen are related to the applied signal voltage by means of the vertical deflection factor indicated on the CRT face mask. By means of the indicated deflection factor it is possible to make accurate voltage measurements from the oscilloscope screen. The accuracy of the voltage measurements made with the Type 519 depends to a large extent on how accurately vertical measurements are made on the oscilloscope display. It is important to obtain sufficient vertical deflection, if possible, for accurate measurements to be made. Also, care must be used so that the with of the trace is not included in the measurements. All measurements should be consistently made from the same point on the trace; for example the top, center or bottom. If the center of the trace is used for one reading, it should be used for all successive readings.

To make a voltage measurement on a waveform, the following method can be used (see Fig. 3-1).

 By using the graticule, measure the vertical deflection in centimeters between the appropriate two points on the waveform.

2. Multiply the vertical distance by the vertical deflection factor and by the attenuation factor; divide the result obtained by the amplification factor (if either attenuation or amplifiers are used). The figure obtained is the voltage difference between the two points on the displayed waveform.

As an example, assume that using a 5X attenuator, you measure a vertical distance of 1.2 centimeters between two points on the waveform. Also assume that the vertical deflection factor is 8.7 volts per centimeter. In this case then, 1.2 centimeters multiplied by a deflection factor of 8.7 volts per centimeter gives a product of 10.44 volts. This figure multiplied by the attenuation factor of 5 gives the actual voltage difference of \$2.2 volts.



Fig. 3-1. Measuring voltages from the oscilloscope display.



Fig. 3-2. Measuring the time between two events on the oscilloscope display.

Time Measurements

Due to the calibrated linear sweep rates of the Type 519 Oscilloscope, any horizontal distance on the screen represents a definite known intervol of time. Using this feature the time between two displayed events can be accurately measured directly from the oscilloscope screen. Time measurements can be made as follows (see Fig. 3/2):

 Using the graticule, measure the horizontal distance in centimeters between the two displayed events whose time interval you wish to find.

Multiply the distance measured by the setting of the NANOSEC/CM switch to obtain the actual time interval.

For example, assume that the NANOSEC/CM switch setting is 10 and that you measure a horizontal distance of 4 contimeters between the two desired points. The time interval is then 4 centimeters multiplied by 10 nanoseconds per centimeter or 40 nanoseconds.

Frequency Measurements

By using the method described for measuring time intervals, the period filme required for on cycle) of a recurrent waveform can be easily calculated since frequency is the reciprocal of the period. Frequencies which may be measured with the 519 using this method range from approximately 200 kc to over 1000 mc. For example, if the period of a recurrent waveform is accurately measured and found to be 5 nanoseconds, the frequency is the reciprocal of 5 nanoseconds, or 200 mc.

Another method which is frequently preferable is described as follows: At any given oscilloscope sweep rate, the number of cycles of the input waveform that are displayed in 5 centimeters of the screen is dependent on the frequency of the input waveform. At a sweep rate of 10 nanoseconds per centimeter, for example, 3 cycles are displayed with a 60-mc input signal 2% cycles with a



Fig. 3-3. Measuring the frequency of a repetitive signal by counting the number of cycles which are displayed in 5 centimeters. This example shows 12 cycles in 5 centimeters.

50-mc signal, and 2 cycles with a 40-mc input signal. By utilizing the pattern of these observations the frequency of a waveform can be measured by counting the number of cycles in 5 centimeters on the screen, and multiplying this by the factor given in Table 3-1 for that sweep rate [see Fig. 3-3). Since each sweep rate has a fixed multiplication factor, frequencies can easily be determined by this method ance the multiplication factors for the various sweep rates are known. The appropriate multiplies rate found by taking the treciprocal of the time required for the sweep to move 5 centimeters. The method can be summarized as follows:

 Place the NANOSEC/CM switch at a setting which permits several cycles of the input waveform to be displayed.

2. Count the number of cycles of the waveform that are displayed in 5 centimeters.

Multiply the number of cycles by the multiplication factor for the sweep rate being used.

As an example, assume that with a sweep rate of 5 nanoseconds per centimeter, 3.2 cycles are displayed in 5 centimeters. The multiplication factor for a sweep rate of 5 nanoseconds, or 40 megacycles. The frequency is 3.2 multiplied by 40 megacycles, or 128 megacycles.

FREQUENCY MULTIPLICATION FACTORS FOR 5 CM INTERVAL		
NANOSEC/CM Switch Setting	Multiplication Factor (megacycles)	
2	100	
5	40	
10	20	
20	10	
50	4	
100	2	
200	1	
500	.4	
1000	.2	

TABLE 3-1

Measuring Diode Switching Characteristics

Dynamic test measurements in the nanosecond region can be made with the Type 519 Cscilloscope to study switching and storage times in semiconductor diodes. In addition, diodes can be selected and compared for particular response characteristics. Two basic diode test circuits are illustrated here to demonstrate how the ascilloscope may be used as a pulse generator and indicator unit.

The first test circuit for diodes is shown in Fig. 3-4. This circuit can be used to measure diode reverse recovery time. In this circuit, the Calibration-Step Generator is used to abruptly shut off the forward current passing through the diode. The oscilloscope diplays the current through the diode as a function of time. The current values are then found from the display by dividing the waveform valueg by 125 ahms. From the displayed waveform on the CRT the recovery time of the diode can be measured.





Fig. 3-4. 250-ohm test circuit for measuring diode reverse recovery time. If desired, the diode and pulse polarity may be reversed to obtain an inverted display.



Fig. 3-5. Typical reverse recovery transient for a T12G diode. The upper trace shows only the Calibration-Step Generator waveform which is used to establish the initial reverse current. The lower trace shows the recovery waveform.

The resistors used in the circuit set the forward current through the diode. These resistors and the 125-bm input impedance of the oscilloscope set the forward current at 47 ma when used with a 1-50-with power supply, total power dissipated is 2.35 watts. A convenient amount of vertical deflection is obtained with a forward current of at least 30 ma. The shortest possible leads must be used to construct the system.

A variable dc supply may be used to supply multiples of 5 or 10 ma of forward current. Keep in mind that the maximum forward current must not only be within the test diade rating, but also within resistor dissipation ratings, including the 1.8 watt rating of the 125-banh termination resistor connected at the neck of the CRT. Steady current flow through the CRT termination resistor should never exceed 120 ma.

A typical reverse recovery waveform is shown in Fig. 3-5. The trace was obtained when a positive-going pulse was applied from the Calibration-Step Generator to a T12G



Fig. 3-6. Test circuit for measuring the turn-on time of a diode. If desired, the diode and pulse polarity may be reversed to obtain an inverted display.

test diode. A dc supply voltage of approximately -50 volts is connected to the diode through \$40 ohms to establish a forward reference current of -50 ma. The lower trace shows the diode recovery while the upper trace indicates the reverse current caused by the positive-going step waveform when the diode is shorted out temporarily. With the diode shorted out, the step waveform is adjusted for the desired initial turn-off current during diode recovery (+50 ma in the example). The time required from the application of the switching puble until the current through the diode reaches essentially zero or a predetermined value of current is the reverse recovery time for that diode.

The second diade test circuit, shown in Fig. 3.6, is used to turn the diade on so the turn-on time can be measured. To turn on the diade, a positive-going step is applied from the Calibration-Step Generator to the diade when it is connected directly across the 125-ohn cancial cable.

A typical resultant waveform obtained when the T12G diade is being turned on is shown by the lower waveform in Fig. 3.7 (a). The upper waveform is the step pulse generated by the Calibration-Step Generator with the diade disconnected. The fact that the voltage across the diade does not drop to zero is due to the forward drap across the impedance of the diade. The dynamic diade impedance as a function of time can be determined from the oscilloscope display by means of the following equation:

$$z = \frac{125A}{2(1-A)}$$

where $A = V_2/V_1$ as shown in Fig. 3.7. V_1 is almost constant while V_2 varies considerably with time. A plot of the diode impedance as a function of time obtained using the recovery waveform and the above equation is shown in Fig. 3.7 (b). Turn-on curves for other diodes are shown in Fig. 3.8.



Fig. 3-7. (a) Typical turn-on curve for a T12G diode on the lower trace. The upper trace shows the Calibration-Step Generator waveform. (b) A plot of impedance versus time constructed from the turn-on curve shown in (a).



Fig. 3-8. Typical turn-on curves for various types of diodes.



Fig. 3-9. Connection of a test device into the charge line of the Calibration-Step Generator.

Impedance Measurement by Reflection

A clear picture of transmission-line characteristics is made possible by the use of the Type 519 Oscilloscope. The presence of discontinuities along a transmission line can be determined while the line is under study by means of the oscilloscope display.

The Calibration-Step Generator of the Type 519 provides an excellent means for measuring the impedances of cartoni devices and cables. In an application of this sort, the device is connected as part of the charge line for the Calibration-Step Generator while the output of the step generator is applied to the oscilloscope input. If the impedance of the inserted device is exactly 125 ohms, it will mersib increase the time that the amplitude of the Calibration-Step Generator waveform remains constant. However, if the inserted device is not exactly 125 ohms or does not have a constant impedance, then irregularities can be used to determine the impedance of the inserted device and to determine whether this impedance is constant. The displayed waveform will also indicate double the delay time for the inserted device.



Fig. 3-10. Waveform obtained when a section of 50-ohm cable is connected as part of the charge line for the Calibration-Step Generator. The test device, such as a piece of coaxial cable, a connector samebby, or a delay line, can be connected into the charge line of the Calibration. Step Generator in the manner shown in Fig. 3-9. In Fig. 3-10 the Calibration. Step Generator workform is shown when a length of 50-ohm cable is connected into the charge line in series with two lengths of 125-ohm cable. The portion of the waveform due to the 50-ohm section of the waveform due to the 50-ohm cable to the cable transmitted of the cable to it is evident from the picture that the true delay time of the 50-ohm cable used is approximately 12 neac.

In Fig. 3-10 the relative amplitude of the partions of the waveform beer a definite relationship to the impedance of the device that generated that partian. The impedance of an unknown device can thus be measured by comparing the amplitude of the partian of the Calibration-Step Generator waveform produced by it against the amplitude of the initial portion due to the 125-ohm system. The method is generally limited to the first reflection, unless the deviations are small, due to multiple reflections and reflection losses.

If we call the amplitude produced by the 125-ohm system $V_{\rm O}$ and the amplitude produced by the inserted device V_{X_c} then the impedance of the inserted device is given by the formula:

$$Z = 125 \left(2 \frac{V_{O}}{V_{X}} - 1\right)$$

In Fig. 3-10 the ratio of $V_{\rm O}$ to $V_{\rm X}$ is approximately 0.7. Using this in the above formula gives the correct impedance of 50 ohms used to produce the waveform.

It is essential in applications of the type described here that no shorts, terminations, terminated adaptors, or attenuators having low shunt resistance to ground are used in the charge line of the Calibration-Step Generator. If devices

Applications—Type 519

such as these are used, they will prevent the charge line from charaing to the correct voltage and will thereby prevent the Calibration-Step Generator from producing an output waveform. Where it is necessary to match one type of connector to another, unterminated adaptors should be used. While a series coupling capacitor (125 Ω connectors) may be used to allow the step generator to function, experience will show that the capacitor cannot retain its initial charge indefinitely, so voltage sag across its terminals will show in the observations. Further, the coupling capacitor is already quite large to be charged between operations of the reed switch at 750 cycles per second. External charging voltage connected through a 5000-ohm resistor by a short lead to the center conductor of an insertion unit will permit better operation with the coupling capacitor. The insertion unit must appear on the reed-switch side of the capacitor and the short-circuit or shunt resistance must be connected beyond the series capacitor.

Obtaining Information from Small Deflections

The vertical deflection factor of the Type 519 Oscilloscope is approximately 10 volls per centimeter. Consequently, very small deflections in the oscilloscope display may be important. Careful analysis and proper techniques will allow you to obtain a great deal of information from these small signals. The triggering circuits of the Type 519 are sufficiently sensitive that signals which produce only a few trace widths of deflection will provide a table display.

One of the more obvious means of recovering information from small deflections involves the use of photographic enlargement. Here, the oscilloscope display is photographed using a high quality system and the photograph is then enlarged to a convenient size. Measurements may then be



Fig. 3-11. A typical cross-feed manipulator and microscope which can be used to obtain information from small deflections by the trace-splitting technique.

made from the enlarged photograph. The distance between graticule lines on the enlarged photograph can be used to determine the exact enlargement factors so that distances on the enlarged photograph can be readily converted into measurements of voltage and time. The primary difficulty with this method is the delay involved in obtaining enlarged prints.

A second method for obtaining information almost immediately from small deflection involves the use of a device such as a cross-feed manipulator, as shown in Fig. 3-11. A Polaroid® photograph is taken and mounted on the table of the cross-feed manipulator under a microscope containing cross hairs. The table is then adjusted until the center of the trace are one point of measurement lies directly under the cross hairs. The calibrated dials are then set to zero. The manipulator is then adjusted until the second point of measurement lies under the cross hairs and a second reading on the calibrated dials is made. This distance multiplied by the appropriate deflection factor gives the actual time or valance between the two points.

In both methods described, a line is imagined to run through the exact center of the oscilloscope trace. This line splits the trace into two halves and consequently the technique of measurement involving this imaginary line is sometimes referred to as "trace splitting". All measurements are mode with respect to the center of the trace and thus to the line passing through the center of the trace Measurements mode with respect to the trace-splitting line are much more accurate than those obtained from, say, one side of the trace.

How well the position of the trace-splitting line can be determined will depend to a large extent on the care with which the measurements are made and upon how sharply the trace is focused. Obviously, the better the trace is focused the easier it is to determine exactly where the center of the trace is located. Care should be taken to obtain the best possible trace and camera focus. It is not necessary to actually draw in the trace-splitting line. In most cases this would be very difficult, if not impossible. It is only necessary for you to make all measurements under the microscope for from an enlarged photograph) from the exact center of the trace.

Substitution Method of Frequency Measurement

Occasionally you may want to measure the average repetition frequency of a random input signal. One means of doing this involves the use of the Rate Generator in the Type 519, or some external signal generator with calibrated frequencies. In this application, the random signal is first used to trigger the oscillocope. The output from the DE-LAYED +GATE connector is then used to charge a capacitor. The circuit is shown in Fig. 312. The voltage to which the capacitor charges will depend on the average repetition rate of the signal and on the setting of the NANOSEC/ CM switch. If then the oscillocope is triggered from the Rate Generator output or from the external signal generator, the frequency can be adjusted to give the same voltage across the capacitor.
 R
 From DELAYED

 + GATE 50 Ω
 Connector

 Correction
 Connector

 0r Q6 - 1000
 Connector

Fig. 3-12. Circuit used to measure the average repetition rate of a random signal using the substitution method. The circuit and technique can also be used to measure the frequency of periodic signals with frequencies too low to be measured directly from the oscilloscope screen.

When the same voltage is obtained, the repetition rate of the Rate Generator or external signal generator is the same as the average repetition rate of the random signal. The repetition rate can be read from the Rate Generator or axternal signal generator controls. A VTVM is used to measure the voltage across the capacitor in both cases. The average repetition rate of the random signal cannot exceed the highest repetition rate of the Rate Generator (approximately 30 kc) if the Rate Generator is used for the frequency comparison. Care must be taken to insure that the oscilloscope is triggered once for each random input signal. The value of the capacitor used in the application must be chosen to give a substantial voltage reading so that reasonably accurate comparisons. Can be made.

The settings of the NANOSEC/CM control must be made as a compromise between two factors. Since the width of the gate obtained from the DELAYED +GATE connector depends on the setting of the NANOSEC/CM switch, the switch must be set to produce a substantial charge on the capacitor. The maximum charge is obtained with the widest gate and thus the slowest sweep. However, in order to measure fairly high frequency signals (up to 400 kc) the NANO-SEC/CM switch should be set to allow the highest possible sweep repetition rate. The maximum possible sweep repetition rate is obtained when the NANOSEC/CM switch is set for the fastest sweep. A typical value for the capacitance is 0.5 uf. The resistance is 10 k. If the voltage reading on the meter drops below 1 volt, short out the 10-k resistor. This method should permit you to cover the complete range of 3 cps to 400 kc.

An example of the use of the substitution method is shown in Fig. 3-13. In this application a random repetition-rate signal is being generated by a high-voltage corona discharge between a metal plate and a discharge rod. The metal plate is connected to a +20-kv voltage source. The discharge rod is mounted 4/₃ inches from, and perpendicular to, the metal plate. The corona signal is applied to the SICNAL 125 Ω connector of the Type 519 through a 125-ohm coaxial cable.

The average repetition rate of the corona discharge in this example was measured and found to be 2300 pps.

Use of the Instrument in Rapid-Changing High Fields

The Type 519 Oscilloscope is a well shielded instrument. However, shielding provides only attenuation and not complete exclusion of extraneous fields. Therefore, in the presence of very large fields such as those set up by large



Fig. 3-13. Using the substitution method for determining the average repetition rate of a corona discharge signal.

surge generators, erratic triggering may result as the triggering circuits operate from energy obtained from the external fields. It is important to recognize that the presence of a strong external field can adversely affect the operation of the instrument unless adequate precautions are taken.

Extremely large fields may produce false signals on the oscilloscope screen. If difficulties in triggering or false waveforms result from external fields, it will be necessary to further shield either the instrument or the generator, or to move the instrument to a point farther away from the source of the stray field.

Use of Ferrite Cores

In many applications for the Type 519, large ground currents are present. These ground currents enter the Type 519 either from the power line or signal cables, and may cause stray triggering or enter into the oscilloscope dipploy. Stray signals generally enter the signal circuit through leaky coaxial cables or poor grounding. Solid cable is better than braided cable because it has lower outer conductor impedances and no electrosticit leakage. Corre must be taken not to include transient voltage drops in inductive ground connections into the signal-input circuit.

Large transient ground currents can frequently be attenuated by passing either the power cord or signal cooxial cables through a ferrite torroid. The impedance to ground current is related to the square of the number of times the lead passes through the core. The power or signal is not affected by the core. Ferrites should be placed as near as possible to the connections to equipment under test.

NOTES



SECTION 4

CIRCUIT DESCRIPTION

BLOCK DIAGRAM

A simplified block diagram of the Type 519 Oscilloscope is shown in Fig. 4.1. This diagram can be used to gain a general understanding of the operation of the instrument after which the schematic diagrams at the rear of the manual can be used for more detailed information. The information which follows describes briefly the function and purpose of each of the blocks shown in the block diagram. This is followed by a detailed circuit description of the instrument. Input signals to the Type 519 are applied to the SIGNAL I250 a connector on the front panel of the instrument. The signal is then applied through a trigger takeoff and a 45nsec delay line to the terminated 125-ohm vertical deflection system of the CRI. The delay permits the horizontal sweep to be started before the vertical signal arrives at the CRI. The trigger takeoff obtains a sample of the input signal which is then applied to the Trigger Channel as a triggering signal.

A TRIGGER SOURCE switch in the Trigger Channel selects the triggering signal used to initiate or synchronize the hori-



Fig. 4-1. Type 519 Oscilloscope simplified block diagram.

zontal sweep. Possible triggering signal sources are the vertical signal, external trigger inputs, the Rate Generator, and the Calibration-Step Generator. The selected trigger signals are amplified, or attenuated, to the required level in the Trigger Channel and then applied to the Trigger and Holdoff Circuit. When very high frequency signals are used, a special countdown circuit in the Trigger Channel reduces the signal frequency applied to the trigger circuits. The countdown circuit parts frequency trequencies to Zhmc.

The triggering signal obtained from the Trigger Channel is used to initiate or synchronize the operation of the First Regenerator Blocking Oscillator. The blocking oscillator produces an output waveform with constant amplitude and shape regardless of the shape or amplitude of the triggering signal. A holdfof circuit prevents the blocking oscillator from being triggered again before it and the sweep circuits have had a chance to reset after a sweep. The holdfof (cicuit also permits single sweeps to be generated by the instrument. Because of the holdfof (circuit, the maximum repetition rate of the blocking oscillator (and the sweep), at 2 nanoseconds per centimeter, is approximately 400 kc. The blocking oscillator is capable of counting down from frequencies up to approximately 50 mc.

The output from the First Regenerator Blocking Oscillator operates a delay circuit which then operates the Second Regenerator Blocking Oscillator after a variable delay time. Since the sweep can thus be delayed through 35 anaseconds with respect to the trigger signal, this ultimately permits the vertical signal to be positioned horizontally within the oscillacore trace. The output from the Second Regenerator Blocking Oscillator is amplified and applied to the Unblanking and Time-Base Gate circuits.

When triggered by the output of the Second Regenerator Blocking Oscillator, the Unblanking Circuit opplies a gate to the cathode of the CRT to unblank the beam for the duration of the horizontal sweep. The time-base gate which is produced by the gate timer circuit is applied to the Time-Base Generator where it is used to gate on the sweep. The duration of the gate is dependent on the setting of the NANOSEC/CM control.

The Time-Base Generator in the Type 519 consist essentially of a clamp tube, timing resistors, the capacitance in the plate circuit of the clamp tube, and a bootstrap circuit. When the clamp tube is gated off by the output of the Time-Base Gate Generator, the plate capacitance of the clamp tube charges, producing the sweep sawtooth. A special bootstrap circuit is used to apply the sawtooth to the positive end of the timing resistor, thereby producing a linear sawtooth by maintaining the charging current for the capacitance relatively constant.

The plote voltage of the clamp tube is regulated by a feedback loop which operates on the grid of the clamp tube. Regulation of the clamp tube plote voltage results in a constant starting voltage for the sweep sourcoth waveform. In addition, a separate regulator circuit automatically adjusts the screen voltage to maintain the control grid of the clamp tube at -32 volts regardless of the setting of the NANOSEC/CM switch. This maintains the correct operating point for the clamp tube.

The sawtooth waveform which is generated is applied to a paraphase amplifier and then to the horizontal deflection plates of the CRT to produce the horizontal sweep. The high accelerating potentials required for operation of the CRT are supplied by the regulated High-Voltage Power Supply. Other operating potentials used by the Type 519 are obtained from the Low-Voltage Power Supplies.

Two signal generators are contained in the Type 519. The Colibration-Step Generator produces output steps which are continuously variable in amplitude and accurately calibrated. The repetition rate of the output steps is approximately 250 steps per second and the risatime is approximately 0.1 anosecond. The Rate Generator produces output putses which are fixed in amplitude but variable in repetition rate. Repetition rate is variable between 3 cps and 30 kc. The risetime of the Rate Generator waveform is nominally 0.5 nanosecond.

VERTICAL SIGNAL CHANNEL

Trigger Takeoff

 $^{\prime}$ Input signals applied at the SIGNAL 125 Ω connector are possed through a trigger takeoff before being applied to the input of the 45-nanosecond delay line. The purpose of the input signal which can then be applied to the oscilloscope triggering acrouits.

A drawing of the trigger takeoff is shown in Fig. 42. A gap is made in the outer conductor of the US 0. coaxial coble while the inner conductor is not disturbed. Between two metal rings at the gap in the outer conductor of the coaxial coble are connected eight short jumpers. Each jumper is connected through a small ferrite core. An output lead is connected through each of the eight cores in series. The eight cores and leads form eight small 1-1 transformers.

The output load on the trigger takeoff is 125 ohms. Consequently, due to the series arrangement, the impedance reflected back into the primaries of each of the small transformers is approximately 16 ohms. Core impedances reduce this primary impedance. The eight small transformers connected in parallel across the gap in the outer conductor of the coaxial cable actually present a total impedance of approximately 1.5 ohms with a decay time constant of 30 nanoseconds. The slight series impedance results in less than 1%, reflection. Over 9%, of the input signal is transmitted through the trigger takeoff into the 45-nanosecond delay line.

The voltage developed across the gap in the caxial cable due to the trigger tacked is approximately 1.3%, of the input voltage. This voltage appears across each of the eight 1.1 transformers connected across the gap. Since the takeoff loop is connected in series through all eight transformers, the voltages of each of the transformers are additive. The net result is that the total output of the trigger takeoff is 8 times 1.3%, or approximately 10%, of the input signal voltage. This triggering signal is obtained at the expense of approximately 0.7%, attenuation of the signal capitel to the ascillacoce.

Since the gap is short-circuited by the cover, the only thing preventing the gap from being shorted out is a small amount of inductance in the outer conductor inside the cover. At low frequencies the inductance in the outer con-



Fig. 4-2. Construction details of the trigger takeoff. The center portion is expanded to show details of the wiring. Only four of the eight jumper wires and small cores are shown for simplicity.

ductor would be insufficient to prevent the gap from being shorted out and consequently no voltage would be developed across the gap. The net result is that no output could be obtained from the trigger takedif. To extend the low-frequency response of the takedif, large ferrite cores are placed around the cable on bath sides of the gap inside the cover. The cores increase the inductance of the outer conductor and thus permits the trigger takedif to operate at much lower frequencies. These cores do not affect the input signal.

Delay Line

Signals transmitted through the trigger takeoff are applied through the 43-nanosecond delay line to the vertical deflection system of the CRT. The delay line is a high-quality, low-loss coaxid line which is especially selected for minimum deterioration of input signal waveshape. In addition, special care is taken in the design of the sweep circuit to insure that the sweep is started as soon as possible after a triggering signal is applied. This permits a minimum length of delay cable to allow display of the leading edge of the signal which produces the trigger.

Cathode-Ray Tube

A distributed deflection system is used in the CRT of the Type 519. The upper vertical deflection plate is, in effect, a large number of very small deflection plates. The signal is applied only to this upper deflection system. The lower deflection plate is internally bypassed to ground and is used only to apply the vertical positioning voltage.

The distributed deflection system is so designed that the velocity of propagation of the signal toward the screen is the same as the velocity of the beam electrons in the deflection system. The overall effects of the large number of effective vertical deflection plates is that their individual deflections are additive. By this means the sensitivity of the CRT is kept relatively high while transit time effects and capacitance in the plates are minimized.

Since the deflection system of the CRT is part of the 125ohm vertical deflection system, it is important that the deflection system also present a constant impedance of 125 ohms. Any deviation from this impedance would cause reflections from the deflection system and would result in distortion of the displayed waveform. To insure that the deflection system in each CRT is exactly 125 ohms, each is tuned by means of 27 trimmers for least reflected energy. This is done before the deflection system is seeded into the envelope of the CRT and is again quality-checked after final processing of the CRT.

After the signal has passed through the deflection system of the CRT it is terminated by a 125-ohm resistor connected to the side of the CRT. Thus the signal energy traveling through the deflection system is absored rather than being reflected back into the deflection system.

TRIGGER CHANNEL

Trigger Source Switch

Triggering signals obtained from the Rate Generator, Calibration-Step Generator, EXTERNAL TRIGGER 125.0 connector, and from the vertical input signal are applied to the RIGGER SOURCE switch. The TRIGGER SOURCE switch determines which triggering signal is applied into the Trigger Channel.

When triggering signals obtained from the vertical input signal, the EXTERNAL TRIGGER 125 Ω connector or the Calibration-Step Generator are not used, the triggering signals are terminated by 120 ohms. A termination is not necessary when the Rate Generator triggers are not used.

Since the Calibration-Step Generator is capable of producing very large steps, an attenuator is placed in the trigger signal lead running to the TRIGGER SOURCE switch. The attenuator, composed of R6 and R7, prevents the triggering signal applied into the Trigger Channel from exceeding about 2 volts. Larger signals could conceivably cause damage in the Trigger Channel in GAIN switch positions other than X.2.

Resistors R8. R9, and R15 are added in series with the triggering signal leads to damp unwanted resonances in unused sections of the TRIGGER SOURCE switch.

In addition to its primary function of selecting the triggering signal, the TRIGGER SOURCE switch also selects the triggering signal slope which produces triggering. In the + positions the positive or rising slope is selected, while in the - positions the negative or falling slope is selected. Positive-slope triggers are required to actually trigger the blocking oscillators in the Trigger Channel, Consequently, when a negative slope is selected, the signal must be inverted to convert the negative slope to a positive slope.

Inversion of the triggering signal is accomplished in the 125-ohm transmission line between the TRIGGER SOURCE and GAIN switches. One conductor of the transmission line is grounded, while the other conductor is connected to the GAIN switch. Triggering signals may be applied to either of the two transmission line leads depending on the position of the TRIGGER SOURCE switch. When the triggering signal is connected to the ungrounded conductor of the transmission line, the signal is passed through the line without inversion. However, when the signal is connected to the arounded conductor, the signal applied to the GAIN switch is inverted. When the triggering signal is connected to the arounded transmission line conductor, the small endto-end inductance of the line prevents ac signals from being shorted out. The transmission line is passed four times through a small ferrite core to increase the inductance of the lead and thus extend the low frequency performance of the inverting network.

Gain Switch

Since triggering signals selected by the TRIGGER SOURCE switch may have a large range of amplitudes, some means must be provided for altering these signal amplitudes to satisfy the requirements of the triggering circuits. Amplifiers and attenuators switched in by the GAIN switch perform this function.

Four positions are provided in the GAIN switch. Amplifier and attenuator connections for each of these positions are as follows:

GAIN Switch Setting X.2 X5 Attenuator NORMAL Sianal Unchanaed X5 X4 Attenuator and X20 Amplifier X20 X20 Amplifier

The various positions of the GAIN switch allow triggering signals arriving at the switch with between 20 my and 10 volts of amplitude to trigger the oscilloscope.

X20 Amplifiers

Two identical wideband trigger amplifiers are arranged to permit stable triggering from small input signals. The amplifiers are arranged on plug-in boards and are interchangeable. Each offers approximately 20 times gain with an upper 3-db point beyond 100 megacycles. The high frequency input impedance of each amplifier is 125 ohms. matching the trigger system transmission line impedance. The amplifiers employ feedback, degeneration, and frequency-sensitive networks to maintain good transient response, bandwidth, and input impedance. Excellent agin stability and dc-voltage stability results.

The input impedance of 125 ohms is closely maintained at various frequencies by frequency-sensitive networks L30, R30, R32, and C32. At the highest frequencies, the signal is applied through C32 directly to the base of Q34 while L30 and R30 supply additional termination for slower input signals which are not easily passed by C32. Feedback through R45 also contributes to the termination of slower trigger signals.

Signals applied at the base of Q34 appear amplified at the collector and are directly coupled to the base of Q44, where they are further amplified. The output of the amplifier is coupled through C46. Each amplifier stage causes a phase reversal but since two stages are used, the output signal has the same polarity as the input sianal.

DC operating voltages are stabilized by emitter-circuit degeneration in both Q34 and Q44. At low frequencies R43 produces a slight amount of degeneration in the emitter circuit of Q44. At high frequencies, R43 is bypassed by C43 and the reduced degeneration helps to compensate for the high-frequency loss in gain in transistors Q34 and Q44. R32 and C32 also tend to compensate for this loss of gain in the transistors. L37 adds high-frequency peaking at the collector of Q34.

A dc potential of -0.4 volt appears at the input of each amplifier. This voltage is necessary for proper action of the pulse-amplitude-selecting diodes, D68 and D69, which precede the Second X20 Amplifier. Diode D69 must be forward biased to permit trigger signals to be applied to the second amplifier.

A voltage divider consisting of R47, R37, and R36 supplies the collector voltage for Q34. The values for the resistors are chosen to provide the correct voltage for the base of Q44.

Function Switch

Signals from the output of the GAIN switch and the optional First X20 Amplifier are applied through a 125ohm transmission line to the FUNCTION switch. In the PULSE and SYNC positions of the FUNCTION switch the triggering signals are applied to the Second X20 Amplifier through the pulse-amplitude-selecting diodes. Front panel selection of trigger level by R66A changes the current through R64 and R63, thereby selecting the trigger height required to send a signal through diode D69 into the Second X20 Amplifier. In the SYNC position, pulse-ampli-tude-selecting diode D69 is forward biased by R63 to pass very small triggers. In the HF SYNC position, triggering signals are connected directly to the Countdown Oxcillator: The output of the Countdown Oxcillator is then applied through D69 to the Second X20 Amplitier. When the Countdown Oxcillator is used, it free-runs at approximately 30 mc but is synchronized by the trigger signal. The countdown operation assures that the output of the Second X20 Amplifier does not exceed 30 mc.

In the PULSE position of the FUNCTION switch, resistors R63, R64, and R66A form a voltage divider between +225 and -265 volts. The divider sets the voltage at the iunction of D88 and D69. Varying the setting of R66A changes the bias applied to the two diodes. When the junction of D88 and D69 is negative with respect to ground D68 conducts, shunting any small incoming signals through R68 to ground. However, if the positive trigger signal is large enough, it will cost D69 to conduct, thereby passing the positive positive than approximately -0.2 volt will cause D69 to conduct, thereby passing the positive portion of the signal on to the Second X20 Amplifier.

By varying the setting of R66Å it is possible to determine the signal amplitude required to cause D64 to cutoff and D69 to conduct. The circuit can be set to pass a certain signal amplitude while rejecting all signals with less amplitude. This amplitude-level-selection feature is necessary to permit stable triggering from a wide range of signal amplitudes. When maximum triggering sensitivity is required, R66Å is set to the grounded end. This places a slightly positive potential at the junction of D68 and D69, thereby allowing all positive trigger signals to pass to the X20 Amplifier without significant amplitude reduction.

In addition to the operations already described, the FUNCTION switch also performs several other operations. These other operations will be described in conjunction with the circuit most nearly related to the operation.

Countdown Oscillator

The Countdown Oscillator (DS0) circuit is connected into the trigger channel only in the HF SYNC position of the FUNCTION switch. Voltage is applied through R56 and L53 to the cathode of the tunnel diode, D50. The circuit configuration causes the Countdown Oscillator to free run at approximately 30 mc. The Countdown Oscillator can be synchronized by trigger signals with frequencies up to approximately 2 kmc. The PULSE AMPLITUDE OR SYNC and VERNIER SYNC controls vary the basic frequency of the Countdown Oscillator slightly to permit steady synchronization to be obtained.

Fig. 4.3 shows a simplified diagram of the Countdown Oscillator. The 5.5-ohn load is produced by R54 and R55 in parallel. Approximately 0.2 volt is obtained from a voltage divider, consisting of R56 and the parallel combination of R54 and R55, from —26.5 volts. The PULSE AMPLITUDE OR SYNC control, R66A, and the VERNIER SYNC control, R67, also exert some influence on the voltage applied to the tunnel didea, ellowing the frequency of the oscillator to be varied as required for stable synchronization by the incoming trigger signal.

Inductor L53 plays an extremely important part in the operation of the tunnel diode oscillator. As power is first applied to the circuit; the voltage across the tunnel diode



Fig. 4-3. Simplified circuit diagram of the Countdown Oscillator.

builds up until the voltage passes over the peak of the diode characteristic curve, shown in Fig. 4.4. This places the diode in its negative resistance region. Any further increase in voltage would cause the current through the diode to decrease. However, as the diode current starts to decrease, the current through LS3 is maintained as its fliux starts to collapse and opposes a change in current. The current through LS3 not taken by D50 flows into stray capacitance and charges the circuit to the voltage at point B where all the current through LS3 is again required by the diode.



Fig. 4-4. Operating cycle of the Countdown Oscillator. The diagram shows the voltàge-current curve for the tunnel diode with the operating cycle indicated.

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As current through L53 decays, the voltage across the diode decreases until point C on the diode characteristic curve is reached. At this time, the voltage again forces the diode into its negative resistance region. Any further decrease in voltage would then cause an increase in diode current. Because L53 does not permit a rapid change in current, the excess diode current discharges circuit capacitances to the voltage at point D in the diagram. When the voltage and current reach point D, the diode voltage and current again slowly build up until the voltage again goes over the peak at A into the negative resistance region. At this point the cycle of operation starts over. The result of the foregoing operation is a continuous oscillation of the output of L53 about the cycle shown in Fig. 4-4. The output signals from the oscillator are approximately 0.5 volt in amplitude.

Varying the voltage applied to the tunnel diade oscillator by means of &66A and &67 selectively changes the time required for the transition from D to A and from B to C, thereby changing the frequency of oscillation. By changing the frequency of oscillation, the oscillator can be brought within the range where it can be synchronized by the applied trigger frequency.

Trigger signals to the Countdown Oscillator are connected from a 125-ohm transmission line through a highpass filter and R51 to the cathode of D50. The filter consists of C51 and L50, and is used to isolate the Countdown Oscillator output from the trigger input circuits. Resistors R50 and R51 are used to terminate the transmission line to the Countdown Oscillator. Inductor L52 isolates the Countdown Oscillator from the output transmission line.

In operation, the input triggering signals are added to the voltages appearing across the tunnel diode. The triggering signals can thus cause the tunnel diode to switch slightly ahead of the time that it normally would. The frequency of the Countdown Oscillator can be adjusted to give stable synchronization.

Second X20 Amplifier

The input to the Second X20 Amplifier is obtained from either the input triggering signals or from the Countdown Oscillator, depending on the setting of the FUNCTION witch. The selected signal is amplified and applied to the First Regenerator Blocking Oscillator. The operation of the Second X20 Amplifier is the same as the First X20 Amplifier.



Fig. 4-5. Trigger Circuit block diagram.

TRIGGER AND HOLDOFF CIRCUITS

First Regenerator Blocking Oscillator

Fig. 4.5 shows a block diagram of the Trigger and Holdoff circuits. Trigger signals from the output of the Second X00 Amplifier are applied through C70 to a trigger amplitude limiting circuit. Low amplitude positive triggering signals are passed through diades D70 and D71 to one of the windings of T70. Large positive signals cause the junction of D70 and D71 to go more positive, threeby causing D71 to turn off. This prevents the large amplitude signals from passing through D71 into T70. Diade D70 cuts off to prevent large amplitude negative signals from being passed.

Only positive signals are effective in initiating the operation of Q70. When a positive triggering signal is applied, current from R71 and D71 which normally flows through 170 and R72 to ground is interrupted. Field callapse in 170 then induces a negative voltage at the base of Q70. This causes Q70 to conduct, thereby permitting current to flow through the collector winding to T70. This induces a still greater negative voltage at the base causing the transistor to conduct more heavily. The regenerative action continues until Q70 saturates. The heavy conduction by Q70 starts to discharge C77. As C77 discharges, the output voltage from T70 is decreased. This causes the drive to the base of Q70 to be reduced, and the blocking oscillator is thereby reset. The cycle of operation is then completed and C77 recharges.

In order to prevent a large backswing in the output due to collapse of the field about 170, diade D72 is placed directly across one of the transformer windings. On the backswing, the diade conducts, effectively shorting out the voltage.

At the completion of each cycle of operation, the First Regenerator Bioching Oxcillator is prevented from operating again for a certain time interval by the Holdoff Circuit. The purpose of the Holdoff Circuit is to issure that the operating frequency of Q70 does not exceed the limits of the circuits which follow. The maximum repetition rate permitted by the Holdoff Circuit at a sweep rate of 2 msc/cm is opproximately 400 kc. If triagering signals are applied with frequencies higher than 400 kc, the circuit counts down.

The bias on the base of Q70 is controlled by a voltage divider network between ground and +225 volts and by the output of the Holdoff Circuit. In the PULSE position of the FUNCTION switch, the setting of R668 influences the bias on Q70. By raising or lowering the base voltage on Q70, R668 determines how much amplitude the triggering signal must have in order to trigger Q70. When R668 is set for minimum resistance, the voltage at the base of Q70 is insufficient to keep the transistor turned off and the circuit free runs at a rate determined by the Holdoff Circuit.

In the SYNC position of the FUNCTION switch, the junction of R155 and R156 is grounded, allowing Q70 to free run. The rate that the circuit free runs is again determined by the Holdoff Circuit. The length of holdoff can be varied by R66A and R67 through R60 and R61, respectively, to permit the circuit to be synchronized by incoming trigger signals.

Resistors R155 and R156 determine the triggering level when the HF SYNC position of the FUNCTION switch is used. The level selected gives correct triggering for amplified signals from the Countdown Oscillator.

Two blocking oscillator outputs are obtained from separate windings of T70. One output is applied to the Holdoff Multivibrator and + Trigger Cathode Follower while the other output operates the Sweep Delay Circuit.

+ Trigger Cathode Follower

A positive output from the First Regenerator Blocking Oscillator is applied through RS and R145 to the grid of V1438. The output signal from the cathode is applied through a length of 50-ohm cable to the \pm TRIGGER 50 Ω connector on the front panel. The output impedance of the cathode follower is approximately 50 ohms during the time the tube is conducting.

Holdoff Multivibrator

A simplified diagram of the Holdoff Multivibrator Circuit is shown in Fig. 4-6. In the normal static state, V114 is cut off and V134 is conducting. A common cathode resistor, R16, insures that both tubes to not conduct simultaneously. The grid of V114 is held approximately 4 volts negative by a voltage divider between -2.5c volts and ground constring of R71, D71, and R72. The grid of V135 is clamped at ground posterial by the grid-to-cathode portion of V138 which operates as a diode. Since the grid of V134 conducts. This causes the common cathode of V114 and V134 to be at approximately ground potential, holding V114 in cutoff.

With V114 cut off, the grid of V123A is at approximately \pm 100 volts, thereby setting the cathode of V123A data at approximately \pm 100 volts. With the control grid of V134 clamped at ground, and the cathode of V123A at \pm 100 volts. capacitors C126 and C123 are charged to approximately 100 volts.

When Q70 operates, a positive pulse of approximately 8 volts is applied thorugh R75 to the control grid of V114. The amplitude of this pulse is sufficient to bring V114 out of cutoff. As V114 conducts its plate voltage drops, causing the cathode voltage of V123A to also drop. The drop in voltage at the cathode of V123A to also drop. The drop involtage at the cathode of V123A to start discharging through R126 and R125. As the capacitors discharge, the voltage at the grid of V134 rises slowly, eventually permitting V134 to again conduct, thereby cutting off V114 ance more. The length of time required for C126 and C123 to discharge sufficiently to permit V134 to conduct is de termined by the voltage at R124. R125. R126, C123, and C126, and by the voltage at R124.

The values of R126 and C126 are selected by the setting of the NANOSEC/CM switch. The length of time that V134 remains cut off is therefore also determined by the setting of the NANOSEC/CM switch. Voltages applied from the VERNIER SYNC control, R67, and the PULS AMPLITUDE OR SYNC control, R667, are applied to the



Fig. 4-6. Holdoff Circuit simplified circuit diagram.

junction of R126 and R125. Both of these controls influence the voltage at this point and thus the time that V134 remains cut off. By modifying the holdoff time provided by the Holdoff Multivibator, the VERNIER SVNC and PULSE AMPLITUDE OR SVNC controls can be used to slightly control the repetition rate of the First Regenerator Blocking Oscillator, G70, in order to obtain stable synchronization.

Tube V123A is used in the circuit to provide a low impedence charge path for C123 and C124. This allows the two capacitors to repidly charge to the full 100 volts when V134 conducts. It is important that the capacitors regain their full charge rapidly in order for the multivibrator to be prepared for the next pulse from Q70. Capacitor C113 acts as a boostrap capacitor for V114 and is used to speed the rise of the pulse the plate of V114.

As previously mentioned, the grid and cathode of V1238 are used as a diode which clamps the grid of V134 at ground potential when V134 is conducting. In addition, the plate circuit of V1238 is used to light the READY lamp on the front panel of the instrument. The lamp lights when V134 is conducting to indicate that the holdoff period has elapsed and that the Holdoff Multivibrator has been rest.

Holdoff Cathode Follower

When the Holdoff Multivibrator is triggered by the output of Q70, the rise in plate voltage of V134 is applied to the grid of Holdoff Cathode Follower, V143A. The resulting positive rise in voltage at the cathode of V143A is then applied to the base of Q70 holding the transistor cut off.

As the cathode voltage of V143A rises, capacitors C143 and C142 are charaed rapidly through the cathode follower to the cathode voltage. Later, as the Holdoff Multivibrator resets, the grid voltage of V143A drops to its original level. Because of capacitors C142 and C143, however, the cathode of V143A is unable to follow, and the cathode follower cuts off. Capacitors C142 and C143 then slowly discharge through the cathode resistors, gradually permitting the base voltage of Q70 to approach the point where it can again be triggered. The effect of the capacitors in the cathode circuit of V143A is to add an additional holdoff time to the holdoff time provided by the action of the Holdoff Multivibrator. The purpose of this additional holdoff time is to permit the Holdoff Multivibrator to become reset and C123 and C126 to become fully charged before Q70 is permitted to operate on the next cycle.
The value of C143 is determined by the setting of the NANOSEC/CM switch. The extension of the holdoff time provided by the Holdoff Cathode Follower is, therefore, selected by the setting of the NANOSEC/CM switch to provide the necessary charging time for C123 and C126.

Single-Sweep Reset

When the NORMAL—SINGLE SWEEP switch is placed at SINGLE SWEEP, R124 is disconnected from +100 volts and reconnected to -26.5 volts. The resulting decrease in the voltage at the grid of V134 course V134 to normally could of the optimum of the

In the SINGLE SWEEP position, the NORMAL-SINGLE SWEEP switch also applies power to Q160. Since Q160 is normally not conducting, C164 and C168 both charge to 26.5 volts. The base of Q160 is held at ground by the connection through R163. When the RESET switch is closed, C168 discharges through R168. As C168 discharges, the voltage at the base of Q160 goes negative, bringing the transistor into conduction. When Q160 conducts, the increase in current through T160 induces a more negative voltage at the base of Q160 causing the transistor to conduct more heavily. This then induces a still greater negative voltage at the base. As C164 discharges into the T160, the output voltage and regeneration base drive decrease. This decrease in the base drive causes the transistor to start to turn off. This induces a positive voltage at the base of Q160 which cuts off the transistor. When the transistor reaches cutoff, the cycle is completed.

If, after a switching cycle, the RESET button is still depresed, Q160 goes back into conduction. However, C164 was discharged by the switching cycle and is therefore unable to support another cycle until sufficient time has elapsed for the capacitor to recharge. Transistor Q160 continues to conduct as long as the RESET button is depresed. The voltage drop across R164 prevents C164 from charging back to -26.5 volts again until the switch is released. The lack of charge on C164 prevents Q164 from charging operation more than once each time the RESET button is presed.

An output pulse from Ti60 is applied to reset the Holdoff Multivitator. A voltage divider consisting of RI60 and RI61 between ground and -12.6 volts biases the junction of diodes DI60 and DI61 at about -9 volts. Diode DI61 charps the amplitude of the pulse at ground level. DI60 resets the circuit only if the cathode voltage level is below ground.

When V134 is conducting, the common cathodes of V114 and V134 are at approximately ground potential. When V114 is conducting the common cathode is about 3 volts negative. Diode D160 is thus held cut off at all times except when it receives the reset pulses from T160. Diode D161 shunts positive signals larger than approximately 9 volts to ground.

When the RESET button is pressed, the positive output from T160 is applied to the common cathode circuit causing V114 to cut off. The rise in plate voltage of V114 causes the cathode voltage of V123A to go to about +100



Fig. 4-7. Sweep Delay Circuit simplified circuit diagram.

volts. Resistors R124 and R125 form a voltage divider between -265 volts and the -100 volts on the cathode of V123A. The voltage divider places a positive voltage on the grids of V134 and V1238 causing them to conduct. This in turn lights the READY lamp and permits Q70 to be triggered on the next triggering signal. When Q70 is triggered, the sweep runs and the positive output from T70 again causes V114 to go back into conduction, cutting off V134. The Holdoff MultiVartator remains in this condition until the RESET button is again pressed to reset the multiVibrator.

Sweep Delay Circuit

A simplified diagram of the Sweep Delay Circuit is shown in Fig. 4-7. A voltage divider between +100 volts and -26.5 volts initially sets the voltage at the junction of D81 and D82. The voltage at this point in the circuit can be varied by means of the DELAY control, R88. The voltage varies from around potential to several volts positive. When Q70 goes through its blocking oscillator cycle, a negative pulse at the junction of D80 and D81 causes D81 to cut off, thereby interrupting the voltage divider network. The sudden drop in current through R82 produces a small negative voltage step at the cathode of D82. This is followed by a negative ramp as C83 charges through R83. When the voltage at the cathode of D82 becomes negative the diode conducts, producing a negative voltage at the base of Q180. When the voltage at the base of Q180 goes sufficiently negative to allow the transistor to conduct, a blocking oscillator cycle is started.

The time duration between the operation of Q70 and the triggering of Q180 is largely determined by the initial charge on C83. The charge on C83, however, is determined by the setting of fthe DELAY control, R88. By changing the setting of R88, the starting point of the ramp-shaped waveform applied to the cathode of D82 can be changed. Thus, the delay in the operation of Q180 can also be varied. The small resistor, R82, provides a small step at the start of the ramp-shaped waveform which overcomes the bias on D82 and allows Q180 to be triggered with minimum delay when R8 is at the ground end.

Second Regenerator Blocking Oscillator

Operation of Q180 is similar to the operation of Q70 and Q160 and will therefore not be described in detail. The purpose of the Second Regenerator Blocking Oscillator is to produce a constant output-amplitude pulse after a varioble time-delay, each time that the oscilloscope is triggered. The output of the Second Regenerator is applied from T180 to a distributed amplifier stage through R185.

Distributed Amplifier Stage

The output of approximately +10 volts from T180 is applied to the grid line of V184 and V194. Bias for the tubes is obtained from the junction of R187 and R188 and is applied through R185 to the grid line.

The positive signal at the grids of VI84 and VI94 produces a negative signal at the plates. The negative signal is applied through CI90 and CI96 to the plate line of the Distributed Amplifier. The main purpose of the amplifier is to develop the large amount of power required to drive the Time-Base Cate and Unblanking circuits. The distributed amplifier circuit is used to minimize the deterioration of the risetime of the output from the Second Regenerator Blocking Oscillator.

When the positive pulse from T180 is applied, the tubes are brought into very heavy conduction, with a total peak current of approximately 0.5 ampere. The peak current, however, is only demanded for approximately 25 nanoseconds out of each 2.5 microseconds or more, resulting in an average current of approximately 5 ma. The large amount of peak current develops approximately a 40-volt pulse across the 100-ohm load at the output (R199) and the 220-ohm reverse termination (R190).

The pulse developed in the plate line of V184 and V194 is negative. However, the Time-Base Generator and Unblanking circuits require a positive drive pulse. Transformer T198 is used to produce the voltage investion. The transformer uses a twin lead transmission line passed seven times through a ferrite core. Energy is coupled into one of the twin leads at the input side and taken from the other lead at the output side with the opposite polarity. The ferrite core is used to extend the low-frequency response of the transformer. This transformer is similar to the one described previously which inverts triggering signals at the TRIGGER SUURCE switch.

The output from TIP8 is a positive pulse of approximately 40 volts. The pulse is passed through DIP3 and DIP8 into a length of coaxial cable, where it is transmitted into the Ubelanking and Time-Base Gate Generator circuits. After a certain period determined by the setting of the NANO-SEC/CM switch (and for only the three fastert sweeps) the pulse is reflected back to the junction of DIP3, DIP8, and DIP9, but in the opposite polarity. The negative reflected pulse causes D199 to conduct allowing the reflected pulse to be terminated and absorbed by R199. This prevents unwanted multiple reflections.

UNBLANKING AND TIME-BASE GATE

Clipping Line

As shown in the block diagram of Fig. 4-8, the positive pulse from T198 is applied through a lumped-constant delay line into the clipping line. The clipping line is used to determine the length of the time-base gate in the 2, 5, and 10 positions of the NANOSEC/CM switch and is composed of 3 sections of coaxial cable. In these positions, the positive pulse travels down the clipping line until it reaches the short provided by a switch contact. The shorted output causes a 100% reflection of the pulse and a polarity reversal. The reflected pulse then turns off the positive gate appearing at the grid of V244. The duration of the positive pulse appearing at the grid of V244 depends on the time required for the pulse to travel to the end of the clipping line, reflect from the shorted end, and return to the arid of V244. This is twice the transit time of the length of cable used. The NANOSEC/CM switch changes the length of coaxial cable used to produce the reflection and thus determines the length of the gate produced.

The maximum gate length which can be produced by means of the clipping line is limited by two primary factors. The first limitation is that the gate cannot last longer than the duration of the positive pulse obtained from T198. The second limitation is imposed by the physical need for additional lengths of cable for the clipping lines as the length of the gate is increased. The long cables required for long gates would make this means of setting the gate lengths impactical.

In all positions of the NANOSEC/CM control except 2, 5, and 10, R250 terminates the clipping line and thereby prevents reflections. In all of these slower sweep-rate positions, the Gate Extender Multivibrator determines the final length of the time-base gate.

Gate Extender

In the 2, 5, and 10 positions of the NANOSEC/CM switch, plate vallage is disconnected from V264, V3938, and V383, thereby disabling the Gate Extender Circuit. In these positions, the length of the gate is determined solely by the clipping line length. In all other positions of the NANOSEC/CM switch, plate voltage is applied to the tubes. Fig. 49 is a simplified diagram of the Gate Extender Circuit.

A voltage divider between -250 volts and ground consisting of R257, R255, and R250 holds V264 in cut off by maintaining the grid a few volts negative. With V264 cut off, R262 and R263 set the plate voltage at approximately +100 volts. Tube V274 is normally conducting heavily. Grid current keeps the grid or approximately ground potential. Capacitor C262 is therefore charged to approximately 100 volts.

When a positive pulse is applied to the grid of V264, the tube conducts, causing the plate voltage to drop. The



Fig. 4-8. Unblanking and Time-Base Gate Circuit block diagram.

drop in plate voltage causes VZ74 to cut off. When C262 hos discharged through R271 and R270, VZ74 again conducts and ends the time-base gate. The time that VZ74 remains cut off is controlled by the setting of C262 and by the value of R270 selected by the NANOSEC/CM switch. At the 1000 setting of the NANOSEC/CM switch the duration of the gate is approximately 9 microseconds.

During the time that V274 is cut off, its plate voltage is near +100 volts. The positive rise in voltage is divided across R200 and R281 and applied to the grid of V3938. The purpose of the divider is to obtain the correct dc level for the grid of V3938. The output from cathode follower V3938 is then applied to the grids of both sections of V283. The resultant rise of approximately 20 to 30 volts at the cathodes of V283 courses the grid of V244 to go positive by a similar amount. The voltage obtained from V283 extends the pulse obtained from T198 for the time interval required for a particular sweep duration. During the time that the pulse from T198 is present, the positive voltage at the cathodes of V283 holds V283 in cutoff.

The resulting positive gate is applied to the Sweep-Gate Amplifier, V244, to V264, and to the Delayed-Gate and Unblanking Amplifier, V214. The positive signal at V264 regenerates the input signal after expiration of the drive pulse and causes the Gate Extender to operate as a oneshot multivitartor.

Gate Amplifier

V244 is a high GM pentode which is used to obtain the high current necessary to drive the Time-Base Generator. Both the cathode and dynode voltages are regulated. The cathode voltage of both V214 and V244 is set at approxi-



Fig. 4-9. Gate Extender Circuit simplified circuit diagram.

mately +7.5 volts by emitter follower Q238. The positive voltage on the cathodes holds both V214 and V244 in cutoff while permitting the grids to be near ground potential.

When a positive pulse is applied to the control grid, V244 conducts heavily causing the plate voltage to drop. The resulting negative pulse at the plate is applied to the Time-Base Generator where it is used to gate on the sweep.

Delayed-Gate and Unblanking Amplifier

With no positive gate applied to the grid of V214, the tube is biased beyond cutoff by Q238, and the plate voltage is approximately +650 volts. When the positive gate is applied to the grid of V214, the tube conduct heaving via the plate voltage drops. Diodes D214 and D215 limits the amount that the plate voltage of V214 can drop. The anode of D214 is set at approximately +540 volts by Zener diode D220. Thus, when the cathode voltage of D215 drops to approximately +540 volts, the diodes conduct, preventing of ruther decrease in voltage. The limiting action of D214 and D215 permits the plate voltage of V214 to decrease by only about 110 volts.

The drop in plate voltage of V214 is applied through C228 and other components to the cathode of the CRT. The negative pulse at the cathode unblanks the CRT for the duration of the time-base positive gate at the grid of V214.

The positive pulse at the base of Q214 saturates both Q214 and Q215 emitter-coupled switching network providing a +12-volt pulse at the front-panel connector labeled DELAYED +GATE 50 $\Omega.$

TIME-BASE GENERATOR

Block Diagram

A block diagram of the Time-Base Generator is shown in Fig. 4-10. Two conditions are automatically maintained between sweeps in order for the generator to function properly. The first condition is that the plate voltage of V331 is maintained at +155 volts to insure that the sweep sawtooth waveform starts from the same voltage for each



Fig. 4-10. Time-Base Generator block diagram.

sweep. The second condition is that the control grid voltage of V331 is maintained at approximately -3.2 volts. Two feedback loops are used to provide these starting conditions. The two loops are shown on the block diagram.

The first feedback loop operates if the plate voltage of V331 deviates from +155 volts between sveyers. When this occurs, an error signal is generated in the Plate-Voltage Regulator Circuit (fast loop) which is amplified and fad through the Disconnect Diades to the control grid of V331. The error signal causes V331 to change its operating point to bring the plate voltage back to +155 volts. However, the signal at the grid voltage remain to -23 volts and thus causes the grid voltage errons int -32 volts and thus causes the Screen-Grid Suppl volta plate the screen ford Suppl voltage is not at -32 volts of V331 against a reference. If the control grid voltage of V331 against a reference. If the control grid voltage of v331 much a manner as to cause the fast loop to move the control grid to -32 volts.

When a negative gate is applied to the Sweep Generator, both feedback loops are temporarily disabled and V331 is cut off. The sawtooth waveform is then generated in the plate circuit of V331 and applied through the Output Cathode Follower and the Paraphase Amplifier to the deflection plates of the CRT. The sawtooth waveform at the CRT moves the electron beam horizontally across the screen to form the sweep. A bootstrap cathode-follower circuit is used to improve the linearity of the basic sweep generator by driving the more-positive and of the timing resistors.

Plate-Voltage Regulator

After each sweep is produced, the plate voltage of V331is returned to +155 volts by the fast regulator loop. Any trendency for this voltage to shift between sweeps would result in a horizontal movement of the start of the sweep on the screen.

The plote voltage of V331 is applied directly to the grids of output cathode follower V343. The level of the voltage at the cathodes of V343 is reduced approximately. 250 volts by Zener diodes D344 and D345 and 2pplied through R344 to the control grids of V363. The cathode voltage of V336 is in twa rapplied through R375 and L375 to the cathode of V3748 where it is compared with the voltage on the control grid obtained from R374. If the plate voltage of V331 is higher than normal, for example, the cathode voltage of V3748 ulb emore positive than normal. This reduces conduction through V3748, causing the plate current to decrease at R373. The resulting rise in voltage is applied to the base of Q318 causing the current through Q318 and Q328 to decrease.

The current through disconnect diodes V312 and V322 is set by transitors 0.328 and Q318. Therefore, as the current through the transitors is reduced, so also is the current through the disconnect diodes. The voltage drop across R380 is therefore reduced, allowing the control grid of V331 to move in the positive d'rection. This permits V331 to conduct more heavily, lowering the plate voltage to the proper level. A decrease in the plate voltage of V331 is compensated for similarly. The grid voltage of V331 would be forced more negative by the feedback loop to decrease conduction and thereby allow the plate voltage to rise to the proper level.

Screen-Grid Supply

As a result of action by the Plate-Voltage Regulator loop, the control grid of V331 may tend to not be at the normal -32 volts. This is corrected by the slow loop as follows: The voltage at the grid of V331 is also applied to the grid of V3934, where it is compared with the voltage at the grid of V394. This produces an increase in conduction through V394. This produces a drop in the plate voltage of the grid of V394. This produces a drop in the plate voltage of the grid of V403 causing a decrease in the screen-grid voltage of V403.

When the screen-grid voltage of V331 is decreased, conduction through V331 also decreases, causing the plate voltage to rise. The Plate-Voltage Regulator loop makes the control grid of V331 more positive to increase the current and return the plate voltage to normal. Thus, the change in screen-grid voltage produced by the Screen-Grid Supply causes the Plate-Voltage Regulator loop to return the control grid to -3.2 volts. The reference voltage which the control-grid voltage of V331 is compared against can be adjusted by means of R3% to obtain the correct regulated control-grid voltage.

Capacitor C333C in the cathode of V403 prevents the Screen-Grid Supply from operating as fast as the Plate-Voltage Regulator loop. The main function of the Screen-Grid Supply is to readjust screen voltage as the Sweep Range switch is operated and as V331 eventually ages or is replaced.

Sawtooth Generator

When a negative gate is produced by the Time-Base Gate circuit, the amplified gate is applied through Zener diades D205, D306, D307 and D308 and developed across R380. This negative gate is approximately 45 volts in amplitude. The four Zener diades decrease the dc level of the gate without affecting its amplitude. The negative gate cuts off V312 and V322 and drives V393 A toward cutoff, thereby disabiling the feedback loops. The negative gate also cuts off V313 to initiate the generation of the sawtooth waveform. A simplified diagram of the Time-Base Generator is contained in Fig. 4-11.



Fig. 4-11. Time-Base Generator simplified circuit diagram.

Approximately 16 pf total capacitance to around appears in the plate circuit of V331. When the tube cuts off, the capacitance starts to charge toward approximately +475 volts through R336. The rate at which the stray capacitance charges is determined by the particular value of R336 selected by the NANOSEC/CM switch and by the exact voltage obtained from the +475-volt supply. The rate at which the stray capacitance charges determines the rate of rise of the sawtooth waveform produced across the stray capacitance and ultimately the sweep rate. The voltage obtained from the +475 volt supply is adjustable by means of a separate potentiometer for each range of the NANOSEC/CM switch. The potentiometers are adjusted to provide the necessary voltage to obtain correct sweep timing on their respective sweep ranges (refer to the Power Supply Diggram.)

The sawtooth waveform produced by the charging of the stray capacitance is applied to the grids of the Output Cathode Follower, V343. At the cathode of V343 the dc level of the sawtooth waveform is decreased by about 250 volts by diodes D344 and D345 while the amplitude of the sawtooth is unchanged. The sawtooth waveform is then applied to the grids of V353, the Bootstrap Cathode Follower. The sawtooth waveform appearing at the cathodes of V353 is applied through C356 and C357 to the cathode of V332, thereby cutting V332 off. Since the voltage at the cathode of V332 rises at approximately the same rate as the voltage at the plate of V331, the voltage across the timing resistor, R336, remains approximately constant. With the voltage across the timing resistor constant, the current charging the stray capacitance is also constant. The Bootstrap Cathode Follower thus causes the capacitance to charge linearly rather than in the normal exponential manner. The linear charging of the capacitance produces a linear output sawtooth waveform. The output sawtooth waveform is applied from the junction of D345 and R344 to the grids of V424.

Tube V388 is used to rapidly discharge the capacitance in the cathode circuit of V343 ofter each sweep. This is important in order to return the grids of V353 to normal as rapidly as possible to allow C356 and C357 to discharge in time to be rady for the next sweep. Normally V388 is cut off by the bias voltage obtained from the junction of R381 and R383. When the negative gate is applied to start the sweep, the signal passing through C380 is clamped by D384, charging C380.

At the end of the sweep gate, the positive rise in voltage is coupled to the grid of V388 through C380. This causes V388 to go into conduction to supply the current necessary to discharge the capacitance in the cathode circuit of V343. The current available to discharge the capacitance causes the cathode voltage of V344 to fall napidly, as is required.

Paraphase Amplifier

The output savicatin waveform is applied to the grids of the Paraphase Amplifier, V424. A positive-going savicatin is obtained from the cathode and a negative-going savicatin is obtained from the plate. The amplitude of both savitooths is 150 volts or more. Zener diodes D430 and D431 are used to decrease the de level of the savication waveform obtained from the plate to the same average voltage as the savication obtained from the cathode. The negative going waveform is applied to the left hand deflection plate of the CRT while the positive going waveform is applied to the right hand deflection plate. Positioning vollages are applied to the deflection plates from R441 through isolation resistors R440 and R442.

The Paraphase Amplifier is designed to reset somewhat slowly. If it ware to reset radjidly, the sweep retrace would be visible on the screen unless the CRT is very radjidly blonked. In order to produce this radjid blanking, complex blonking circuitry would be required. Therefore to keep the blonking circuitry as simple as possible, the beam is held at the right side of the screen by V424 until the blanking circuit has a chance to blank the beam. V424 then slowly resets. Even on the fastest sweeps V424 has much more than sufficient time for reset.

HIGH-VOLTAGE POWER SUPPLY AND CRT CIRCUITS

High-Voltage Power Supply

The High-Voltage Oscillator, V800, operates at approximately 20 k due to the resonant circuit of C808 and the high-voltage transformer. The primary voltage of T801 is stepped up at the secondary and applied to the high voltage rectifiers. A voltage tripler circuit made up of V802, V812, and V822 produces approximately -20k which is applied to the post-accelerating terminal of the CRT. The cathode of V823 is connected to a tap on the secondary of T801. The rectified voltage at the plate of V832 is approximately -42 kv.

A sample of the voltage obtained from the plate of V823 is applied from a voltage divider network to the grid of V8148. If the supply voltage changes from the preset level, a voltage change appears at the grid of V8148. The voltage change is amplified by V8148 and V814A and causes a change in the screen voltage of V800 which adjusts the amplitude of the ascillations to compensate for the change in high voltage. The high voltage is set by adjusting R81, which controls the voltage at the arid of V8148.

CRT Circuits

A voltage divider between the plate of V832 and ground is used to obtain operating potentials for the acthode, control grid, and focus grid of the CRT. Four neon bulbs, 8853, 8854, 8855, and 8856, are used to regulate the cathode, control grid, and focus grid voltages to prevent changes in cathode current from affecting the operating voltages of the CRT.

The control grid voltage for the CRT is obtained from R856 which is at a more negative point than the point where the cathode voltage is obtained. The negative voltage on the grid normally holds the CRT cort off until the unblanking circuit. The INTENSTY control, however, has sufficient range to overcome the cut off bias in the absence of an unblanking puble.

Circuit Description—Type 519

Vertical positioning voltages are obtained from R655 and applied through R867 to R868 and the bypassed lower vertical deflection plate of the CRT. Other operating voltages for the CRT are also obtained from simple voltage divider networks.

LOW-VOLTAGE POWER SUPPLIES

Primary Power

Line voltage is applied through F601 and SW601 to the primary windings of T601. Two primary windings are used. The two windings are connected in parallel for 117-volt operation and in series for 234-volt operation. The blower is connected across one of the primary windings and operates whenever the POWER switch is closed.

Voltage applied at the primary winding of T601 eneraizes the secondary windings of the transformer. Voltage obtained from terminals 18, 19, 22 and 23 is rectified by D650 and D651 and applied to energize the regulated heater supply. Negative 26.5 volts from the regulated heater supply is applied to the time delay relay K601. After approximately a 45-second delay, the contacts of K601 close, thereby energizing K602. The contacts of K602 then energize K603. The purpose of the time delay relay is to delay application of power supply voltages to oscilloscope circuits until the tube filaments have had a chance to heat. Relays K602 and K603 control the application of the power supply voltages to the other circuits of the oscilloscope and to all Power-Supply regulator circuits except the -250-volt and -26.5-volt regulators. If the temperature inside the instrument becomes excessively high, the contacts of the thermal cutout, TK602, open to de-energize relays K602 and K603 and thereby remove power from the oscilloscope circuits.

-250-Volt Power Supply

Voltage obtained from terminals 6 and 11 of T601 is applied through a full-wave bridge rectifier circuit and filter to the regulator circuit of the -250-Volt Power Supply. By varying the drop in voltage across V627A, the negative output voltage from the power supply can be controlled. The greater the voltage dropped across V627A, the less the output voltage for the power supply.

A voltage divider consisting of 6K44, fs47, and K648 is used to set the voltage on one of the grids of V646. The voltage on one of the grids of V646 the voltage on the other grid is determined by voltage regulator tube V639 and fs433. The voltage across V639 holds constant at about 85 volts by gas-tube regulator action. This places pin 7 of V646 normally at -165 volts. The voltage across V639 holds out of the power supply.

When the output voltage has been set, any change in that voltage produces a voltage change at the grids of V646 which constitutes an error signal. If the output voltage of the power supply becomes less negative than normal, as for additional load, the grid voltages of V646 obs start to rise. Due to the constant voltage drop across V639 the change in voltage at pin 7 is approximately 3 times greater than the change at pin 2. Consequently the net change in the voltage between the grids of V464 is approximately 2/3 the change in the output voltage of the power supply. The larger change at pin 7 causes that section to conduct more heavily. The amplified signal at pin 6 is then opplied to the grid of V624 as a negativegoing signal. This causes V624 to conduct less and the further amplified array and the plate of V624 is applied to the grid of V627A as a positive-going signal. The positive signal on the grid of V627A causes that tube to conduct more heavily, thereby reducing the voltage drop across the tube and increasing the output voltage of the power supply to the normal level.

A tendency for the power supply output to increase is eliminated in the same manner except that the polarity of all the signals is reversed from the previous example. Thus the regulator compensates for either an increase or a decrease in the output of the power supply to insure that the voltage remains at the normal level.

The -250-Volt Regulator operates at all times, and is not dependent on the operation of K601, K602, and K603.

Regulated Heater Supply

Voltage for the Regulated Heater Supply is obtained from terminals 18, 19, 22, and 23 of T601. Negative 35 volts is obtained from D650 and D651 and applied to the collector of the series regulator, Q777. A voltage divider between —250 volts and ground sets the reference voltage on the base of Q764A. The voltage on the base of Q764B is set by a voltage divider between —265 volts and grown. Because of the common emitter resistor for Q764A and Q764B, the voltage at the base of Q7664B is compared against the voltage at the base of Q7668 is base of Q766B. This error signal is produced at the base of Q766B. This error signal is produced at the base of Q766B. This error signal is applied though and papiled through Q773 to the base of Q777 in the direction which returns the output voltage to normal.

A voltage divider between -26.5 volts and ground sets the base of Q767 at approximately -13 volts. The emitter of Q767 stabilizes the -12.6 volt output from the supply.

Other Regulator Circuits

The other regulator circuits are similar in operation to the -250-Volt Regulator and the -265-Volt Regulator (Regulated Heater Supply). Each of the regulators uses the -220-Volt Supply as part of its reference. Voltages obtained from the output of each supply are compared to a reference voltage. An error signal is produced when the output voltage deviates from its normal value. The error signal is then amplified and used to control the operation of a series regulator twb. The polarity of the error signal applied to the series regulator is always such as to bring the output voltage back to normal.

The +475-Volt Supply is somewhat different from the other circuits in that several potentiometers are used to adjust the output of the supply. A separate potentiometer is provided for each position of the NANOSEC/CM switch. The voltage at the grid of V724 can be adjusted by means of the particular R323 selected by the NANOSEC/CM switch. By setting the output-voltage divider ratio, R732 determines the output voltage of the power supply. The output voltage is adjusted on each setting of the NANO-SEC/CM control by means of the respective R732 for correct sweep timing on that range.

CALIBRATION-STEP GENERATOR

Reed Switch and Charge Line

A mechanical dry-read switch and associated charge line are used to produce the output steps from the Calibration-Step Generator. The charge line is a polystyrene-loam insulated, rigid coaxial line with a characteristic impedance of 125 ohms and a one-way transit time of 1.5 nanoseconds. The read switch is located inside the charge line and forms part of the 125-ohm system.

Charging voltage for the charge line is obtained from voltage divider networks and opplied through a special charging network to the charge line. The POLARITY switch determines the polarity of the charging voltage and the RANGE switch determines the magnitude of the voltage applied to the VOLTS control, R879. In the 1 V and 10 V positions of the RANGE switch, the charging voltage obtained from R879 is applied through R882 and R883 to the charge line. In the VARIABLE position of the RANGE switch the charging voltage is obtained from the VARIABLE potentiometer, R875. The amplitude of the output steps from the Calibration-Step Generator into a 125-chm load is exactly half of the charging voltage.

As the dry metal read switch starts to class, the capacitance of its contacts increases considerably. The added capacitance at the contacts draws charge out of the charge line. This would have a tendency to affect the amplitude of the output pulse unless some provision is made to reduce this effect. Copacitor C&BS prevents the charge drawn from the line from affecting the amplitude by any significant amount.

When the switch contacts close, the charge line acts as a 125-ohm source supplying energy to a 125-ohm load. As a result only half of the charging voltage appears across the output 125-ohm load. At the same time that the contracts close, a backware is propagated down the charge line toward the charging network. Resistor R883 is used to terminate the backware and thus eliminate a reflection from the charging network. Resistor R883 is used to terminate the backware and thus eliminate a reflection from the charging network. Resistor R883 is used to terminate the backware and thus eliminate a reflection from the charging network the tortarist time of the charge line, after which the output applivade remains relatively constant for twice the transit time of the charge line, after which the output endplivade decays as C883 discharges. When the switch contacts again open, the charge network to the initial voltage before the contacts close again.

The charging network can be disconnected from a fixed charge line in the instrument. This permits additional lengths of charge line and other devices to be added to the internal charge line. The charging network is then connected to the end of the additional charge line.

A 125-ohm trigger takeoff is used in the charge line to obtain a triggering signal for the oscilloscope sweep. The trigger takeoff is identical in function to the trigger takeoff at the vertical input of the oscilloscope. The takeoff obtains a suitable triggering signal from the backwave propagated down the charge line.

The dry metal read is used in the Calibration-Step Generator because of the relative ease of making the switch appear to be part of a 125-ohm transmission line. It is important that the switch be part of the 125-ohm system in order to prevent reflections and mismatches occuring at the switch. Use of a mercury switch would further complicate the design of the transmission system. The dry read is selected because for its excellent waveform. The lifetime of the switch is frequently quite short because of the severe requirements of the application.

Oscillator

An oscillator is used to drive the reed switch and thereby produce the output steps. The oscillator is composed of tubes V885 and V895A.

The basic circuit is that of a Wien-Bridge Oscillator, as can be seen from the simplified diagram in Fig. 4-12. Two stages of triade amplification provide gain necessary to maintain oscillations. The amplitude of the oscillations is very high, resulting in highly clipped waveforms. Adjustment of the FREQUENCY control simultaneously varies the settings of R820A and R827B. This shifts the frequency of which the peak positive feedback occurs and thus changes the frequency of the oscillator.



Fig. 4-12. Simplified circuit diagram of the Calibration-Step Generator oscillator.

The output of the oscillator is taken from the plate of pentode V885. Plate current is divided through R885 and 1885. The amount of current passing through L885 is determined by the setting of R885. The field set up by L885 actuates the read at the frequency of the oscillator.

In normal operation, the frequency of the oscillator is adjusted near the mechanical resonant frequency of the read switch (approximately 700 to 800 cycles). Operation near the natural resonant frequency of the read switch reduces the tendency for the contacts to bounce and reduces the amount of drive which must be provided by L885.

When the DRIVE control is rotated fully counterclockwise to the SINGLE CLOSURE position, SW885 closes sending current from R886 through L885. The current through the coil causes the reed to close once.

RATE GENERATOR

Multivibrator

Tubes V915 and V895B form a free-running multivibrator circuit which is used to control the repetition rate of the Rate Generator. For the purpose of explaining the operation of this circuit, we will assume that V915A conducts as power is applied to the circuit and that the MULTIPLIER switch is set to X1. This connects C920A between the cathode of V915B and the grid of V895B and connects C920B between the plate of V895B and the grid of V915A. As V915A conducts, its plate voltage drops causing a decrease in voltage at the cathode of V915B. The drop in voltage is coupled through C920A to the grid of V895B, cutting the tube off. The resulting increase in the plate voltage at V895B is coupled to the grid of V915A through C920B driving V915A farther into conduction. Capacitor C920A then discharges through R925 and R923 permitting the grid of V895B to slowly rise. When the charge on C920A becomes insufficient to hold V895B cut off, V895B goes into conduction. The time that V895B remains cut off is determined by the values of C920A, R925, and R923 and by the voltages obtained from R923.

When V895B conducts, the drop in its plate voltage is coupled to the grid of V915A, causing V915A to cut off. This causes the plate of VPISA and the cathode of VPISB to rise, coupling a positive signal to the grid of VPSPB to drive VPSPB farther into conduction. The time that VPISA is held cut off is determined by the time required for C220B to discharge through PP28, R929 and R911. As C920B discharges, the grid of VPISA rises and eventually permits the tube to come back into conduction. When VPISA conducts again, another cycle of operation is initiated.

The frequency of the multivibrator operation is determined by the values of C202 selected by the MULTIPLER witch and by the setting of the CYCLES/SEC control. The values of C202 affect the period of time that both V955A and V959S remain cut off. The setting of R223 affects the time that V959S remains cut off and controls the output wing of V915B, reducing the amplitude of drive to the grid of V959B at the clockwise end of the CYCLES/SEC control.

Cathode follower V915B also serves to couple the output of the multivibrator to the avalanche circuit through C930.

Avalanche Circuit

The output pulses from the multivibrator circuit are oppiled through (293) to the collector circuit of Q934. The AVALANCHE SET control R931 adjusts the collector voltage of Q934 to the point where the transistor is just short of avalanching. When a positive pulse is applied from the multivibrator as V915A cuts off, the additional voltage is enough to cause Q934 to avalanche, thereby discharging the fixed charge line in the collector circuit. The duration of the outpulse is twice the transit time of the charge line, or approximately 10 nanoseconds for the charge line used with the instrument.

The output from Q924 is taken from the emitter circuit. The components added to the emitter circuit are used to fully terminate the discharge of the charge line and to shope the output waveform. The circuit can be used to drive a 50-ohm load directly from the +RATE 50 Ω connector on the front panel.

A sample of the output pulse is developed through R939 and applied to the triggering circuit of the oscilloscope. This permits the oscilloscope to be triggered easily from the rate generator without any external patching of the signals.



SECTION 5

MAINTENANCE

PREVENTIVE MAINTENANCE

Air Filter

The Type 519 Oscilloscope is cooled by air drawn through a washable filter located at the rear of the instrument. The filter is constructed of adhesive-coated aluminum wool. If the filter become secssively diry, it will restrict the flow of air into the instrument and may cause overheating. High internal temperatures will not only reduce the lifetime of the instrument components but may also cause the thermal cutout to open at a crucial point in an experiment. Any time that the thermal cutout opens, the filter should be checked immediately.

The filter should be visually checked every few weeks. It should be cleaned at least every three or four months, more often if required. To clean the filter, first remove the loose dirt by tapping the filter genity on a hard surface. Then wash the filter by running hat soapy water through it until it is clean. After rinsing and allowing it to dry, cost the filter with an adhesive such as "Handi-Kater" or "Filtercost" (products of the Research Products Corporation). These products are generally available from air-conditioner suppliers.

Cleaning the Exterior

Losse dust accumulating on the auside of the Type 519 can be removed with a lintfree cloth or a small point brush. The paint brush is particularly useful for dislodging dust an and around the front-panel controls. A soft cloth dampened with water and a small amount of liquid detergent can be used to remove the harder coating of dirt. Abrasive cleaness should not be used.

To clean the graticule and the face of the CRT, first remove the four slotted graticule nuts and remove the cover and mask assembly. Then unscrew the knutled knob used to position the graticule and remove the graticule from the mask assembly. Clean the graticule and the face of the CRT with a soft, lint-free cloth dampened with denatured alcohol.

The graticule cover and mask assembly are remounted by reversing the order of their removal.

Removal of Panels

The side, top, and bottom panels of the Type 519 can be removed separately for maintenance work. The side panels are held in place by small screwhead fasteners. To remove the side panels, use a screwdriver or coin to rotate the fasteners approximately two turns counterclockwise. Then pull the upper portion of the panels outward from the carrying handles. To pand bottom panels are held in place by small screws. After first removing the screws, the top and bottom panels can be lifted off.

Cleaning the Interior

Although air entering the Type 519 is filtered, some dust may penetrate into the interior of the instrument. This dust should be removed occasionally to prevent instrument failures due to the conductivity of the dust under high humidity conditions. Perhaps the best way to keep the interior of the instrument (ean is to blow dust off using compressed air. A very high velocity air stream should be avoided, however, to prevent damage to some of the components. Persistent dir can be removed using a damp cloth or a small paint brush.

Special attention should be given to the high-voltage circuits, including parts inside the high voltage box. Since most of the high-voltage parts are enclosed in a transpoent plantic box, very little dust should accumulate on these parts. If dust does accumulate it should be removed, since excessive dust combined with high humidity can produce arcing and possible high-voltage failures. Presence of arcing will normally cause false triggering of the instrument, particularly for X20 trigger agin.

A cloth dampened in denotured alcohol may be used to clean dirt and dust off the high voltage anode lead and around the anode of the CRT. A cotton-tipped applicator can be used for cleaning in narrow spaces and for cleaning caramic strips.

Visual Inspection

Many potential and existant troubles can be detected by a visual inspection of the instrument. For this reason, you should perform a complete visual check every time the instrument is calibrated or repaired. Visual checks should also be made during other routine maintenance work.

Defects which may be detected visually include such things as loose or broken connections, loose set screws in the knobs or shaft couplers, loose or damaged coaxial connectors, improperly seated tubes or transistors, scorched or burned parts, and broken terminal strips. The remedy for most of these troubles is a opparent. However, particular care must be taken when heat-damaged components are detected. Overheating of parts is often the result of other, less apparent, defects in the circuit. It is essential that you determine the cause of overheating before replacing heatdamaged parts in order to prevent further damage.

Blower Motor

The blower motor bearings are factory lubricated and sealed. No additional lubrication is required for the life of the instrument.

Tube Checks

Periodic tube checks on the tubes used in the Type 519 Oscilloscope are not recommended. Tube testers in many cases indicate a tube to be defective when that tube is operating quite satisfactory in a circuit, and fail to indicate tube defects which affect the performance of the circuits. The ultimate criterion of the usability of a tube is whether or nor the tube works properly in the circuit. If it does not, then it should be replaced. If it is working correctly, it should not be replaced. Unterestary replacement of tubes is not only expensive but may also result in needless recalibration of the instrument.

Recalibration

The Type 519 Oscillacope is a stable instrument that will provide many hours of trouble-free operation. However, to insure the reliability of measurements we suggest that you recalibrate the instrument offer each 500 hours of operation (or every six months it used intermittent). A complete step-by-step calibration procedure is given in the Calibration Procedure section of this manual.

REMOVAL AND REPLACEMENT OF PARTS

General Information

Most parts in the Type 519 Oscilloscope can be replaced without detailed instructions. Other parts, howere, can best be removed if a definite procedure is followed. Instructtions for the removal of some of these parts are contained in the following paragraphs. Because of the nature of the instrument, replacement of certain parts will require that you recalibrate portions of the scaliloscope to insure proper operation. Refer to the Calibration Procedure portion of this manual for the applicable calibration steps.

Removal of Cathode-Ray Tube

If it becomes necessary to replace the CRT, the CRT and shield should be removed as a single unit. To remove the unit, first disconnect all leads to the CRT. Disconnect the input coaxial cable and termination from the cRT base by board. Disconnect the CRT socket from the CRT base by pressing back on the plastic flanges attached to each side of the socket. The sockets can be worked loase by pressing first on one flange and then the other until the socket is free of the CRT. Remove the four slotted graticule nuts which hold the graticule cover and mask. Disconnect the rear supports of the CRT shield and gently remove the complete CRT assembly through the fron-panel opening of the accilloscope, taking extreme core of exposed hardware. The CRT can then be removed from the shield only if absolutely necessary. Separation from the shield is usually undesirable due to (1) cost of the tube and (2) possibility of damage to exposed tube pins and resultant loss of vaccum. CRT's are shipped from the factory with a shield already installed.

Installation of CRT

With the CRT and shield reinserted and the shield bolted into the instrument, small leads can be reconnected by following the color code information printed on the shield. Then connect the anode lead, coaxial cable, and termination. Replace the mask in the araticule cover assembly making certain the serial numbers of the CRT and mask agree. After replacing the CRT and mask, recheck the vertical sensitivity to determine if this measurement agrees with the figure on the mask. A slight adjustment in high voltage may be necessary to correct the deflection factor: however, any future adjustment of either the -250-volt supply or high-voltage supply will affect the vertical sensitivity. Adjust the AXIS ROTATION control to align the trace with the horizontal araticule markings and check the calibration of the sweeps with the 1 KMC Timing Standard. Detailed instructions for completing these steps can be found in the Calibration Procedure section of this manual.

Replacement of Switches

Methods for removal of defective switches are, for the most part, obvious, and only a normal amount of care is required. Single wafers are normally not replaced in the switches used in the Type 519. If one wafer is defective, the entire switch should be replaced. Switches can be ordered from Tektronik, either wired or unwired, as desired.

Because of the complexity of the NANOSEC/CM switch, some special core is required to remove this witch. First loosen the shaft coupling just in front of the forward support for the switch. Then slide the coupling down the shaft. Disconnet the two metal supports for the switch from the chossis. Unsolder all leads coming to the switch while making a careful drawing of the lead connections. Remove the lock nut which holds the switch in the front support, and slide the switch out of the support. It should then be possible to remove the besidth for the instrument.

The NANOSEC/CM switch should be replaced by reversing the order of the steps required for removal. The wiring diagram made during removal can be used to wire in the new switch.

Tube Replacements

Care should be taken both in preventive and corrective maintenance that tubes are not replaced unless they are

actually causing trouble. Many times during routine maintenance it will be necessary for you to remove tubes from their sockets. It is important that these tubes be returned to the same sockets unless they are actually defective. Unnecessary replacement or switching of tubes will often necessitate recalibration of the instrument. If tubes do require replacement, it is recommended that they be replaced by previously checked high-quelity tubes.

To replace V331, first loosen the screws which hold the anode strap of the tube in place. Then rotate the anode strap out of the way. The tube can then be pulled from the socket.

Replacing the Reed Switch

The reed switch in the Calibration-Step Generator can be replaced as follows:

Remove the four mounting screws which hold the QUT. PUT 125 Ω connector to the front panel. Grass the connector and pull straight out from the front panel to extract the attached coax assembly. The complete assembly is about 18rd long and contains the read switch (see Fig. S-1). Disassemble the OUTPUT 125 Ω connector to expose the read switch. Grip the metal end of the read switch with long-noscd piters and pull the read switch out.

To install the replacement reed switch, remove the 125ohm connector double button assembly from the connector sleeve. Insert one end of the reed switch into the button assembly and use the button assembly as o holder. Plug the other end of the reed switch into the center conductor located inside the tube. Lock through the small hole in the side of the tube to see that the switch and conductor are properly connected. Allian the slots in the connector



Fig. 5-1. Exploded view of the coax tube assembly. The Snap Ring, Coupling Nut, and Retaining Nut need not be removed from the tube when the reed switch is replaced.

parts and reassemble the connector. Insert the coax tube assembly through the front panel while holding the OUT-PUT 125 Ω connector so that the flanges are in the same position as the EXTERNAL TRIGGER 1250 connector. This insures that the connector at the rear of the coax tube assembly will mate properly with the connector at the rear of the instrument. Mount the OUTPUT 125 Ω connector securely to the front panel with the four screws.

Soldering Precautions

In the production of Tektronix instruments a special silver-bearing solder is used to establish a bond to the caramic terminal strips. This bond may be broken by repeated use of ardinary tin-lead solder, or by excessive heating of the terminal strip with a soldering iron. Occasional use of ordinary 50-50 solder will not break the bond unless excessive heat is applied.

If you frequently perform work on Tektronix instruments, it is advisable thet you have a stock of solder containing about 3% silver. This type of solder is used quite often in printed circuitry and is generally available locally. It may also be purchased directly from Tektronix in one-pound rolls, order by part number 251-514.

Because of the shape of the terminals of the ceramic terminal strips, you may wish to use a wedge-shaped tip on your soldering iron. These tips allow you to apply heat directly to the solder in the terminals and reduces the amount of heat required. It is important to use as little heat as possible while producing a full-flow joint.

Due to the high-frequency requirements of the Type 519, many of the components are soldered in place with very short leads. This is necessary to reduce the lead inductance. When these components are replaced, the leads should again be made as short as possible. The proper technique for soldering and unsoldering short-lead components requires: (1) the use of fong-nose pilers to hold the lead securely between the component and the point where the heat is applied, allowing the pilers to serve as a heat sink, (2) the use of a hot iron for a short time; and (3) careful monipulation of the leads to prevent lead breakace.

Ceramic Terminal Strips

Damaged ceramic terminal strips are most easily removed by unsoldering all connections, then using a plastic or hard rubber mallet to knock the yokes out of the chassis. This is can be done by using the mallet to hit the ends of the yokes protruding through the chassis. The strip with the two yokes can then be removed as a unit. The spacers will probably come out with the yokes. If not, the spacers can be pulled out separately of derwards.

Another way of removing the terminal strip is to use diagonal cutters to cut off the side of the yoke holding the strip. This method permits the strip to be removed from a difficult area where the maller cannot be used effectively. The remainder of the yokes and the spacers can be pulled out separately after the removal of the strip. Since a replacement strip is supplied with yokes already altached, the old yokes need not be salvaged. However, the old spacers can be used at least twice before new ones need be ordered.

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When the damaged strip and yoke assembly has been removed, place the spacers into the holes in the chasis. Then set the ends of the yoke pins into the spacers. Then press or top lightly directly above the yokes to drive the yoke pins down through the spacers. Be certain that the yoke pins are driven completely through the spacers. Using a pair of diagonal cutters, cut off the portion of the yoke pin protruding through the spacers. Fig. 5-2 shows how the ceramic strip parts fit logether.



Fig. 5-2. Installation of ceramic terminal strips.

TROUBLESHOOTING

General Troubleshooting Information

This partian of the Instruction Manual includes information that will enable you to more efficiently troubleshoot the Type S19 in the event that a trouble develops. During troubleshooting work, you should correlate information contained in this section with information obtained from other parts of the manual.

When a trouble occurs in the instrument, you should first recheck the settings of all controls to see that they are set properly. Then operate the front panel controls to see what effect, if any, they have on the trouble. The normal or adhormal operation of each control will allow you to firmly establish the trouble symptoms in your mind. (The location of a trouble which occurs only in certain positions of a control can usually be determined immediately from the trouble symptoms.)

After the trouble symptoms are clearly established, look first for the obvious causes of the trouble. Check to see that the pilot light is on; feel for irregularities in the operation of the controls, listen for any unusual sound; see that the tube filaments are lit; visually check the entire instrument. The type of trouble will indicate the checks to make.

In general, a troubleshooting procedure can be thought of as consisting of two parts; circuit isolation and circuit troubleshooting. In many cases, the general procedure outlined will enable you to accomplish the first part of the procedure. You have then only to find the exact cause of the trouble in the isolated circuit. If the above procedure does not enable you to isolate the defective circuit, then additional checks will be required. After the defective circuit will allow you to determine the exact cause of the trouble will allow you to determine the exact cause of the trouble Tables 5.2 and 5.3 can be used to troubleshoot the Type 519. Table 5.2 is used first to isolate certain troubles to a particular circuit or stage. Table 5.3 is then used to locate the trouble in the isolated circuit or stage. References in Table 5.2 direct you to the appropriate step or steps in Table 5.3. It is clearly not practical to include every possible trouble in the troubleshooting tables, and therefore only those troubles most likely to occur are included. Troubles not found in the tables must be located using the general method of first isolating the defective stage and then determining the cause of the trouble within the stage.

Table 5-3 can also be used independently of Table 5-2. The steps in Table 5-3 are arranged so that they can be used to perform a quick check on the operation of each circuit. The table is subdivided into separate sections for the major circuits contained in the instrument so that if you are able to immediately isolate trouble to a circuit, you can proceed directly to the appropriate section of Table 5-3 without first using Table 5-2.

Although the Type 519 is a stable instrument, it is possible for circuits to get out of colibration thereby producing an apparent trouble. Before proceeding with any detailed trouble analysis be sure that the trouble cannot be corrected by means of some adjustment. If there is any doubt, recalibrate the entire suspected circuit using the information in Section 6.

Unusual troubles may occur due to a failure in one of the power supplies. Also, the circuit configurations used in the Type 519 make it possible for an incorrect power-supply voltage to affect one circuit more than others. Consequently, a power supply trouble should be considered as a possibility in virtually any type of failure which may occur in the instrument. If there is any doubt as to whether a power supply may be causing the trouble, the power supply regulated voltages and ripple should be checked before proceeding with the troubleshooting procedure. If the output and ripple voltages of the regulated power supplies are correct, the power supplies can be assumed to be operating correctly.

When trouble has been isolated to a circuit, perform a complete visual check of that circuit. Many troubles can be found most easily by visual means. If a visual check fails to detect the cause of the trouble, check the tubes used in the circuit by substitution. Approximately 90% of the troubles which occur in Tektronk instrument result from tube failures. Be sure to return any tubes found to be good to their original sockets.

Transistor defects usually take the form of the transistor either opening or shorting. In the case of the trigger amplifiers, signal-tracing around the suspected transistors should first be done. The plug-in trigger amplifiers may also be interchanged. A transistor-curve display instrument, such as the Tektronix Type 575, can also be of help in finding abnormal transistor difficulties. However, in power supply circuits such as the -26.5-volt supply, most failures can be located with an ohmmeter. A check for open or shorted transistors can be made using an ohmmeter. Use of the $R \times 1$ scale of the ohmmeter should be avoided, however, because the low resistance in series with the transistor and the voltage source could conceivably cause damage to a good transistor. Checks should be made with the ohmmeter leads connected both ways across the transistor so that the effects of the polarity reversal of the voltage across the transistor can be observed. If there is doubt about whether a transistor is good or not, substitute another transistor for it in the circuit. Be sure first, however, that the voltages and loads on the transistor are normal before making the substitution. If a transistor is substituted without first checking out the circuit, the new transistor may immediately be damaged by some defect in the circuit.

Separate circuit diagrams for each circuit are contained in the back of this manuel. In addition, a block diagram provides an overall picture of instrument operation. The reference designation of each electronic component in the instrument is shown on the circuit diagrams as well as important voltages and waveforms. The following is a list of the reference designations associated with each circuit.

Numbers less

than 100	Trigger Processing Channel
100 numbers	Triggering Circuits
200 numbers	Time-Base Gate and Un- blanking
300 and 400 numbers	Time-Base Generator
600 and 700 numbers	Low-Voltage Power Supplies
800 numbers	CRT Circuit and Calibration- Step Generator
900 numbers	Rate Generator

Switch wafers shown on the circuit diagrams are coded to indicate the position of the wafer on the actual switches. The number portion of the code refers to the wafer number on the switch assembly. Wafers are numbered from the front of the switch to the rear of the wafer is used to perform the particular switching function.

All wiring in the Type 519 is color coded to facilitate circuit tracing. The power-supply buses are identified by the following code; the widest stripe identifies the first color in the code.

+650 v	Blue-green-brown on white
+475 v	Yellow-violet-brown on white
+450 v	Yellow-green-brown on white
+225 v	Red-red-brown on white
+100 v	Brown-black-brown on white
—250 v	
—26.5 v	Red-blue on dark background
—14 v .	Black-yellow on white
—12.6 v	Black-orange on white

In the troubleshooting tables that follow, reference is made in several places to the use of an oscilloscope to check the waveform at some point in the circuit. Because of the extremely short times involved in many of the waveforms throughout the instrument, it is necessary that you use a wide-band oscilloscope for these checks. A 30-mc instrument such as the Tektronix 540-series oscilloscopes is a minimum. If possible, a 100-mc oscilloscope should be used, such as the Tektronix type 580-series instruments.

TABLE 5-1. V331 SCREEN VOLTAGE

SWEEP RATE	AVERAGE	NORMAL UPPER LIMIT	NORMAL LOWER LIMIT
2 NSEC/CM	115	135	95
5 NSEC/CM	65	80	50
10 NSEC/CM	50	60	40
20 NSEC/CM	35	45	25
50 NSEC/CM	25	35	15
100 NSEC/CM	22	30	14
200 NSEC/CM	18	25	10
500 NSEC/CM	15	25	5
1000 NSEC/CM	13	25	5

5-5

TROUBLE	PROBABLE CAUSES OF TROUBLE	CHECK	IF NORMAL	IF ABNORMAL
 Pilot light and tube fil- aments do not light. 	Line power not ap- plied. Fuse F601. Pow- er switch SW601. Power Transformer T601.	Check to see that os- cilloscope is properly connected to power source. Then check for correct line voltage between terminals 1 and 4 of T601.	T601 is probably de- fective.	F601 or SW601 prob- ably defective.
 Pilot light does not dim after normal time delay. No dc power. 	-250-Volt Power Sup- ply26.5-Volt Pow- er Supply. K601, K602, K603, TK602.	1. Check for —26.5 volts at output of the —26.5-Volt Power Supply.	Check power supply relays.	Proceed to check no. 2.
		2. Check for —250 volts at output of —250-Volt Power Sup- ply.	Trouble is in —26.5- Volt Power Supply. Proceed to step 18 of Table 5-3.	Trouble is in —250- Volt Power Supply. Proceed to step 4 of Table 5-3.
 No spot or trace on screen of oscilloscope. 	Beam positioned off screen. Beam deflect- ed off screen. CRT circuit, Unblanking cir- cuit. Time-Base Gen- erator. High-Vollage Power Supply. Loose CRT Socket.	 Set PUISE AMPLI- TUDE OR SYNC con- trol fully clockwise to free run the sweep. Turn up intensity. A trace should appear. 	Adjust the INTENSITY control for normal brightness, then stop the sweep by setting the FUNCTION switch or PUISE AMPLITUDE OR SYNC control fully counterclockwise. The spot and trace should disoppear. If so, in- blanking circuit is de- fective; proceed to step 54 of Toble 53.	Proceed to check no. 2.
		2. Disconnect any in- put signals. Set HOR- IZONTAL control fully clockwise. Ground vertical positioning plate at neck pin of CRT. A spot or trace should appear.	The vertical position- ing circuit is defec- tive. Check R865, R867, and R868.	Proceed to check no. 3.
		3. Set VERTICAL con- trol to midrange. Stop the sweep, and short between the horizon- tal deflection plates. A spot should appear on the screen.	Trouble is in Time- Base Generator. Pro- ceed to step 56 of Table 5-3.	Trouble is in CRT cir- cuit. Proceed to step 19 of Table 5-3.
 Trace not focused prop- erly. 	ASTIGMATISM or FO- CUS controls not set correctly. CRT Circuit.	Adjust FOCUS and ASTIGMATISM con- trols as described in Operating Instruc- tions. Trace should focus properly.	Instrument is operat- ing correctly.	Trouble is in CRT Cir- cuit. Proceed to step 27 in Table 5-3.
 Incorrect trace geom- etry. 	GEOMETRY control misadjusted. CRT Cir- cuit.	Set GEOMETRY con- trol as described in CALIBRATION proce- dure. Adjustment should correct trouble.	Instrument is operat- ing correctly.	Trouble is in CRT Cir- cuit. Proceed to step 27 in Table 5-3.

TABLE 5-2, CIRCUIT ISOLATION

Table 5-2, (continued)

TROUBLE	PROBABLE CAUSES	CHECK		
	OF TROOBLE	CHECK	IF NORMAL	IF ABNORMAL
6. Sweep inoperative.	Trigger and Holdoff Circuit. Time-Base Gate. Time-Base Gen- erator. Single-Sweep Switch.	 Set PULSE AMPLI- TUDE OR SYNC con- trol fully clockwise and check for approx- imately a +30-volt gate at junction of D198 and D199. 	Proceed to check no. 2.	Trouble is in Trigger and Holdoff Circuit. Proceed to step 34 of Table 5-3.
		 Check for negative gate at the plate of V244; should be ap- proximately 40 volts or more. 	Trouble is in Time- Base Generator. Pro- ceed to step 56 of Table 5-3.	Trouble is in the Time-Base Gate Gen- erator. Proceed to step 46 of Table 5-3.
 Sweep operates but cannot be triggered or synchronized. 	X20 Amplifiers. TRIG- GER SOURCE switch SW10, GAIN switch SW20, Trigger Take- off. Corona in High- Voltage Power Sup- ply, Large external fields. Extraneous sig- nals due to ground loops.	1. Attempt to trigger oscilloscope by apply- ing a signal to the EXTERNAL TRIGGER 125Ω connector. The oscilloscope should trigger properly.	Check the trigger takeoff and SW10.	Check that trigger amplifiers are correct- ly installed. Set GAIN at NORMAL. If proper triggering is still not obtained interchange the two amplifiers. Then proceed to check no. 2.
		2. Place FUNCTION switch at PULSE. Us- ing a test oscillo- scope, trace the trig- gering signal through triggering channel to Q70. The signal should not disappear at any point.	The triggering chan- nel is normal.	The point where the triggering signal dis- appears will isolate the trouble.
 Sweep cannot be syn- chronized in HF SYNC position of FUNCTION switch. 	Countdown Oscillator D50. FUNCTION switch SW50.	With the FUNCTION switch at H.F. SYNC and no triggering sig- nal applied, check that countdown os- cillator D50 is oper- ating at a frequency of approximately 10- 30 mc.	Check FUNCTION switch.	Trouble is in the Countdown Oscillator. Proceed to step 37 of Table 5-3.
9. Short sweep length.	Time-Base Gate. Time- Base Generator. CRT horizontal deflection lead disconnected.	Check width of gate at plate of V244. Width of the gate should be at least 7 times the sweep time per centimeter.	Trouble is in the Time-Base Generator. Check value of timing resistor and adjust- ment of +475-Volt Power Supply.	Trouble is in the Time- Base Gate. Proceed to step 46 in Table 5-3.
 Start of sweep shifts horizontally on screen. 	Holdoff Circuit. Time- Base Generator.	Disconnect triggering signals and free run the sweep. Maximum repetition rate of the triggering circuit at various sweep rates should be approxi- mately as shown in Table 2-2.	Trouble is in Time- Base Generator. Pro- ceed to step 56 in Table 5-3.	Trouble is in Hold- off Circuit. Proceed to step 38 in Table 5-3.
11. Incorrect sweep timing.	Incorrect Timing Ad- justment. Time-Base Generator. +475-Volt Power Supply.	1. Check timing ad- justment. Adjustment should correct diffi- culty.	Instrument is operat- ing correctly.	Proceed to check no. 2.

TROUBLE	PROBABLE CAUSES OF TROUBLE	CHECK	IF NORMAL	IF ABNORMAL
		2. Check output volt- age of +475-Volt Power Supply. Out- put should be vari- able between approx- imately +360 and approximately +550 volts.	Trouble is in the Time-Base Generator Check value of tim- ing resistor R336.	Trouble is in +475- Volt Power Supply. Proceed to step 17 in Table 5-3.
12. Nonlinear sweeps.	Time-Base Generator.	Proceed to step 63 in Table 5-3.		
 No single sweeps ob- tainable. 	Holdoff Circuit.	Proceed to step 43 of Table 5-3.		
14. No vertical deflection.	Improper triggering delay. Loose input connection. Faulty ex- ternal attenuator. Faulty signal source.	 Check delay line connections to CRT. Check for 125 ohms as measured at SIG- NAL 125 connector. Adjust oscilloscope for internal triggering from the applied waveform. The oscil- loscope should trig- ger and the sweep should run. 	Adjust the DELAY control to move waveform on screen. If it appears, the in- strument is operating correctly. If not, the trouble is in the sweep delay circuit. Check R82, C83, R83, R87, and R88.	Proceed to check no. 2.
×		2. Check input con- nections. Substitute external attenuators. Waveform should ap- pear.	Trouble has been cor- rected.	Check for open or shorted input cables. Check signal source. Substitute known in- put waveform.
15. Waveform distortion.	Loose input connect- ors. Improper imped- ance matching of ex- ternal cables and sig- nal sources. CRT ter- mination missing. Ex- cess external cabling. Faulty signal source.	Check for distortion using Calibration- Step Generator. Check for loose input connectors, proper im- pedance matching, and that the termina- tion for the CRT is in place. Proper waveform should ap- pear.	Trouble has been cor- rected.	Check external cable lengths to determine if excess cabling is causing a loss of per- formance. Check the signal source.
 Calibration-Step Gener- ator does not operate properly. 	Calibration-Step Gen- erator.	Proceed to step 68 of Table 5-3.		
 Rate Generator does not operate properly. 	Rate Generator.	Proceed to step 74 of Table 5-3.		

CIRCOIL INCODELSIIOOIIII	TABLE	5-3,	CIRCUIT	TROUBLESHOOTING
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IF NORMAL	IF ABNORMAL
•	•
Proceed to step 2.	Proceed to step 4.
Proceed to step 3.	Check C613, C646, and C647.
Power supply is operating correctly.	Check power-line waveform for ex- treme flat-top or peaking. Check D610, D611, D612 and D613, Check V627, R627 and K602-4.
If power supply output voltage is within approximately 25 volts of -250 volts, proceed to step 7. If more negative than -275 volts, proceed to step 8. If less negative than -225 proceed to step 9.	Proceed to step 6.
Proceed to step 6.	If voltage is abnormally high, check R610. If R610 is normal, there is an unusually high load on the power supply. Check for cause of extra loading. If voltage is low, proceed to step 6.
Check D610, D611, D612, and D613. Proceed to step 7.	Check T601.
Trouble is apparently corrected. Check vertical deflection factor and sweep timing.	If output voltage varies but cannot be set to correct value, check V639, R646, R647, and R648. Recheck step 5. If the output voltage does not vary
	check V646, V639, V624, and V627.
Proceed to step 11.	Be certain —250-Volt Power Supply is adjusted and regulating. Then proceed to step 13.
Proceed to step 12.	Check C661, C662, C663, and C668.
Power supply is operating normally.	Check power line waveform for ex- treme flat-top or peaking. Check V627, R677, D660, D661, D662, and D663.
	IF NORMAL Proceed to step 2. Proceed to step 3. Power supply is operating correctly. If power supply output voltage is within approximately 25 volts of250 volts, proceed to step 7. If more negative than -225 volts, proceed to step 7. Proceed to step 8. If less negative than -225 proceed to step 9. Proceed to step 6. Check D610, D611, D612, and D613. Proceed to step 7. Proceed to step 11. Proceed to step 12. Power supply is operating normally.

	STEP	IF NORMAL	IF ABNORMAL
13.	Check voltage across C661, C662, and C663.	If output voltage of the power sup- ply is near normal, check R688 and R689. If output voltage is consider- ably higher than normal, proceed to step 15. If the output voltage is much lower than normal proceed to step 16.	Proceed to step 14.
14.	Check voltage drop across R660 and R661.	Check D660, D661, D662, D663, and T601.	If voltage is abnormally high, check R660 and R661. If the resistors are normal, there is an unusually high load on the power supply. Check for cause of extra loading. If voltage is low, check D660, D661, D662, D663, and T601.
15.	Check for the following tube conditions: V627B shorted, V674 open (open filament or low emission), or V686 shorted. Use voltage and resistance checks if neces- sary.		
16.	Check for the following tube conditions: V627B open (open filament or low emis- sion), V674 shorted, or V686 open. Use voltage and resistance checks if neces- sary.		
+4	75-Volt Power Supply		
17.	Use the same general troubleshooting procedure as for the —250-Volt Power Supply.		
Oth	er Low-Voltage Power Supplies		
18.	Use the same general troubleshooting procedure as for the $+225$ -Volt Power Supply. Always check supplies in this order: (1) -250 , (2) -26.5 , (3) $+225$, (4) $+100$, $+450$, $+475$, or $+650$.		
CRT	Circuit		-
19,	Check for4 kv at the junction of R854 and R855 (HIGH VOLTAGE TEST POINT) being careful to avoid contact with the circuit. Switch off power while making connections.	Proceed to step 20.	If the voltage is near zero, proceed to step 22. If the voltage is not near zero, pro- ceed to step 25.
20.	Vary setting of the INTENSITY control, and vary line voltage between 105 and 125 volts while checking voltage at junc- tion of R854 and R855. The voltage should remain in regulation.	Proceed to step 21.	Proceed to step 26.
21.	Use a high-voltage probe and a great deal of care to measure the voltage at the ungrounded side of C835. The volt- age should be approximately 20 kv.	The high voltage power supply is operating normally. The trouble is probably a defective CRT.	Check V802, V812, V822, and their associated components.
22.	Check Fuse F801 and R801.	Proceed to step 23.	Replace F801 or R801.
23.	Check to see if oscillator, V800, is oper- ating by seeing if the filaments of V802, V812, V822, and V832 are lit. This should be apparent without removing the metal shield.	Proceed to step 24.	Check for 20-kc oscillator waveform at pin 5, V800. Check V800 or T801.

	STEP	IF NORMAL	IF ABNORMAL
24.	Check voltage at plate of V832. This voltage should be more than $-4 \mathrm{kv}.$	Check R856, R855, B853, B854, B855, and B856.	Check T801, V832, C832, and C833.
25.	Use R841 to adjust voltage at junction of R854 and R855 to4 kv. The adjust- ment should be possible.	The trouble has been corrected.	If voltage varies with setting of R841, check resistance of R844, R843, R842, R841, and R840. If voltage does not vary, check V814A and V814B and their associated com- ponents.
26.	Check that neon bulbs B853, B854, B855, and B856 are lit at all settings of the INTENSITY control.	Check V814, V800 and associated components.	Check R855, R856 and the neon bulbs. The drop across each neon bulb should be from 55 to 70 volts.
27.	Perform step 19.	Proceed to step 28.	Follow information given for steps 19 and 21.
28.	Check voltages obtained from FOCUS, GEOMETRY, and ASTIGMATISM controls.	Check CRT tube socket connection. Check the CRT.	Check the circuit of the appropriate control.
Tim	e-Base Trigger and Holdoff Circuit		
29.	Set the FUNCTION switch to H.F. SYNC and use test oscilloscope to check (at junction of R63 and D69) whether the Countdown Oscillator is running. Approx- imately a 10- to 30-mc signal should be present.	Proceed to step 30.	Proceed to step 37.
30.	Set the FUNCTION switch to PULSE and PULSE AMPLITUDE OR SYNC control just short of the free-running position. Con- nect a triggering signal to the EXTERNAL IRIGER I250 connector and set RIG- GER SOURCE switch to either + EXT. or -EXT. The triggering signal should ap- pear at the junction of R63 and D69.	Proceed to step 31.	Trace the triggering signal through the triggering channel to point where signal is lost. The point where signal disappears will isolate the trouble.
31.	Check that the triggering signal appears at input to Second X20 Amplifier.	Proceed to step 32.	Check D68, D69, R63, R64, and R66A.
32.	Check that the triggering signal appears at output of Second X20 Amplifier.	Proceed to step 33.	Trouble is in X20 Amplifier. Verify by interchanging amplifiers. Trace signal through amplifier to detect cause.
33.	Check that the triggering signal appears at junction of D71 and D72.	Proceed to step 34.	Check components between the out- put of Second X20 Amplifier and T70.
34.	Determine whether Q70 is being trig- gered by observing the waveform at the collector of Q70. If the transistor is trig- gering, a large positive-going swing should appear at the collector.	Proceed to step 35.	Check the READY lamp on the front panel. If the lamp is lit, check Q70 and T70. If the lamp is not lit, pro- ceed to step 38.
35.	Using a test oscilloscope, check to see whether Q180 is being triggered by ob- serving the workform at the collector. If Q180 is triggering, a positive-going swing should be apparent at the collec- tor.	Proceed to step 36.	Try all settings of the DELAY con- trol. If any setting of R88 causes (2180 to trigger, the trouble is probably in the delay circuit. Re- sistance checks can be used to locate the cause of the trouble. If (2180 does not trigger at any set- ting of R88, check Q180, D82, T70, and T180.

	STEP	IF NORMAL	IF ABNORMAL
36. Check 30 vc D198	for a positive gate approximately lts in amplitude at the junction of and D199.	The Time-Base Trigger Circuit is operating correctly.	Check V184, V194, and their associ- ated components.
37. Set th Check of D5	the FUNCTION switch to H.F. SYNC. to about -0.2 volt at the cathode 0 .	Check D50 and L52.	Check R56, R54, and R55 and the —26.5-Volt Power Supply.
38. Set the the P fully SING Check condu measu sistors	the FUNCTION switch to PULSE and ULSE AMPLITUDE OR SYNC control counterclockwise. Set the NORMAL- LE SWEEP switch to NORMAL- that V114 is cut off and V134 is citing. This can be determined by uring the drop across the plate re- of the two tubes.	Check V123B. Proceed to step 39 if necessary.	Check V114, V134, and the grid re- sistors for the two tubes.
39. Rotate contro of po of V1	 the PULSE AMPLITUDE OR SYNC of fully clockwise. Check for a series sitive 7- to 10-volt pulses at pin 2 14. 	Proceed to step 40.	Check for —0.3 v at cathode of D144. Also check Q70, R78, R77 and T70.
40. Check positiv The +100	that V114 conducts each time a ve pulse is applied to the grid. plate voltage should drop from volts.	Proceed to step 41.	Check V114 and its plate resistors. Check the grid-to-cathode voltage to be sure that positive pulses ap- plied to the grid cause the tube to conduct.
41. Check V114 cut o the se When should	that V134 cuts off each time that conducts, and that V134 remains ff for a period which varies with etting of the NANOSEC/CM switch. V134 cuts off, its plate voltage d rise to approximately +75 volts.	Proceed to step 42.	Check V123A and its associated components.
42. Check The b time remai deper SEC/0 val, tially.	the voltage at the base of Q70, asse of Q70 should go positive each that Q70 completes a cycle and n there for a period of time that ds on the setting of the NANO- ZM switch. After a definite inter- the voltage should drop exponen-	The holdoff circuit apparently is operating correctly.	Check V143A and its cathode circuit.
43. Place switch TION AMPL count tons.	the NORMAL-SINGLE SWEEP at SINGLE SWEEP. Set the FUNC- switch at PULSE and the PULSE ITUDE OR SYNC control fully erclockwise. Press the RESET but- The READY lamp should light.	The reset circuit is operating normally.	Proceed to step 44.
44. Conne tion o positi time	ect a test oscilloscope at the junc- of D160 and D161 and check that a ve pulse occurs at this point each the RESET button is pressed.	Check D160.	Proceed to step 45.
45. Check It sh —26.	the collector voltage of Q160. would normally be approximately 5 volts.	Check Q160 and the base circuit.	Check Q160, C164, R164, SW160, SW168.
Time-Base	Gate and Unblanking Circuit		
46. Check pin 1	for approximately +3.7 volts at of V244.	Proceed to step 47.	Check Q238, R230, and R231.

STEP	IF NORMAL	IF ABNORMAL
47. Set the PUISE AMPLITUDE OR SYNC con- trol fully clockwise and set the NORMAL. SINGLE SWEEP switch to NORMAL Set the NANOSEC/CM switch to 10. Check the duration of the gate appearing of pin 2 of V284. The duration of the gate should be 70 nanoseconds.	Proceed to step 48.	Check the NANOSEC/CM switch and the coaxial cables used for gate timing. See that Q180, V184, V194, D197, and D198 are functioning properly.
48. Set the NANOSEC/CM switch to 1000 and check the length of the gate appearing at pin 2 of V244. The gate duration should be approximately 9 μ sec.	Proceed to step 49.	Proceed to step 50.
49. Check the waveform at pin 8 of V244. The waveform should be a negative gate of more than 40 volts with a duration of approximately 9 µsec.	The Time-Base Gate Generator is operating correctly.	Check V244 and the associated cir- cuitry.
 Set the NANOSEC/CM switch to 20 and check that a series of positive pulses or gates appear at pin 1 of V264. 	Proceed to step 51.	Check R256 and R260.
 Check for a series of negative-going pulses (near 160 volts amplitude) at pin 5 of V264. 	Proceed to step 52.	Check V264 and the associated com- ponents.
 Check waveform at pin 8 of V274. The waveform should be a series of positive- going gates with a duration of approxi- mately 180 nsec. 	Proceed to step 53.	If the gates do not appear, check C262 and V274. If the duration of the gates is incorrect, check the set- ting of C262 using the Calibration Procedure. Then check R270 and R271.
53. Check for a series of 180-nsec positive gates at pin 8 of V393B.	Check V283 and the associated com- ponents.	Check V393B and R280, R281, and R282.
54. Adjust the oscilloscope for a free-running sweep, and set the NANOSEC/CM switch to 1000. Check for positive gates at pin 2 of V214. The gate duration should be approximately 9 µsec.	Proceed to step 55.	Check L250.
55. Check the waveform at pin 8 of V214. This should be a series of negative gates approximately 9 µsec in duration and ap- proximately 110 volts in amplitude.	The unblanking circuit is operating normally.	If no gate is present, check V214. If the gate is abnormally large, check V223 and D220.
Time-Base Generator		
	CAUTION	

Certain failures in the Time-Base Generator may cause damage to the screen grid of V331. Consequently, when troubles are traced to the Time-Base Generator, the screen-grid and control-grid voltages of V331 should be checked immediately. If the screen-grid voltage is more than 140 volts, and the control-grid voltage is within 1 volt of ground, the instrument should not be left on for more than 30 seconds at a time (measured from the end of the time delay period). Voltage readings must be made within the 30-second periods.

56. Observe CAUTION above. Check for +155 volts at the plate of V331 with the NANOSEC/CM switch at 2.	Proceed to step 57.	If plate voltage is zero, check V332, R336, and R334. If the plate is near +475 volts first check the setting of R374, then check V331. If the plate is near +155 volts try to adjust to sergest utility with P274. If traible is
		correct value with R374. If trouble is
		not corrected, proceed to step 37.

STEP		IF NORMAL	IF ABNORMAL	
57.	Check for -3.2 volts at the control grid of V331 with the NANOSEC/CM switch at 2.	Proceed to step 58.	Use R396 to adjust voltage to cor- rect level. If trouble cannot be cor- rected, proceed to step 60.	
58.	Check that screen-grid voltage of V331 is less than ± 140 volts.	Circuit is operating correctly.	Replace V331.	
59.	Check the control-grid voltage of V331. If the plate voltage of V331 is high, the control-grid voltage should be less than -3.2 volts. If the plate voltage of V331 is low, the control-grid voltage should be more than -3.2 volts.	The Plate-Voltage Regulator is oper- ating, Proceed to step 61.	Check V312, V322, Q318, Q328, V374, V363 and V343. Check volt- ages in Plate-Voltage Regulator against those indicated on the sche- matic diagram.	
60.	Check the screen-grid voltage of V331 against the values in Table 5-1.	Circuit should be operating correct- ly. Recheck the setting of R396.	Proceed to step 61.	
61.	Check plate voltage of V394. The voltage should be higher than normal if the control-grid voltage of V331 is less than -3.2 volts. The voltage should be less than normal if the control-grid voltage of V331 is more than -3.2 volts.	Check V403.	Check V393A and V394.	
62.	Free-run the sweep and check the ampli- tude of the gate at the grid of V331. The gate should drive the grid of V331 neg- ative by at least 40 volts.	Proceed to step 63.	Check D305, D306, D307 and D308, and C312.	
63.	Set the NANOSEC/CM switch to 1000 and free run the sweep. Check for a sawtooth waveform of approximately 150 volts or more at pin 3 of V343. Use a 10X probe on the test oscilloscope.	Proceed to step 64.	Check V343, V332, and R336.	
64.	Check for approximately a 150-volt or more sawtooth waveform at pin 2 of V424.	Proceed to step 65.	Check D344, D345, and C344.	
65.	Check for approximately a 150-volt or more sawtooth waveform at pins 3 and 8 of V353.	Proceed to step 66.	Check V353.	
66.	Check for a linear sawtooth of approximately 150 volts at pins 3 and 6 of V424. Pin 6 negative waveform may have low amplitude due to capacitance of test probe.	Proceed to step 67.	Check V424. See that deflection plate leads are connected properly.	
67.	Check for linear sawtooth waveform of approximately 150 volts or more at the junction of R440 and R445, and at the junction of R442 and R446.	Time-Base Generator is operating correctly.	Check D430, D431, R433, R435, C435, C433, and C430.	
Cal	ibration-Step Generator			
68.	Set the RANGE switch to 10V TO 125 $\Omega.$ You should hear the reed switch operating.	Proceed to step 70.	Adjust the FREQUENCY and DRIVE controls to see if the reed will vi- brate. If it does, proceed to step 70. If not, proceed to step 69.	
69.	Check that the oscillator is operating by observing the waveform at pin 5 of V885. The normal waveform should be present.	Check L885 and R885.	Check V885 and V895A.	

Table 5-3, (continued)

	STEP	IF NORMAL	IF ABNORMAL	
70.	Check the frequency of oscillator opera- tion while varying the setting of the FREQUENCY control. The frequency should vary between approximately 250 and 1100 cycles per second.	Proceed to step 71.	Check the network in the plate cir- cuit of V895A and the grid circuit of V885.	
71.	Display the output waveform of the Cali- bration-Step Generator on the Type 519. It should be possible to obtain triggering.	Proceed to step 72.	If the sweep cannot be triggered properly, check the charging net- work, charging voltage attenuators, and the reed switch. Be sure Cali- bration-Step Generator has a dc load (no series capacitor).	
72.	Check the waveform displayed on the Type 519. There should be no appreci- able time or amplitude jitter.	Proceed to step 73.	Check the reed switch.	
73.	Check the amplitude of the displayed step waveform on the Type 519 against the settings of the POLARITY, RANGE and VOLTS controls.	Calibration-Step Generator is ap- parently operating correctly.	Check the charging voltage attenu- ators for proper values.	
Rate	e Generator			
74.	Set the MULTIPLIER switch to X1000 and the CYCLES/SEC control to 30. Check for approximately a 30-kek switching waveform at pin 3 of V9158.	Proceed to step 75.	If the multivibrator is not operating, check to see if other settings of the MULTIPLER and CYCLES/SEC con- trols will cause the multivibrator to run. If so, check the components used only in the inoperative posi- tions. If not, check V915 and V895, and components common to all set- tings of the controls. If the fre- quency of operation is incorrect check K923, R924, and the appropri- ate C920 values.	
75.	Connect the output waveform from the $+RATE 50 \Omega$ connector to the input of the Type 519 through a T50/T125 adapter. Check the appearance of the waveform. It should be approximately 7-10 volts in amplitude, as displayed on the screen, and approximately 10 nsec in duration.	Proceed to step 76.	If no pulses are present, adjust the AVALNCHE SET control as des- cribed in the Calibration Procedure. If the trouble cannot be corrected, heck Q934 and its emitter and col- lector circuits. If the pulse amplitude is incorrect or if the waveform is severely distorted, check Q934 and the emitter circuit.	
76.	Set the MULTIPLIER switch to OFF and check to see that the waveform disap- pears from the screen of the Type 519.	The Rate Generator is operating correctly.	The Avalanche stage is free run- ning. Adjust the AVALANCHE SET control to the point where the gen- erator just stops free running. If the adjustment cannot be made, check R933, R932, R931 and Q934.	

NOTES



INTRODUCTION

The following paragraphs outline the procedure used to calibrate the Type 519 Oscilloscope. The instrument should not require frequent recalibration, but occasional adjustments will be necessary when tubes and other components are changed. Also, a periodic recalibration is desirable from the standpoint of preventive maintenance.

Apparent troubles in the instrument are occasionally the result of improper calibration of one or more circuits. Consequently, calibration checks should be an integral part of any troubleshooting procedure. Abnormal indications occuring during calibration checks will often aid in isolating troubles to a definite circuit or stage.

In the instructions that follow, the steps are arranged in the proper sequence for a complete calibration of the instrument. Each numbered step contains the information required to make one check or adjustment or a series of related checks or adjustments. The steps are arranged to avoid unnecessary repetition of checks or adjustments.

EQUIPMENT REQUIRED

The following equipment or its equivalent is required to perform a complete calibration of the Type 519 Oscilloscope.

 An accurate dc voltmeter with a sensitivity of 20,000 ohms per volt or more.

 A nonloading dc voltmeter such as John Fluke 800, if available.

 An accurate ac voltmeter capable of reading voltages from 105 to 125 or from 210 to 250 volts.

4. An ohmmeter.

 An autotransformer with output voltage variable between 105 and 125 or 210 and 250 volts. Minimum rating of 1 kva.

6. A test oscilloscope with a bandpass to at least 30 mc and a maximum sensitivity of at least 50 millivolts per centimeter. Must also have a sensitivity of at least 5 milliv volts per centimeter at reduced bandpass. Oscilloscope such as a Tektronix Type 540-Series and Type L Plug-In Unit is suitable.

7. Time-mark generator capable of generating 1 µsec markers and 5, 10, and 50 mc sine waves, such as the Tektronix Type 180A.

CALIBRATION PROCEDURE

8. A 125Ω , 1 KMC Timing Standard, such as Tektronix Part No. 017-019, or accurate source of sine waves with frequency to 1000 mc.

9. Miscellaneous adapters and cables.

10. Miscellaneous alignment tools and other hand tools,

PRELIMINARY PROCEDURE

Make a complete visual check of the instrument. Then use an ohmmeter to make a check on the resistance at the regulated bus of each power-supply lead to ground at the test points shown in Fig. 6-1. The values of resistance should be approximately as follows:



Fig. 6-1. Locations of low-voltage power supply test points on the top of the instrument near the CRT.

TABLE 6-1

POWER SUPPLY LEAD	RESISTANCE TO GROUND
—250 volts	12 k
-26.5 volts	2 Ω
-14 volts	2 Ω
+100 volts	3 k
+225 volts	4 k
+320 volts (unregulated)	4 k
+450 volts	50 k
+475 volts	30 k
+650 volts	40 k or higher

Set the INTENSITY control fully counterclockwise and connect the power cord and ac voltmeter to the output of

Calibration Procedure—Type 519

the variable autotransformer. Set the POWER switch to ON and adjust the autotransformer for an output of 117 volts (or other voltage for which the instrument is wired). Allow the instrument to warm up for several minutes before proceeding with the calibration steps.

ADJUSTMENT PROCEDURE

1. Adjust -250-volt Power Supply

Connect the dc voltmeter to the -250-volt test point shown in Fig. 6-1. Set the -250 V control for exactly -250 volts.

2. Check the Low-Voltage Power Supplies

Stop the sweep by setting the FUNCTION writch to PUISE and the PUISE AMPUTIDE OR SYNC control fully counterclockwise. Use the dc voltmeter to check the output of each of the low-voltage power supplies. Yary the output of the autotransformer between 105 and 125 for 210 and 250) volts with checking that the power supplies regulate over the entire range. Check for proper regulation at 125 volts with the NANOSEC/CM switch set at 1000, and at 105 volts with the NANOSEC/CM switch set at 2. Using the high gain plug-in unit for the test oscilloscope, check the ripple voltage of each power supply. Voltages and ripple voltages should be approximately as follows:

NOTE

Before making Power Supply ripple measurements, be sure the Rate Generator is OFF and the Calibrator is in the standby position with the Reed Drive in other than Single Occurrence.

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		RIPPLE VOLTAGE (peak-to-peak
POWER SUPPLY	OUTPUT VOLTAGE	typical)
-250 volts	-250 volts	3 mv
-26.5 volts	-26.5 ± 0.5 volts	5 mv
—14 volts	-14 ±0.5 volts	3 mv
+100 volts	+100 ±3 volts	3 mv
+225 volts	+225 ±5 volts	2 mv
+450 volts	450 ±10 volts	10 mv
+475 volts	+400 to 550*	100 mv
+650 volts	+650 ±25 volts	50 mv

*Depends on setting of NANOSEC/CM switch.



Fig. 6-2. Location of the V331 control-grid test point.



Fig. 6-3. Location of the 400 KC REP. RATE adjustment, R126A.

3. Set Sweep Repetition Rate

Set the PULSE AMPLITUDE OR SYNC control fully clockwise and the NANOSEC/CM switch to 2. Connect the probe of the test oscilloscope to the grid line of V331 at the point shown in Fig. 6-2. Adjust R126A [see Fig. 6-3] for 2.5 µsec between the gates displayed on the test oscilloscope. Check the repetition rates of other ranges using the following data:

NANOSEC/CM SETTING	APPROXIMATE TIME BETWEEN GATES
5	5.5 µsec
10	11 µsec
20	24 µsec
50	60 µsec
100	100 µsec
200	200 µsec
500	500 µsec
1000	l ms

TABLE 6-3

4. Set Position of PULSE AMPLITUDE OR SYNC Knob

Set the PULSE AMPUTUDE OR SYNC control counterclockwise and the NANOSEC/CM switch to 2. Connect a dc voltmeter to the center terminal of R76 (located an the front ceramic strip on the Time-Base Trigger chassis) and with a screwdriver set the potentiometer for a reading of maximum negative volts. Turn the PULSE AMPUTUDE OR SYNC control solw/ clockwise until the sweep time free runs, then



Fig. 6-4. Location of the SWEEP GATE LENGTH adjustment, C262.

back the control off slightly. Adjust R76 until the sweep just stops running. The dot on the PULSE AMPLITUDE OR SYNC knob should be just above the RECURRENT line on the front panel. If not, loosen the knob on the shaft and rotate it to the proper position. Relighten the knob. The sweep should now free run any time that the knob is positioned clockwise past the line.

5. Set Time-Base Gate Duration

Set the PULSE AMPLITUDE OR SYNC control fully clockwise and connect the test oxicilloscope to the grid line of V331, at the test point shown in Fig. 6.2. Set the NANO-SEC(CM witch to 1000 and odjust C262 gee Fig. 6.4] for 9-usec gates displayed on the test oscilloscope. Measure the duration of the gate at the b5% vollage level. Check the duration of the gates at the other sweep rates according to the following table:

TABLE 6-4

NANOSEC/ SETTING	GATE DURATION (at 50% voltage level) 30-Megacycle Oscilloscope
500	4.5 μsec minimum
200	1.8 µsec minimum
100	900 nsec minimum
50	450 nsec minimum
20	180 nsec minimum
10	70 nsec approximately
5	35 nsec approximately
2	14 nsec approximately

6. Check Gate and Unblanking Amplitudes

Set the NANOSEC/CM switch to 10 and free run the sweep. Check for the following gate amplitudes:

TABLE 6-5

LOCATION	AMPLITUDE	
Pin 2 of V244	+20 volts minimum	
Plate of V214	Approximately	
Grid line of V331	-35 volts minimum	

7. Check + Trigger Pulse

Free run the sweep and connect the probe of the test oscilloscope to the $+TRIGGER~50\,\Omega$ connector on the front ponel of the Type 519. The pulses displayed on the test oscilloscope should be approximately 50 nsec in duration as measured at the 50% voltage level, and should have a peak amplitude of approximately 4 volts when unterminated.

8. Set Delayed + Gate

Set the NANOSEC/CM switch to 1000 and free run the sweep. Check the amplitude of the waveform at the DELAYED +GATE 50 Ω connector by connecting the probe of the test oscilloscope directly to the connector. The gates displayed on the test oscilloscope should have a peck amplitude of approximately 9 to 10 volts when unterminated, falling off to approximately 50% of this at the end of the gate (see Fig. 6-5).

Connect the output of the +TRIGGER 50 Ω connector to the external triggering connector of the test oscilloscope and trigger the oscilloscope from this signal. Set the NANOSEC/CM switch to 10. Observe the output of the DEIAYED +GATE 501 connector while rotating the DEIAY control. Adjust C84 until it is possible to move the delayed +gate 35 ness in time by means of the DEIAY control.



Fig. 6-5. Typical Delayed + Gate as displayed on a 30 mc test oscilloscope.

9. Check Single-Sweep Operaton

Set the NORMALSINGLE SWEEP switch to SINGLE SWEEP and the PULSE AMPLITUDE OR SYNC control fully clockwise. Set the NANOSEC/CM switch to 1000. The sweep should run once each time the RESET button is pressed.

Set the PULSE AMPLITUDE OR SYNC control fully counterclockwise. The READY lamp should light when the RESET button is pressed, but the sweep should not run.

On instruments with serial numbers above 403, an external SINGLE SWEEP triggering jack was installed. When making connections to the plug, use only the tip and ring connections. If the shield is grounded, improper operation will result.

10. Set Grid Voltage of V331

Set the NANOSEC/CM switch to 2 and stop the sweep. Connect the dc voltmeter to the grid test point of V331 and adjust the 4CX250F GRID -3.2 VOLTS control for exactly -3.2 volts.

11. Set Plate Voltage of V331

Connect the dc voltmeter to the plate of V331 at the plate strap of the tube. Adjust R374 (see Fig. 6-6) for exactly +155 volts at the plate.



Fig. 6-6. Location of the V331 PLATE + 155 V adjustment, R374.

12. Check Calibration-Step Generator Charging

Voltage

Set the VOLTS control to 10 and the RANCE switch to 10V TO 1250. Measure the de voltage at the input to the charging network. The voltage should be approximately 202 volts, using a nonloading voltmeter. Set the RANCE switch to 1 V TO 500 and again check the voltage at the input to the charging network. The voltage should now be approximately 8.8 volts, using a nonloading voltmeter. Set the RANCE switch to STANDBY.

13. Adjust High Voltage

Turn off the oscillacscope power. Connect the dc voltmeter to the NY TEST POINT on the right side of the instrument (be sure the voltmeter is capable of indicating at least 4 ky.) Turn on the oscillacscope and allow the normal warmup period to pass. Then adjust the HIGH VOITAGE contol for -4 ky. Check to see that the high voltage power supply regulates by first positioning the intensity up and down. Then adjust the adjusting the intensity up and down. Then adjust the output voltage from the variable autotransformer between 105 and 125 (or 210 and 250) volts and again check that the high voltage power supply remains in regulation. Switch off the oscillacscope power and disconnect the meter. Then turn on the oscillacscope power and disconnect the meter.

Apply Calibration Step Generator OUTPUT to the 125 Ω SIGNAL Input connector and display the waveform on the CRT. Set the Calibrator Step Generator RANGE switch to 10 V TO 125 Ω and turn the helidial to the same voltage that appears on the CRT facemask (vertical deflection factor). Adjust the HIGH-VOLTAGE control (R841) for exactly one centimetr of CRT deflection.

14. Set Maximum Intensity

With the PULSE AMPUTUDE OR SYNC control fully counterclockvise and the FUNCTION switch in the PULSE position, slowly rotate the INTENSITY control until a low intensity spot appears on the CRT. Use the FOCUS and ASTICMA-TISM controls to bring the spot into shorp focus. Rotate the Max Intensity (R855) control fully counterclockwise and the INTENSITY control fully colchwise. Slowly turn the R855 control clockwise until the spot reappears and a halo forms around the spot. Then turn the R855 control cour.terclockwise until the halo just disoppears.

6-4

15. Adjust Axis Rotation

Free run the sweep and position the trace near the center horizontal graticule line. Set the NANOSEC/CM switch to 1000. Adjust the AXIS ROTATION control so that the trace runs parallel to the horizontal graticule line.

16. Adjust Geometry

Position the trace to the top of the screen and adjust the GEOMETRY control to minimize bowing of the trace. Position the trace to the bottom of the screen and recheck the setting of the control. Make the final setting of the GEOMETRY control so that the best overall geometry is obtained.

NOTE

Make final setting of high voltage to produce correct vertical deflection factor before proceeding with sweep timing adjustments.

17. Set Sweep Timing

Set the NANOSEC/CM switch to 50 and apply 50-mc sine waves from the time-mark generator to the input of the Type 519. Obtain a stable display of the sine waves on the screen of the oscilloscope. Adjust the 50 nSEC control to obtain 2/5, cycles per centimeter over the righthand 5 centimeters of the display. Adjust C425 for optimum linearity while the 50 nSEC control is being set.

Set other sweep rates as indicated in Table 6-6.

TABLE 6-6

SWEEP RATE	INPUT SIGNAL	CONTROL	ADJUST FOR
1000 nsec/cm	1-μsec time markers	1 μSEC	1 marker/cm
500 nsec/cm	5-mc sine waves	500 nSEC	21/2 cycles/cm
200 nsec/cm	5-mc sine waves	200 nSEC	1 cycle/cm
100 nsec/cm	10 mc sine waves	100 nSEC	1 cycle/cm
20 nsec/cm	50-mc sine waves	20 nSEC	1 cycle/cm
10 nsec/cm	50-mc sine waves	10 nSEC	1 cycle/2 cm

Connect the 1 KMC Timing Standard to the SIGNAL 1250 connector on the front panel of the Type 519. Connect a 125-ohm cable from the OUTPUT 1250 connector to the 1 KMC Timing Standard. Set the VARIABLE control fully clockwise, the TRIGGER SOURCE switch to + CAL, and the POLARITY switch to +. Be sure that the trigger takeoff or the Calibration-shep Generator is installed in the INTR GRE RANCE switch to VARIABLE. Adjust the DRIVE and TREQUENCY controls for table operation of the read switch and adjust the triggering controls for a stable display of the timing swerform on the screen of the sourceforms, the start of the sweep when the NANOSEC/CM switch is set to 5.

With the timing waveform displayed and the NANO-SEC/CM switch set to 5, adjust the 5 nSEC control for 5 cycles per centimeter over the middle 4 centimeters of the display. Set the NANOSEC/CM switch to 2, position the trace to start approximately 1 centimeter to the left of the

graticule marks and readjust the DELAY control so the first graticule mark is lined up with the first peak of the timing waveform. Adjust the 2 nSEC control for 2 cycles per centimeter over 6 centimeters.

18. Set Rate Generator Frequency

Connect the 10X attenuator probe of the test oscilloscope to pin 3 of V9158. Set the MULTIPLIER control to X10 and the CYCLES/SEC control to 3. Adjust the FREQUENCY RANGE control for a multivibrator frequency of 30 cps (33.3 msec between waveforms).

Set the MULTIPLIER switch to X1000 and the CYCLES/SEC control to 3. Set C920D (see Fig. 6-7) for 3 kc (333 μ sec between multivibrator waveforms).

Set the CYCLES/SEC control to 30. Adjust C920E (see Fig. 6-7) for 30 kc (33.3 μsec between multivibrator waveforms).

19. Adjust the Avalanche Generator

Set the CYCLES/SEC control to 30 and the MULTIPLIER switch to X1000. Display the output of the rate generator

Calibration Procedure—Type 519



Fig. 6-7. Location of the Rate Generator frequency adjustments, C920D and C920E.

on the Type 519 by connecting the 50-bhm end of a T50/T125 adaptor into the +FATE 50 Ω connector and then connecting from the T50/T125 adaptor to the SIGNAL 125 Ω connector. Set the NANOSEC/CM switch to 100. Set the AVALANCHE SET control fully clockwise and then turn the control slowly counterclockwise until the avalanche generator just stops free running (as indicated by a sudden decrease in the number of pulses displayed on the screen).

NOTES

PARTS LIST ABBREVIATIONS

BHB	binding head brass	int	internal
BHS	binding head steel	lg	length or long
cap.	capacitor	met.	metal
cer	ceramic	mtg hdw	mounting hardware
comp	composition	OD	outside diameter
conn	connector	OHB	oval head brass
CRT	cathode-ray tube	OHS	oval head steel
csk	countersunk	PHB	pan head brass
DE	double end	PHS	pan head steel
dia	diameter	plstc	plastic
div	division	PMC	paper, metal cased
elect.	electrolytic	poly	polystyrene
EMC	electrolytic, metal cased	prec	precision
EMT	electrolytic, metal tubular	PT	paper, tubular
ext	external	PTM	paper or plastic, tubular, molded
F & I	focus and intensity	RHB	round head brass
FHB	flat head brass	RHS	round head steel
FHS	flat head steel	SE	single end
Fil HB	fillister head brass	SN or S/N	serial number
Fil HS	fillister head steel	SW	switch
h	height or high	TC	temperature compensated
hex.	hexagonal	THB	truss head brass
HHB	hex head brass	thk	thick
HHS	hex head steel	THS	truss head steel
HSB	hex socket brass	tub.	tubular
HSS	hex socket steel	var	variable
ID	inside diameter	w	wide or width
incd	incandescent	WW	wire-wound

PARTS ORDERING INFORMATION

Replacement parts are available from or through your local Tektronix, Inc. Field Office or representative.

Changes to Tektronix instruments are sometimes made to accommodate improved components as they become available, and to give you the benefit of the latest circuit improvements developed in our engineering department. It is therefore important, when ordering parts, to include the following information in your order. Part number, instrument type or number, serial or model number, and modification number if applicable.

If a part you have ordered has been replaced with a new or improved part, your local Tektronix, Inc. Field Office or representative will contact you concerning any change in part number.

SPECIAL NOTES AND SYMBOLS

imes000	Part first added at this serial number
00 imes	Part removed after this serial number
*000-0000-00	Asterisk preceding Tektronix Part Number indicates manufactured by or for Tektronix, Inc., or reworked or checked components.
Use 000-0000-00	Part number indicated is direct replacement.
0	Screwdriver adjustment.
	Control, adjustment or connector.

SECTION 7 ELECTRICAL PARTS LIST

Values are fixed unless marked Variable.

Bulbs

Ckt.	No.	Tektronix Part No.	Serial/Model Eff	No. Disc	De	scription	
B128 B128 B601 B606 B607		Use 150-0027-00 150-0030-00 150-0001-00 150-0001-00 150-0001-00	101 820	819	Neon, NE-23 Neon, NE-2V Incandescent #47 Incandescent #47 Incandescent #47	Ready Pilot Light Graticule Light Graticule Light	
8853 8854 8855 8856 8856 8857		Use 150-0027-00 Use 150-0027-00 Use 150-0027-00 Use 150-0027-00 150-0027-00	X470		Neon, NE-23 Neon, NE-23 Neon, NE-23 Neon, NE-23 Neon, NE-23		
B858 B870		150-0036-00 150-0018-00	X988		Neon, A1D Incandescent #12	ON	

Capacitors

Tolerance ±	20% Unless otherwise	indicated.					
C29 C32 ¹ C33 ¹ C43 ¹ C44 ¹	Use 283-0057-00 281-0509-00 283-0026-00 283-0026-00 281-0518-00			0.1 µf 15 pf 0.2 µf 0.2 µf 47 pf	Discap Cer Discap Discap Cer	200 v 500 v 25 v 25 v 500 v	10%
C46 ¹ C47 ¹ C51 C69 C70	283-0024-00 283-0010-00 281-0500-00 Use 283-0057-00 283-0028-00			0.1 μf 0.05 μf 2.2 pf 0.1 μf 0.0022 μf	Discap Discap Cer Discap Discap	30 v 50 v 500 v 200 v 50 v	±0.5 pf
C72 C77 C77 C83 C84	283-0026-00 281-0536-00 283-0114-00 281-0521-00 281-0036-00	101 910 X138	909	0.2 μf 1000 pf 0.0015 μf 56 pf 3-12 pf, Var	Discap Cer Discap Cer Cer	25 v 500 v 200 v 500 v	10% 10%
C86 C100A,B,C C113 C123 C126	281-0524-00 Use 290-0089-00 281-0503-00 281-0517-00 281-0546-00			150 pf 3 x 20 μf 8 pf 39 pf 330 pf	Cer Elect. Cer Cer Cer	500 v 350 v 500 v 500 v 500 v	±0.5 pf 10% 10%

¹There are two parts of this description in your instrument.

Capacitors (Cont)

Ckt. No.	Tektronix Part No.	Serial/Model No. Eff Disc		Description			
C128 C131 C141 C142 C143	283-0004-00 283-0006-00 281-0525-00 283-0518-00 281-0536-00	X303		0.02 μf 0.02 μf 470 pf 330 pf 1000 pf	Discap Discap Cer Mica Mica	150 v 600 v 500 v 500 v 500 v	10% 10%
C144 C146 C151 C160 C164	281-0543-00 281-0523-00 281-0523-00 283-0000-00 283-0000-00			270 pf 100 pf 100 pf 0.001 μf 0.001 μf	Mica Mica Mica Discap Discap	500 v 350 v 350 v 500 v 500 v	10%
C166 C168 C170 C172 C183 C183	283-0005-00 283-0024-00 283-005-00 283-0024-00 283-0028-00 283-0001-00	X403 X403 101 1225	1224	0.01 μf 0.1 μf 0.01 μf 0.1 μf 0.0022 μf 0.005 μf	Discap Discap Discap Discap Discap Discap	250 v 30 v 250 v 30 v 50 v 500 v	
C187 C190 C196 C200A,B,C C211	283-0024-00 283-0002-00 283-0002-00 Use 290-0089-00 283-0006-00			0.1 μf 0.01 μf 0.01 μf 3 × 20 μf 0.02 μf	Discap Discap Discap Elect. Discap	30 v 500 v 500 v 350 v 600 v	
C213 C214 C214 C217 C217	283-0080-00 290-0020-00 290-0217-00 283-0006-00 283-0026-00	X743 101 743 101 743	742 742	0.022 μf 275 μf 250 μf 0.02 μf 0.2 μf	Discap Elect. Elect. Discap Discap	25 v 6 v 12 v 600 v 25 v	
C221 C222 C225 C226 C228	283-0013-00 283-0013-00 285-0590-00 285-0590-00 283-0036-00	101 101	742X 742X	0.01 µf 0.01 µf 0.22 µf 0.22 µf 0.025 µf	Discap Discap PTM PTM Discap	1000 v 1000 v 1000 v 1000 v 6000 v	
C229 C230 C231 C240A,B,C C241	281-0510-00 290-0149-00 290-0002-00 290-0115-00 283-0006-00	101 X743 X743 101	742X 742X	22 pf 5 μf 8 μf 3 × 20 μf 0.02 μf	Cer Elect. Elect. Elect. Cer	500 v 150 v 450 v 350 v 600 v	
C243 C244 C244 C246 C256	283-0080-00 290-0020-00 290-0217-00 283-0006-00 281-0504-00	X743 101 743	742	0.022 μf 275 μf 250 μf 0.02 μf 10 pf	Discap Elect. Elect. Discap Cer	25 v 6 v 12 v 600 v 500 v	10%
C262 C280 C312 C318 C320	281-0007-00 281-0518-00 Use *285-0672-00 Use 283-0057-00 Use 283-0057-00			3-12 pf, Var 47 pf 0.1 µf 0.1 µf 0.1 µf	Cer Cer PTM Discap Discap	500 v 600 v 200 v 200 v	

7-2
Capacitors (Cont)

Cht No	Tektronix Part No	Serial/M	odel No.		Desert		
CKI. 140.	1411 140.	L11	Disc		Descri	Stion	
C332 C333A,B,C C334A C334A	Use 290-0089-00 290-0115-00 283-0006-00 283-0022-00	101 580	579	3 × 20 μf 3 × 20 μf 0.02 μf 0.02 μf	Elect. Elect. Discap Discap	350 v 350 v 600 v 1400 v	
C334B	283-0006-00	101	579	0.02 µf	Discap	600 v	
C334B C336 C342A C342B C344	283-0022-00 283-000-00 283-0013-00 283-0013-00 283-0006-00	580		0.02 μf 0.001 μf 0.01 μf 0.01 μf 0.02 μf	Discap Discap Discap Discap Discap	1400 v 500 v 1000 v 1000 v 600 v	
C346 C352A,B C356 C357 C370	281-0529-00 283-0006-00 Use 283-0022-00 Use 283-0022-00 283-0002-00			1.5 pf 2 × 0.02 μf 0.02 μf 0.02 μf 0.01 μf	Cer Discap Cer Cer Discap	500 v 600 v 1400 v 1400 v 500 v	±0.25 p
C380 C382 C385 C399 C422	281-0542-00 283-0006-00 283-0006-00 283-0004-00 283-0006-00			18 pf 0.02 μf 0.02 μf 0.02 μf 0.02 μf	Cer Discap Discap Discap Discap	500 v 600 v 600 v 150 v 600 v	10%
C425 C430 C433 C435 C601	281-0007-00 283-0006-00 283-0006-00 283-0006-00 281-0559-00			3-12 pf, Var 0.02 μf 0.02 μf 0.02 μf 0.0015 μf	Cer Discap Discap Discap Cer	600 v 600 v 600 v 500 v	
C602 C613 C634 C646 C647	281-0559-00 Use 290-0017-00 285-0515-00 285-0515-00 Use 290-0006-00			0.0015 μf 125 μf 0.022 μf 0.022 μf 2 × 15 μf	Cer Elect. PTM PTM Elect.	500 v 450 v 400 v 400 v 350 v	
C650 C655A,B ² C661 C662 C663	290-0142-00 290-0022-00 Use 290-0017-00 Use 290-0017-00 Use 290-0017-00			4000 µf 2 x 1000 µf 125 µf 125 µf 125 µf	Elect. Elect. Elect. Elect. Elect.	50 v 15 v 450 v 450 v 450 v	
C680 C688 C698 C702 C703	285-0511-00 285-0510-00 285-0510-00 Use 290-0077-00 Use 290-0077-00			0.01 μf 0.01 μf 0.01 μf 2 × 100 μf 2 × 100 μf	PTM PTM PTM Elect. Elect.	600 v 400 v 400 v 350 v 250 v	
C718 C728	285-0515-00 285-0511-00			0.022 μf 0.01 μf	РТМ РТМ	400 v 600 v	

²Note: Cover not included. For cover order Tektronix Part No. 200-0093-00.

Capacitors (Cont)

	Tektronix	Serial/N	Nodel No.				
Ckt. No.	Part No.	Eff	Disc	Description			
C742	Une 200 0010 00			2 v 20 v f	Float	450	
C743	Use 295 0010-00			0.022 uf	PTAA	400 v	
C750	200 0117 00			50f	Flort	400 V	
C760	290-0117-00			01.4	Discarp.	20.4	
C70/	203-0024-00			0.1 /	Discup	200	
0//1	038 203-0037-00			0.1 µl	Discup	200 V	
C801	290-0002-00			8 µf	Elect.	450 v	
C802	281-0556-00			500 pf	Cer	10,000 v	
C803	283-0000-00			0.001 µf	Discap	500 v	
C807	285-0511-00			0.01 µf	PTM	600 v	
C808	285-0502-00	101	1001X	0.001 µf	PTM	1000 v	
C812	Use 283-0096-00			500 pf	Cer	20.000 v	
C822	Use 283-0096-00			500 nf	Cer	20,000 v	
C832	Lise *050-0223-00	101	589	Replacement Kit		20,000 1	
C832	283.0071.00	590	507	0.0048 uf	Cer	5000 v	
C833	1100 *050.0223.00	101	589	Replacement Kit	601	5000 4	
0000	030 030-0220-00	101	507	Replacement Ri			
C833	283-0071-00	590		0.0068 μf	Cer	5000 v	
C835	283-0037-00			500 pf	Discap	30,000 v	
C841	Use 283-0013-00			0.01 µf	Discap	1000 v	
C842	Use 283-0071-00			0.0068 µf	Cer	5000 v	
C857	283-0033-00			0.001 µf	Discap	6000 v	
C858	283-0021-00			0.001 <i>"</i> f	Discan	5000 v	
C862	283-0033-00			0.001 uf	Discap	6000 v	
C869	283-0001-00	X914		0.005 uf	Discap	500 v	
C8833	200 0001 00	70714		0.001 uf	Discap	500 v	
C888	285,0526,00			0.1f	PTAA	400 v	
0000	203-0320-00			0.1 <i>µ</i> 1	1 IM	400 1	
C895	283-0509-00			180 pf	Mica	500 v	10%
C896	283-0509-00			180 pf	Mica	500 v	10%
C916	Use 290-0143-00			20 µf	Elect.	500 v	
C919	281-0549-00	X482		68 pf	Cer	500 v	10%
C920A,B,C	*291-0031-00			0.1 × 0.01 × 0.001	μf Mylar	Timing Series	
C920D	281-0023-00			9-180 nf. Var	Mice		
COOLE	281 0012 00			7.45 of Var	Cor		
C930	281-0510-00	101	181	22 of	Cer	500 v	
C030	281.0508.00	492	401	12 of	Cer	500 v	5%
C020	201-0508-00	101	401 V	150 mf	Cer	500 v	5%
C730	201-0324-00	101	401A	130 pi	Cer	300 V	

Diodes

D50	Use *152-0203-00	101	589	Tunnel	Tek Spec	20 MA	4 pf
D50	152-0177-00	590		Tunnel	TD253B	10 MA	2 pf
D68	152-0026-00			Q6100			
D69	152-0026-00			Q6100			
D70	Use *152-0075-00			German	nium Tek Sp	bec	

³If replacement is necessary, order Tektronix Part No. 132-0055-00.

Diodes (Cont)

	Tektronix	Serial/Model	No.	Description
CKI. INO.	Full No.	EII	Disc	Description
D71 D72 D72 D73 D80 D81	Use *152-0075-00 Use *152-0075-00 *152-0185-00 *152-0185-00 152-0026-00 152-0026-00	101 967 X1220	966	Germanium Tek Spec Germanium Tek Spec Silicon Replaceable by 1N4152 Silicon Replaceable by 1N4152 Q6100
D82 D144 D160 D161 D180 D197	152-0026-00 152-0026-00 152-0008-00 152-0008-00 *152-0185-00 152-0025-00	X1220 X105	1224X	Q6100 Q6100 T12G T12G Silicon Replaceable by 1N4152 1N634
D198 D199 D214 D215 D220	152-0025-00 152-0025-00 *152-0233-00 *152-0233-00 152-0039-00	X743 X743 101	149 249	1N634 1N634 Silicon Selected from 1N3606 Silicon Selected from 1N3606 Zener 1.5 w 110 v
D220 D220 D221 D226 D227	*153.0009-00 152.0150-00 152.0150-00 *152-0233-00 *152-0233-00	250 743 X743 X759 X759	742	Zener 1.5 w Selected Zener 1N3037B 1 w 51 v 59 Zener 1N3037B 1 w 51 v 59 Silicon Selected from 1N3606 Silicon Selected from 1N3606
D305 D305 D305 ⁴ D306 D306 ⁴	152-0037-00 152-0038-00 *153-0010-00 152-0038-00 *153-0010-00	101 235 250 101 250	234 249 249	Zener 1.5 w 200 v Zener 1.5 w 140 v Zener 1.5 w Selected Zener 1.5 w 140 v Zener 1.5 w Selected Zener 1.5 w Selected
D307 D307 D307* D308 D308* D318	152-0037-00 152-0038-00 *153-0010-00 152-0039-00 *153-0010-00 Use *152-0075-00	101 235 250 X235 250	234 249 249	Zener 1.5 w 200 v Zener 1.5 w 140 v Zener 1.5 selected Zener 1.5 w 110 v Zener 1.5 w Selected Germanium Tek Spec
D344 D345 D344 D345 D345 D384 D431 D431	Use *153-0026-00 *153-0026-01 152-0005-00 *153-0004-00	101 1280	1279	Zener 1.5 w 140 v Zener 1.5 w 110 v Zener 1 w 120 v Checked pair TI3G Zener 1.5 w 200 v Zener 1.5 w 140 v
D610 D611 D612 D613 D650	152-0040-00 152-0040-00 152-0040-00 152-0040-00 152-0036-00	101	476	1N2615 1N2615 1N2615 1N2615 1N2615 1N1582C

*Furnished as a unit. Selected for total voltage drop of 500 to 550 v.

Diodes (Cont)

	Tektronix	Serial/Model	No.	D i.i.
Ckt. No.	Part No.	ETT	Disc	Description
D650 D651 D651 D655 D655	152-0088-00 152-0036-00 152-0088-00 152-0047-00 152-0066-00	477 101 477 101 870	476 869	Silicon 1N3209 1N1582C Silicon 1N3209 Silicon 1N3262 (or equal) Silicon 1N3194
D656 D656 D660 D661 D662	152-0047-00 152-0066-00 152-0040-00 152-0040-00 152-0040-00	101 870	869	Silicon 1N/2862 (or equal) Silicon 1N3194 1N2615 1N2615 1N2615
D663 D700 D700 D701 D701	152-0040-00 152-0047-00 152-0066-00 152-0047-00 152-0066-00	101 870 101 870	869 869	1N2615 Silicon 1N2862 (or equal) Silicon 1N3194 Silicon 1N3194 Silicon 1N3194
D702 D702 D703 D703 D740	152-0047-00 152-0066-00 152-0047-00 152-0066-00 152-0047-00	101 870 101 870 101	869 869 869	Silicon 1N2862 (or equal) Silicon 1N2194 Silicon 1N2862 (or equal) Silicon 1N2194 Silicon 1N2862 (or equal)
D740 D741 D741	152-0066-00 152-0047-00 152-0066-00	870 101 870	869	Silicon 1N3194 Silicon 1N2862 (or equal) Silicon 1N3194
			Fuses	
F601 F601 F606	159-0036-00 159-0028-00 159-0019-00	X323		7 Amp 3 AG Slo-Blo 117 V oper. 50 & 60 cycle 4 Amp 3 AG Slo-Blo 234 V oper. 50 & 60 cycle 1 Amp 3 AG Slo-Blo
F650 F651 F801	159-0014-00 159-0038-00 159-0027-00	X403 X522		5 Amp 3 AG Fast-Blo 15 Amp 3 AG Fast-Blo 0.25 Amp 3 AG Fast-Blo
			Jack	
J170	131-0267-00	X403		3 conductor phone jack
			Relays	8
K601 K602 K603	148-0006-00 148-0005-00 Use 148-0020-00			Thermal, Time Delay 26NO45T 32 v 32 v

Electrical Parts List—Type 519

Inductors

	Tektronix	Serial/Model	No.	
Ckt. No.	Part No.	Eff	Disc	Description
L30 ⁵	*108-0170-00			0.5 μh
L375	*108-0215-00			1.1 μh
L50	315-0121-00			Formed from the leads of R50
L52	*108-0182-00			0.3 <i>u</i> h
L53	*108-0182-00			0.3 μh
L184	*108-0216-00			4 Section Grid Line
L192	*108-0129-00			18 ub
1193	*108-0129-00			18 //h
1194	*108-0217-00			2 Section Plate Line
1250	*108-0218-00			5 Section Grid Line
1051	*100 0005 00	¥105		0.05
L251	*108-0235-00	X105	161X	0.05 µh
1312	*108-0219-00			3 Section Plate Line
L336	108-0225-00	X162		0.75 mh, 3 section
L375	*108-0022-00			88 µh
L860°				Beam Rotator
L885	*108-0222-00			Reed Drive
LR936	*108-0223-00	101	481	0.06 μh
LR936	*108-0266-00	482		4 turns on 27 Ω Resistor
			Delay	Line
				100 -
	*119-0011-00			12512
			Iransis	fors
0345	151-0027-00			201700
0445	151-0027-00			2N700
070	Lise 151-0107-00	101	909	2NI967
070	*151-0083-00	910	/0/	Selected from 2NI964
0140	151.0015.00	10		OC170
9100	151-0015-00			Cellis
0190	11 151 0107 00	101	000	010/7
0100	Use 151-0107-00	010	909	21N70/
	*151-0083-00	910		Selected from 21964
Q214	*151-0108-00	X/43		Replaceable by 2N2501
Q215	*151-0133-00	X/43		Selected from 2N3251
Q238	Use 151-013/-00			2N2148
Q318	151-0044-00			2N705
Q328	Use 151-0137-00			2N2148
Q766A	151-0007-00			2N270
Q766B	151-0007-00			2N270
Q767	151-0002-00			2N277
Q773	Use 151-0137-00			2N2148
Q777	151-0002-00	101	402	2N277
Q777	151-0102-00	403		MP504
Q934	Use *050-0140-00	101	481	Replacement Kit
Q934	*153-0523-00	482		Tek Spec, Checked

⁵There are two parts of this description in your instrument.

⁶Furnished as a unit with CRT.

			Resisto	rs		
Ckt. No.	Tektronix Part No.	Serial/Model Eff	No. Disc		Descripti	on
Resistors are	fixed, composition,	$\pm 10\%$ unless othe	rwise indicat	ed.		
R6 R7 R8 R9 R10	Use 302-0680-00 Use 302-0101-00 302-0220-00 302-0220-00 301-0241-00			68 Ω 100 Ω 22 Ω 22 Ω 240 Ω	½ ₩ ½ ₩ ½ ₩ ½ ₩ ½ ₩	5%
R11 R13 R14 R15 R16	301-0241-00 301-0241-00 301-0241-00 302-0100-00 301-0241-00			240 Ω 240 Ω 240 Ω 10 Ω 240 Ω	½ ₩ ½2 ₩ ½2 ₩ ½2 ₩	5% 5% 5% 5%
R17 R18 R19 R21 R22	301-0241-00 301-0241-00 301-0241-00 304-0820-00 301-0510-00			240 Ω 240 Ω 240 Ω 82 Ω 51 Ω	½ ₩ ½ ₩ ½ ₩ 1 ₩ ½ ₩	5% 5% 5% 5%
R23 R26 R27 R28 R30 ⁷	316-0820-00 301-0750-00 301-0680-00 315-0750-00 315-0181-00			82 Ω 75 Ω 68 Ω 75 Ω 180 Ω	1/4 w 1/2 w 1/2 w 1/4 w 1/4 w	5% 5% 5%
R32 [†] R33 [†] R36 [†] R37 [†] R43 [†]	315-0271-00 315-0820-00 315-0122-00 315-0122-00 301-0242-00			270 Ω 82 Ω 1.2 k 1.2 k 2.4 k	1/4 w 1/4 w 1/4 w 1/4 w 1/4 w 1/2 w	5% 5% 5% 5% 5%
R44 ⁷ R45 ⁷ R46 ⁷ R47 ⁷ R50	315-0270-00 315-0272-00 301-0222-00 315-0470-00 315-0121-00			27 Ω 2.7 k 2.2 k 47 Ω 120 Ω	1/4 w 1/4 w 1/2 w 1/4 w 1/4 w	5% 5% 5% 5% 5%
R51 R51 R53 R54 R55	315-0121-00 315-0181-00 315-0110-00 315-0110-00	101 590	589	120 Ω 180 Ω Selected 11 Ω 11 Ω	1/4 w 1/4 w (nominal value) 1/4 w 1/4 w	5% 5% 5%
R56 R57 R58 R60 R61	308-0067-00 316-0222-00 316-0223-00 302-0275-00 302-0685-00			750 Ω 2.2 k 22 k 2.7 meg 6.8 meg	5 w 1/4 w 1/4 w 1/2 w 1/2 w	WW 5%

⁷There are two parts of this description in your instrument.

	Tektronix	Serial/Ma	odel No.		- · · ·		
Ckt. No.	Part No.	Eff	Disc		Descrip	otion	
R63	301-0224-00			229 k	1/2 W		5%
R64	301-0182-00			1.8 k	1/2 w		5%
R66A)	311-0215-00			500 k, Var			
R66B)	311 0018 00			1 k, Var 20 k Var			
107	011-0010-00			20 K, 101			
P40	201 0121 00			120.0	1/		E 0/
R70	301-0241-00			240.0	1/2 W		5%
R71	301-0332-00			334	1/2 W		5%
R72	301-0561-00			560 0	1/2 W		5%
R75	302-0471-00			470 Ω	½ w		• //
R76	311-0480-00	X910		500 Q. Var			
R77	301-0102-00			1 k	1/2 W		5%
R78	301-0681-00	101	909	680 Ω	1/2 w		5%
R78	302-0471-00	910		470 Ω	1/2 W		- /
R79	315-0220-00	X1220		22 Ω	1⁄4 w		5%
R81	302-0221-00			220 Ω	1∕₂ w		
R82	316.0820.00			82 0	1/		
R83	309-0159-00			5 k	1/4 W	Prec	1%
R84	316-0270-00	X138		27 Ω	1/. w		• /6
R85	310-0115-00			15 k	1 w	Prec	1%
R86	309-0273-00			866 Ω	½ ₩	Prec	1%
R87	310-0132-00			19.6 k	lw	Prec	1%
R88	311-0021-00			30 k, Var			
R100	302-0101-00			100 Ω	1/2 W		
R101	302-0101-00			100 Ω	1/2 w		
R102	302-0101-00			100 Ω	½ w		
R113	301-0302-00			3 k	½ ₩		5%
R114	301-0272-00			2.7 k	½ w		5%
R116	308-0178-00			15 k	8 w	ww	5%
R120	302-0470-00			47 Ω	½ w		
R122	305-0104-00			100 k	2 w		5%
R124	301-0913-00			91 k	½ w		5%
R125	301-0304-00			300 k	1/2 w		5%
R126A	311-0078-00			50 k, Var			
R126B	301-0154-00			150 k	½ w		5%
R126C	301-0394-00			390 k	1∕2 w		5%
R126D	301-0105-00			1 meg	½ w		5%
R126E	301-0275-00			2.7 meg	½ w		5%
R126F	301-0565-00			5.6 meg	1/2 W		5%
R128	302-0333-00			33 k	1/2 W		5.01
K130	305-0622-00			6.2 K	2 w		5%
R131	302-0822-00			8.2 k	½ w		
K132	302-0101-00			100 Ω	1/2 W		5.0
K134	305-0682-00			6.8 K	2 w	Dese	5%
K136	309-0036-00			10 K	1/2 W	rrec	1%
K13/	309-0228-00			12.5 K	1/2 W	rrec	1%

	Tektronix	Serial/M	odel No.			
Ckt. No.	Part No.	Eff	Disc		Description	
R142	301-0102-00			1 k	1/2 w	5%
R143	301-0562-00			5.6 k	1/2 W	5%
R144	301-0221-00			220 Ω	1/2 w	5%
R145	302-0471-00			470 Ω	1/2 W	• /6
R146	302-0153-00			15 k	1/2 W	
	002 0100 00			15 K	/2 **	
P147	202 01 52 00			151	17	
D1 40	201 0121 00			190.0	1/2 W	50/
D161	201-0101-00			100 12	72 W	5%
DIEA	201-0221-00			220 12	72 W	5%
R134	301-04/2-00			4./ K	/2 W	5%
K155	304-0003-00			08 K	I w	
P156	201 0242 00	101	1010	246	17	50/
P156	301 0392 00	1000	1217	201	/2 W	5%
P120	202 01 54 00	1220		3.7 K	/2 W	5%
D1/1	302-0134-00			150 K	√2 W	
RIOI D1/O	301-04/3-00			4/ K	1/2 W	5%
K103	302-0821-00			820 12	1/2 W	
R164	302-0334-00			330 k	1/2 W	
D1/6	200 0102 00	¥020		101		
KI00	302-0163-00	7830		18 K	1/2 W	
K166	302-0103-00			10 k	1/2 W	
R16/	302-0153-00			15 k	1/2 W	
R168	302-04/1-00			470 Ω	1/2 w	
R169	302-0103-00	X403		10 k	½ w	
P170	202 01 52 00	× 100		161	14	
D171	202-0103-00	X403		101	72 W	
R171	302-0183-00	X400		10 K	72 W	
R1/2 D101	300-04/1-00	7403		4/0 12	2 W	
K181	301-02/1-00			2/0 \	% ₩	5%
K182	307-0055-00			3.9 Ω	1∕2 w	5%
P193	301-0681-00			690.0	1/	E 9/
D104	204 0221 00			220 0	72 W	5%
N104	300-0331-00			330 12	2 W	
K185	302-0330-00			33 12	√2 W	
K186	315-0101-00			100 \	1/4 W	5%
K18/	305-0101-00			100 12	2 w	5%
K188	301-04/0-00			47 Ω	1∕2 ₩	5%
D100	000 0001 00			000 0		
K190	302-0221-00			220 11	1/2 W	
K192	301-0221-00			220 \	1/2 W	5%
R194	301-0221-00			220 Ω	1/2 w	5%
R199	301-0101-00			100 Ω	1∕2 w	5%
R210	302-0101-00			100 Ω	½ w	
R211	302-0100-00			10 Ω	½ w	
R214	302-0100-00			10 Ω	½ w	
R215	301-0471-00	X940		470 Ω	1/2 w	5%
R216	302-0102-00			1 k	1/2 W	
R217	302-0270-00	101	742X	27 Ω	1∕2 w	
R218	302-0220-00	101	742X	22 <u>Ω</u>	½ w	
R220	302-0473-00			47 k	1/2 W	
R221	316-0150-00	101	742X	15 Ω	1/4 w	
R222	316-0150-00	101	758X	15 Ω	1/4 w	
R223	302-0100-00	101	742X	10 Ω	1/2 W	

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	Tektronix	Serial/N	Nodel No.				
CKI. 140.	Full No.	511	Disc		Desci	ription	
R224	302-0100-00	101	742X	10 Ω	1/2 W		
R225	302-0471-00	101	742X	470 Ω	1/2 w		
R226	302-0101-00			100 Ω	1/2 W		
R227	302-0102-00			1 k	1/2 W		
R228	302-0101-00	101	742X	100 Ω	½ ₩		
R229	316-0470-00	101	742	47 Ω	1/4 w		
R229	301-0301-00	743		300 Ω	1/2 W	Selected (nominal value)	5%
R230	305-0563-00	101	742	56 k	2 w		5%
R230	308-0213-00	743		25 k	7 w	ww	5%
R231	301-0911-00			910 Ω	½ w		5%
R234	304-0124-00			120 k	1 w		
R236	304-0154-00	101	742X	150 k	1 w		
R237	309-0014-00	101	742X	1 mea	1/2 W	Prec	1%
R238	Use 309-0015-00	101	229	1.5 meg	1/2 W	Prec	1%
R238	309-0015-00	230	742X	1.11 meg	½ ₩	Prec	1%
R239	315-0221-00	X743		220 Ω	1/4 w		5%
R240	315-0100-00	X743		10 Ω	1/4 w		5%
R241	302-0100-00	101	742X	10 Ω	½ w		
R242	306-0333-00	X105		33 k	2 w		
R243	305-0153-00	101	104	15 k	2 w		5%
P242	204 0272 00	105		07 h	0		
R243	300-02/3-00	105		2/ K	,2 w	Colored Level and a Level	5.00
R244	301-0390-00	X/43		3911	י∕₂ w	Selected (nominal value)	5%
R240	302-0100-00			1012	1/2 ₩		
R24/	302-0101-00			100 Ω	1/2 W		
KZ3U	301-0101-00			100 12	7₂ ₩		5%
R256	301-0222-00			2.2 k	1/2 w		5%
R257	304-0124-00			120 k	î.w		- /-
R260	302-0220-00			22 Ω	1/2 W		
R262	301-0563-00			56 k	1/2 w		5%
R263	301-0433-00			43 k	1/2 w		5%
R265	302-0101-00			100 Ω	½ ₩		
R270A	301-0204-00			200 k	½ w		5%
R270C	301-0394-00			390 k	½ w		5%
R270E	301-0624-00			620 k	½ ₩		5%
R270G	301-0165-00			1.6 meg	½ w		5%
R270J	301-0245-00			2.4 meg	1/4 w		5%
R271	306-0104-00			100 k	2.		J /0
R272	316-0470-00			47 0	1/. w		
R274	308-0127-00			3.5 k	5	ww	5%
R275	308-0091-00			2 k	5 w	ww	5%
R280	301-0203-00			20 k	½ w		5%
R281	309-0091-00			120 k	½ w	Prec	1%
R282	302-0470-00			47 Ω	½ ₩		
R284	Use 306-0473-00	101	239	220 k	½ w		5%
R284	306-0473-00	240		47 k	2 w		

a	Tektronix	Serial/M	odel No.				
Ckt. No.	Part No.	Eff	Disc		Descrip	tion	
P285	316.0470.00			17.0	1/		
P284	214 0470 00			47.0	74 W		
R287	304-0222-00			224	74 **		
P320	301 0491 00			490.0	1/		E o/
P321	202 0142 00	101	1070	141	72 W		5%
K321	303-0162-00	101	10/9	1.0 K	l w		5%
R321	303-0132-00	1080		1.3 k	1 w		5%
R331	302-0100-00			10 Ω	1/2 W		
R332	308-0105-00			30 k	8 w	WW	5%
R333	304-0562-00			5.6 k	1 w		- 70
R334	307-0023-00			4.7 Ω	1/2 W		
R336A	304-0100-00	101	161	10.0	1 w		
R336A	*308-0193-00	162	277	184	25 w	ww	
R336A	*308-0214-00	278	2	164	25 w	ww	
R336C	*310-0579-00	101	161	64	3	14/14/	19/
R336C	*308-0194-00	162	277	5 k	20 w	ww	1 /0
R336C	*308-0215-00	278		15k	20.54	14/14/	
P336D	*310.0579.00	101	1417	4.J K	20 w	NACIA/	1.0/
P224E	*210 0590 00	101	141	144	4 W		1%
D222E	*210 0595 00	140	077	10 K	4 W	VV VV	1%
P224E	*210.0500.00	070	1004	20 K	4 W	VV VV	1%
ROODE	-310-0390-00	2/0	1224	18 K	4 W	VV VV	1%
R336E	*310-0590-01	1225		18 k	8 w	ww	1%
R336F	*310-0580-00	101	161	16 k	4 w	ww	1%
R336F	*310-0585-00	162	277	20 k	4 w	ww	1%
R336F	*310-0590-00	278	1224	18 k	4 w	WW	1%
R336F	*310-0590-01	1225		18 k	8 w	ww	1%
R336G	*310-0578-00	101	161	17 k	4 w	ww	1%
R336G	*310-0585-00	162	277	20 k	4 w	WW	1%
R336G	*310-0590-00	278	1224	18 k	4 w	ww	1%
R336G	*310-0590-01	1225		18 k	8 w	ww	1%
R336J	*310-0581-00	101	277	46 k	4 w	WW	1%
R336J	*310-0589-00	278		43 k	4 w	ww	1%
R336L	310-0122-00	101	277	182 k	1 w	Prec	1 %
R336L	310-0142-00	278		172 k	1 w	Prec	1 %
R336M	310-0122-00	101	277	182 k	1 w	Prec	1 %
R336M	310-0142-00	278	2.77	172 k	1	Prec	1%
R336N	310.0122.00	101	277	182 4	1	Proc	1 %
R336N	310-0142-00	278	2//	172 k	1 w	Prec	1%
R336Q	310-0123-00	101	277	453 k	1.w	Prec	1 %
R3360	310-01/3-00	278	2//	432 L	1	Prec	1 %
23345	310 0097 00	101	977	402 K	1	Prec	1 %
P224C	210 01 44 00	070	2//	900 K	1	Prec	1%
D227	200 00/5 00	101	1/17	000 K	05	Prec	1%
R338	308-0108-00	101	161X	15 k	25 W 5 w	ww	5% 5%
P340	201 0101 00			100 0	1/		E ^/
D241	201 0101-00			100 12	1/2 W		5%
P3/2	302 0101-00			100 1	1/2 W		5%
D244	200 0155 00			100 12	72 W	Deres	1.01
P2/7	200 0155 00			40 k	72 W	r rec	1%
P250	214 0641 00			40 K	72 W	rrec	1%
K330	310-0301-00			200 11	% W		

Cat. For. Pair For. Disc Description 2351 316.0561.00 77 15 k 1 w 2352 304.0153.00 101 277 15 k 1 w 2352 304.0153.00 101 277 56 k 1 w 2352 304.0153.00 101 277 56 k 1 w 2353 304.0553.00 101 277 56 k 1 w 2357 304.0553.00 278 51 k 1 w 5% 2357 304.0553.00 278 51 k 1 w 5% 2357 304.0554.00 101 100 Ω 1/4 w 5% 2370 309.052.00 101 104 390 k 1/2 w 5% 2371 309.052.00 101 104 200 k 1/2 w 5% 2371 309.052.00 101 104 200 k 1/2 w 5% 2372 316.010.00 50 L 239 200 k 1/2 w 5%<		Tektronix	Serial/M	odel No.		Decede		
R351 316-051.00 27 50 Ω Y_{1} w R352 304-013.00 10 277 15 k 1 w R353 304-053.00 101 277 56 k 1 w R354 304-053.00 278 51 k 1 w 5% R355 303-0513.00 278 51 k 1 w 5% R357 304-0553.00 101 277 56 k 1 w 5% R357 304-0553.00 101 278 51 k 1 w 5% R351 316-011-00 100 Ω Y_{4} w 5% 7% R364 308-0051-00 105 239 220 k Y_{4} w 5% R371 301-0204-00 105 239 220 k Y_{4} w 5% R372 316-0101-00 100 Ω Y_{4} w 5% 530 1 w 5% R373 301-0204-00 240 270 Ω Y_{4} w 5% 5% R373 301-0204-00 200 k Y_{2} w 5% 5% 5% 5%	CKT. INO.	Parr No.	EIT	Disc		Descrip	ofion	
R352 304.0103.00 101 277 15 k 1 w R352 304.0103.00 278 10 k 1 w 5% R355 303.0513.00 278 51 k 1 w 5% R357 304.0553.00 101 277 56 k 1 w 5% R357 303.0513.00 278 51 k 1 w 5% R357 304.0553.00 101 100 Ω 1/2 w 5% R357 304.0553.00 101 100 Ω 1/2 w 5% R357 303.0513.00 278 51 k 1 w 5% R357 304.0554.00 101 100 Ω 1/2 w 5% R371 309.052.00 101 104 390 k 1/2 w 5% R371 301.052.00 101 104 390 k 1/2 w 5% R373 301.0561.00 100 Ω 1/2 w 5% 5% R373 301.0561.00 50 L 1/2 w 5% 5% R374 311.0798.00 X105 50 L 1/2 w </td <td>R351</td> <td>316-0561-00</td> <td></td> <td></td> <td>560 Ω</td> <td>14 w</td> <td></td> <td></td>	R351	316-0561-00			560 Ω	14 w		
R352 304.0103.00 278 10 k 1 w R356 304.0553.00 101 277 56 k 1 w R356 303.0513.00 278 51 k 1 w 5% R357 304.0553.00 278 51 k 1 w 5% R357 304.0513.00 278 51 k 1 w 5% R361 316.0101.00 100 C 1/2 w 5% 100 C 1/2 w R361 316.0101.00 105 390 k 1/2 w 5% 1% R370 301.0394.00 105 390 k 1/2 w 5% 877 R371 309.0322.00 101 104 220 k 1/2 w 5% R371 301.0204.00 240 200 k 1/2 w 5% R372 316.0101.00 100 Ω 1/2 w 5% R373 301.0204.00 200 k 1/2 w 5% R373 301.0204.00 200 k 1/2 w 5% R373 301.0204.00 200 k 1/2 w 5% R380	R352	304-0153-00	101	277	15 k	1 w		
R356 304.0563.00 101 277 56 k 1 w 5% R357 304.0563.00 101 277 56 k 1 w 5% R357 304.0563.00 278 51 k 1 w 5% R357 303.0513.00 278 51 k 1 w 5% R357 303.0513.00 278 51 k 1 w 5% R361 316.011.00 100 Ω 1/2 w 5% 7% R364 308.0051.00 4 k 5 w WW 5% R377 301.0374.00 105 104 399 k $1/2 w$ 7% 5% R371 301.0244.00 105 239 200 k $1/2 w$ 5% R372 316.0101.00 100 Ω 1/4 w 5% 5% 5% R373 301.0244.00 1405 50 Ω 1/2 w 5% 5% R373 301.0244.00 1405 50 L, Vor 0.1 w 5% 5% R374 311.078.00 1105 50 L, Va w 5% 5%	R352	304-0103-00	278		10 k	1 w		
R356 303-0513-00 278 51 k 1 w 5% R357 304-0563-00 101 277 56 k 1 w 5% R357 303-0513-00 278 51 k 1 w 5% R360 316-0101-00 100 Ω V_{4} w 5% R361 316-0101-00 100 Ω V_{4} w 5% R370 307-0056-00 101 104 370 k V_{2} w Prec 1% R371 307-0052-00 101 104 370 k V_{2} w Prec 1% R371 307-0052-00 101 104 220 k V_{2} w 5% R371 301-0204-00 240 200 k V_{2} w 5% R372 310-0511-00 270 L V_{2} w 5% R375 301-0201-00 270 L V_{2} w 5% R375 302-0271-00 91 Ω V_{2} w 5% R380 301-0270-00 20 k V_{2} w	R356	304-0563-00	101	277	56 k	1 w		
R357 304.0563.00 101 277 56 k 1 w 5% R357 303.0513.00 278 51 k 1 w 5% R361 316.0101.00 100 Ω 1/4 w 5% R361 316.0101.00 100 Ω 1/4 w 5% R371 309.0554.00 101 104 390 k 1/2 w 5% R371 309.016294.00 105 390 k 1/2 w 7% 5% R371 309.0224.00 105 200 k 1/2 w 5% R371 309.0224.00 105 200 k 1/2 w 5% R373 301.0244.00 125 239 200 k 1/2 w 5% R373 301.0244.00 240 200 k 1/2 w 5% R374 311.028.00 X105 500 Ω 1/2 w 5% R380 301.0244.00 470 k 1/2 w 5% R381 301.0203.00 20 k 1/2 w 5% R383 301.0474.00 470 k 1/2 w 5% R38	R356	303-0513-00	278		51 k	i w		5%
R357 304.0563.00 101 277 56 k 1 w 57, R357 303.013.00 278 51 k 1 w 57, R361 316.0101.00 100 Ω V_{1} w 57, R361 316.0101.00 100 Ω V_{1} w 57, R370 309.0056.00 101 104 390 k V_{2} w 57, R370 301.0394.00 105 390 k V_{2} w 97, 397, R371 309.052.00 101 104 220 k V_{2} w 5%, R371 301.0204.00 105 239 220 k V_{2} w 5%, R372 316.0101.00 100 Ω V_{4} w 5%, 837, R373 301.0204.00 100 Ω V_{4} w 5%, 836, R373 301.0204.00 20 k V_{4} w 5%, 837, R380 301.0204.00 20 k V_{4} w 5%, R383 301.0204.00 <td< td=""><td></td><td></td><td></td><td></td><td></td><td></td><td></td><td>0 /0</td></td<>								0 /0
R357 303.0513.00 278 51 k 1 w 5% R360 316.0101.00 100 0 1/0 w 5% R361 316.0101.00 100 0 1/0 w 5% R361 316.0101.00 101 100 0 1/0 w 5% R370 309.0055.00 101 104 370 k 1/2 w 5% R371 309.0052.00 101 104 220 k 1/2 w 5% R371 309.0052.00 101 104 220 k 1/2 w 5% R371 301.0204.00 240 200 k 1/2 w 5% 5% R372 316.0101.00 100 0 1/2 w 5% 5% R374 311.005.00 500 k, Vor 0.1 w 5% R380 301.4070.40 91 0 1/2 w 5% R381 301.4070.40 40 k 1/2 w 5% R383 301.4070.40 40 k 1/2 w 5% R384 315.0662.00 6.8 k 1/2 w 5% R385 302.0101.00 <t< td=""><td>R357</td><td>304-0563-00</td><td>101</td><td>277</td><td>56 k</td><td>1 w</td><td></td><td></td></t<>	R357	304-0563-00	101	277	56 k	1 w		
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	R357	303-0513-00	278		51 k	1 w		5%
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	R360	316-0101-00			100 Ω	14 w		- /0
R364 308-0051-00 4 k 5 w WW 5% R370 301-0394-00 105 390 k ½ w 5% R371 301-0394-00 105 390 k ½ w 5% R371 301-0204-00 105 239 220 k ½ w 5% R371 301-0204-00 105 239 220 k ½ w 5% R371 301-0204-00 105 239 220 k ½ w 5% R372 301-0204-00 240 200 C 100 G ½ w 5% R373 301-0204-00 X105 500 G ½ w 5% R380 301-0203-00 91 G ½ w 5% R381 301-0203-00 20 k ½ w 5% R383 301-024-400 470 k ½ w 5% R384 315-0682-00 6.8 k ¼ w 5% R385 304-010-00 100 G ½ w 5% R385 304-010-00	R361	316-0101-00			100 Ω	14 w		
R370 309-0056-00 101 104 390 k Y_{y} w Prec 17, R371 309-0052-00 101 104 220 k Y_{y} w Prec 17, R371 309-0052-00 101 104 220 k Y_{y} w 5%, R371 301-0204-00 105 239 220 k Y_{y} w 5%, R372 316-0101-00 100 0 Y_{y} w 5%, 87, R373 301-0561-00 500 0, V_{y} w 5%, 87, R374 311-0078-00 X105 504, V_{w} w 5%, R380 301-020-00 270 0 Y_{y} w 5%, R381 301-020-00 20 k Y_{y} w 5%, R383 310-020-00 40 k Y_{w} w 5%, R384 315-0482-00 64 k Y_{w} w 5%, R385 304-010-00 100 0 Y_{w} w 5%, R396 311-0005-00 1 meg	R364	308-0051-00			4 k	5 w	ww	5%
3370 3374,3374,337,337,8 γ_{2} w	0070	200.005/.00	101	104	2001			
	K3/U	309-0056-00	101	104	390 K	1/2 W	Prec	1%
R371 Us 00-4052-00 101 104 201k Y ₂ w Prec 1% R371 Us 301-2024-00 120 239 220 k Y ₂ w 5% R371 Us 301-2024-00 240 239 220 k Y ₂ w 5% R371 Us 301-2024-00 240 239 220 k Y ₂ w 5% R372 301-621-00 560 Ω Y ₂ w 5% 57% 57% R375 332-2271, 00 270 L Y ₂ w 5% 5% R380 301-0710-00 270 L Y ₂ w 5% R381 301-0202-00 470 k Y ₂ w 5% R383 310-0262-00 68 k Y ₂ w 5% R384 315.062-00 100 Ω Y ₂ w 5% R397 304-0150-00 1 meg 1 w 5% R397 301-0562-00 56 k 2 w 5% R400 302-0101-00 100 Ω<	K3/U	301-0394-00	105		390 k	1/2 W		5%
R371 Use 301-420-430 105 2.29 200 k V/2 w 5% R371 301-620-400 240 200 k V/2 w 5% R373 301-051-00 500 Ω V/2 w 5% R374 311-0078-00 1105 500 Ω V/2 w 5% R375 302-0271-00 200 k V/2 w 5% R380 301-0274-00 200 k V/2 w 5% R381 301-0203-00 20 k V/2 w 5% R381 301-0203-00 20 k V/2 w 5% R381 301-0203-00 20 k V/2 w 5% R383 301-024-00 470 k V/2 w 5% R384 315-0682-00 6.8 k V/2 w 5% R385 304-010-500 1 meg 1 w 5% R397 301-0551-00 5.6 k 2 w 5% R399 302-010-00 100 Ω V/2 w 5% R401 <td< td=""><td>K3/1</td><td>309-0052-00</td><td>101</td><td>104</td><td>220 k</td><td>1/2 ₩</td><td>Prec</td><td>1%</td></td<>	K3/1	309-0052-00	101	104	220 k	1/2 ₩	Prec	1%
R371 301-3204-30 240 200 k Y_2 w $5\gamma_4$ R272 316-6010-00 100 Ω Y_4 w $5\gamma_4$ R273 301-651.00 560 Ω Y_4 w $5\gamma_6$ R274 311-0078.00 X105 500 Ω Y_4 w $5\gamma_6$ R275 302.6271.00 270 Ω Y_2 w $5\gamma_6$ R380 301-6710.00 270 Ω Y_2 w $5\gamma_6$ R381 301-0203.00 20 k Y_2 w $5\gamma_6$ R383 301-074.00 470 k Y_2 w $5\gamma_6$ R383 301-074.00 470 k Y_2 w $5\gamma_6$ R384 315.062.00 6.8 k Y_2 w $5\gamma_6$ R390 306-053.00 1 meg 1 w 8396 831-0005.00 1 meg 1 w 8396 $301-056.00$ 560 Ω Y_2 w $5\gamma_6$ R397 301-0561.00 500 Ω, Vor 8397 $301-0562.00$ $56 k$ $2 w$ R400	R3/1	Use 301-0204-00	105	239	220 k	1/2 W		5%
R372 316-0101-00 100 Ω V_{2} w 5% R373 301-0561-00 560 Ω V_{2} w 5% R374 311-0078-00 X105 500 k, Var 0.1 w R375 302-02071-00 270 Ω V_{2} w 5% R380 301-0710-00 71 Ω V_{2} w 5% R381 301-0203-00 20 k V_{2} w 5% R383 301-0474-00 470 k V_{2} w 5% R384 315-0682-00 6.8 k V_{4} w 5% R395 302-0110-00 100 Ω V_{2} w 5% R395 304-0105-00 1 meg 1 w 5% R396 301-0550-0 5.6 k 2 w 5% R397 301-0550-0 5.6 k Y_{2} w 5% R401 302-00 5.6 k Y_{2} w 5% R402 316-0101-00 100 Ω Y_{2} w 5% R402 316-0101-00 100 Ω	K3/1	301-0204-00	240		200 k	% ₩		5%
8273 301-0561-00 560 Ω $\sqrt{2}$ w 5% 8274 311-0278-00 X105 50 k, Vor 0.1 w 87 8280 301-0910-00 91 Ω $\sqrt{2}$ w 5% 87 8380 301-0203-00 20 k $\sqrt{2}$ w 5% 8381 301-0203-00 20 k $\sqrt{2}$ w 5% 8383 301-0474-00 470 k $\sqrt{2}$ w 5% 8384 315-0682-00 6.8 k $\sqrt{4}$ w 5% 8385 302-010-00 100 Ω $\sqrt{2}$ w 5% 8390 302-010-00 1 meg 1 w 5% 8395 304-010-50 500 Ω $\sqrt{4}$ w 5% 8397 301-055-00 500 Ω $\sqrt{4}$ w 5% 8397 301-055-00 1 meg $\sqrt{2}$ w 5% 8397 302-010-00 100 Ω $\sqrt{2}$ w 5% 8401 302-2010-00 100 Ω $\sqrt{2}$ w 5% 8402 302-010-00<	R372	316-0101-00			100 Ω	14 w		
8274 311.0078.00 X105 50 k, Vor 0_1^2 w X_1^2 8375 302.0271.00 270 Ω Y_2 w S_2^2 8380 301.0710.00 91 Ω Y_2 w S_2^2 8381 301.0203.00 20 k Y_2 w S_2^2 8383 301.0474.00 470 k Y_2 w S_2^2 8384 315.062.00 6.8 k Y_2 w S_2^2 8385 302.0101.00 100 Ω Y_2 w S_2^2 8385 304.015.00 1 meg 1 w S_2^2 8385 304.015.00 1 meg 1 w S_2^2 8397 304.055.00 560 Ω Y_2 w S_2^2 8398 301.055.00 1 meg Y_2 w S_2^2 8400 302.0101.00 1 00 Ω Y_2 w S_2^2 8401 302.0455.00 3 meg Y_2 w S_2^2 8420 314.6101.00 100 Ω Y_2 w S_2^2 8421 316.0101.00 100 Ω Y_2 w S_2^2	R373	301-0561-00			560 Ω	1/2 W		5%
R375 302.0271.00 270 n $1/3 \text{ w}$ 5% R380 301.0910.00 91 Ω $1/3 \text{ w}$ 5% R381 301.0203.00 20 k $1/3 \text{ w}$ 5% R381 301.0203.00 20 k $1/3 \text{ w}$ 5% R383 301.0474.00 470 k $1/3 \text{ w}$ 5% R384 315.0682.00 6.8 k $1/4 \text{ w}$ 5% R395 302.010.200 1 k $1/3 \text{ w}$ 5% R395 304.0105.00 1 meg 1 w 5% R396 311.0005.00 560 \Omega $1/2 \text{ w}$ 5% R397 301.0561.00 560 \Omega $1/2 \text{ w}$ 5% R397 301.0562.00 1 meg $1/2 \text{ w}$ 5% R401 302.011.50 1 meg $1/2 \text{ w}$ 5% R401 302.011.50 33 k 2 w 5% R401 302.030.01 15 k 2 w 5% R422 30.02.03.0	R374	311-0078-00	X105		50 k. Var	0.1 w		- 70
R380 $301-0910-00$ 91Ω $V_2 w$ 5% R381 $301-0910-00$ $20 k$ $V_2 w$ 5% R383 $301-0474.00$ $470 k$ $V_2 w$ 5% R383 $301-0474.00$ $470 k$ $V_2 w$ 5% R384 $315.0682.00$ $68 k$ $V_2 w$ 5% R385 $302.0102.00$ $1 k$ $V_2 w$ 5% R393 $306.053.00$ $56 k$ $2 w$ 3% R395 $304.050.00$ $1 meg$ $1 w$ 386 R395 $304.050.00$ 500Ω $V_2 w$ 5% R396 $301.055.00$ $1 meg$ $1 w$ 5% R397 $302.051.00$ $56 k$ $V_2 w$ 5% R400 $302.20105.00$ $1 meg$ $V_2 w$ 5% R401 $302.20105.00$ $1 meg$ $V_2 w$ 5% R402 $316.0101.00$ 100Ω $V_2 w$ 5% R421 <td< td=""><td>R375</td><td>302-0271-00</td><td></td><td></td><td>270 Ω</td><td>1/2 W</td><td></td><td></td></td<>	R375	302-0271-00			270 Ω	1/2 W		
R381 301-023-00 20 k V_{1} w 5% R383 301-0474-00 470 k V_{1} w 5% R384 315-062-00 68 k V_{1} w 5% R385 302-0101-00 100 Ω V_{1} w 5% R395 302-0102-00 1 k V_{2} w 5% R395 304-015-00 1 meg 1 w 8396 R396 311-005-00 560 Ω V_{2} w 5% R397 301-0561-00 560 Ω V_{2} w 5% R397 301-0561-00 560 Ω V_{2} w 5% R399 301-0562-00 1 00 Ω V_{2} w 5% R400 302-0101-00 1 00 Ω V_{2} w 5% R401 302-0405-00 6.8 meg V_{2} w 5% R420 316-0101-00 100 Ω V_{2} w 5% R421 316-0101-00 100 Ω V_{2} w 5% R422 302-0302-300 10	R380	301-0910-00			91 Ω	1∕₂ w		5%
Assn Join 2012-2020 Zok Y2 w S2, Y2 w S2, S2, S2, S2, S2, S2, S2, S2, S2, S2,	D201	201 0202 00			20 h	17		50/
ASS3 JUA JUK Y_2 w $3\gamma_4$ R384 J15.062.20 6.8 k Y_4 w $3\gamma_4$ R385 J02.0101.00 100 Ω Y_4 w $3\gamma_4$ R393 J06.0563.00 56 k 2 w R395 J02.0102.00 1 k Y_2 w R396 J1.0005.00 1 meg 1 w R397 J01.0561.00 560 Ω Y_2 w $5\gamma_4$ R397 J01.0561.00 560 Ω Y_2 w $5\gamma_4$ R397 J01.0562.00 1 meg Y_2 w $5\gamma_4$ R400 J02.2015.00 1 meg Y_2 w $8\gamma_4$ R401 J02.6465.00 6.8 meg Y_2 w 840 R401 J02.6465.00 6.8 meg Y_2 w 842 R420 J1.6101.00 100 Ω Y_4 w Y_4 w R421 J1.6101.00 100 Ω Y_4 w $5\gamma_6$ R422 J05.013.00 15 k 2 w $5\gamma_6$	R301	301-0203-00			20 K	72 W		5%
Kasa J J J J J S J S J S J S K V / W S / K V / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / W S / K M / K M / W S / K M / K <thm k<="" th=""> <thm k<="" th=""> <thm k<="" td=""><td>R383</td><td>301-04/4-00</td><td></td><td></td><td>4/0 k</td><td>1/2 W</td><td></td><td>5%</td></thm></thm></thm>	R383	301-04/4-00			4/0 k	1/2 W		5%
Kabb 302/11/1-30 100 ff Y ₂ w R3970 302/11/2-00 1 k Y ₂ w R393 306-0563-00 56 k 2 w R395 304-0105-00 1 meg 1 w R396 311-0005-00 560 Ω V ₂ w 5% R397 301-0561-00 560 Ω V ₂ w 5% R397 301-0561-00 560 Ω V ₂ w 5% R400 302-0105-00 1 meg V ₂ w 5% R401 302-0655-00 3 k 2 w 8% R401 302-0655-00 3 k 2 w 8% R401 302-065-00 33 k 2 w 8% R420 316-010-00 100 Ω V ₄ w 8% R421 316-010-00 100 Ω V ₄ w 5% R422 305-013-00 15 k 2 w 5% R424 Use 303-0243-00 101 104 39 k 1 w 5% R425 305-0393-00	K384	315-0682-00			0.8 K	1/4 W		5%
K390 302/102/20 I K Y ₂ w R393 306/0563.00 56 k 2 w R385 304/0105.00 I mg I w R397 301/0561.00 560 II, Var R397 301/0561.00 560 II, Var R397 301/0562.00 5.6 k Y ₂ w 5% R398 301/0562.00 5.6 k Y ₂ w 5% R400 302/015.00 I meg Y/w 8% R401 302/015.00 1 meg Y/w 8% R420 316/0101.00 100 II Y/w 8% R420 316/0101.00 100 II Y/w 5% R422 303/023.00 101 100 II Y/w 5% R422 303/023.00 101 104 39 k 1 w 5% R425 Use 303/023.00 101 104 39 k 2 w 5% R426 302/032.00 105 39 k 2 w 5% <t< td=""><td>K385</td><td>302-0101-00</td><td></td><td></td><td>100 12</td><td>1/2 ₩</td><td></td><td></td></t<>	K385	302-0101-00			100 12	1/2 ₩		
R393 306-0563-00 56 k 2 w R385 304-0105-00 1 meg 1 w R397 311-0005-00 500 Ω_{1} Var R397 301-0561-00 560 Ω_{1} Var R397 301-0562-00 560 Ω_{1} Var R397 302-0101-00 100 Ω_{1} Var R401 302-0105-00 1 meg R401 302-0105-00	K390	302-0102-00			l k	1∕2 ₩		
R385 304-0105-00 1 meg 1 w R396 311-0005-00 500 Q, Var R397 301-0561-00 560 Q Y ₂ w 5% R398 301-0562-00 5.60 L Y ₂ w 5% R397 301-0562-00 5.60 L Y ₂ w 5% R399 302-0105-00 100 Q Y ₂ w 5% R400 302-0105-00 1 meg Y ₂ w 5% R401 302-02685-00 6.8 meg Y ₂ w 5% R405 304-033-300 100 Q Y ₄ w 5% R422 316-0101-00 100 Q Y ₄ w 5% R422 316-0101-00 100 Q Y ₄ w 5% R422 302-015-300 15 k 2 w 5% R425 Use 302-021-300 101 104 3 k 1 w 5% R425 302-0105-00 39 k 2 w 5% 5% R433 301-0105-00 1 meg Y ₆ w 5%	R393	306-0563-00			56 k	2 w		
R396 311.0005.00 500 μ Var R397 301.0561.00 560 μ ½ w 5% R397 301.0561.00 560 μ ½ w 5% R398 301.0562.00 5.6 k ½ w 5% R400 302.0101.00 1 00 μ ½ w 5% R400 302.20105.00 1 meg ½ w 5% R401 302.6485.00 6.8 meg ½ w 5% R420 316.0101.00 100 μ ½ w 5% R421 316.0101.00 100 μ ½ w 5% R422 305.013.00 15 k 2 w 5% R425 Use 303.0243.30 101 104 39 k 1 w 5% R426 Use 303.03243.30 101 104 39 k 1 w 5% R426 305.0393.00 105 39 k 2 w 5% R432 302.0324.00 300 k ½ w 5% R433 301.0105.00 1 meg ½ w 5%	R385	304-0105-00			1 meg	1 w		
R397 301-0561-00 560 n ½ w 5% R398 301-0562-00 5.6 k ½ w 5% R399 302-0101-00 100 Ω ½ w 5% R400 302-0105-00 1 meg ½ w 8% R401 302-0105-00 1 meg ½ w 8% R404 302-0105-00 1 meg ½ w 8% R405 316-0101-00 100 Ω ¼ w 8% R421 316-0101-00 100 Ω ¼ w 8% R422 305-0153-00 15 k 2 w 5% R425 Use 303-0272-300 101 104 39 k 1 w 5% R426 305-0393-00 101 104 39 k 2 w 5% R426 302-0393-00 105 39 k 2 w 5% R433 301-0105-00 1 meg ½ w 5% R433 301-0105-00 1 meg ½ w 5% R440 302-015-00 1 meg ½	R396	311-0005-00			500 Ω, Var			
R399 301-0562-00 5.6 k ½ w 5% R399 302-0101-00 100 Ω ½ w 5% R400 302-20105-00 1 meg ½ w R401 302-6485-00 6.8 meg ½ w R404 304-0333.00 33 k mg 2 w R420 316-0101-00 100 Ω ¼ w R421 36-01013.00 15 k 2 w R422 305-0103.00 15 k 2 w R422 305-0103.00 12 k 1 w R422 305-0103.00 12 k 1 w R425 Use 303.0243.30 101 104 39 k 1 w R426 Use 303.0243.30 101 104 39 k 1 w 5% R426 305.0393.00 105 39 k 2 w 5% R432 302-0334.00 30 k ½ w 5% R433 301-0105.00 1 meg ½ w 5% R433 301-0105.00 1 meg ½ w 5%	R397	301-0561-00			560 Ω	1/2 W		5%
R399 302.0101.00 100 Ω ½ w R400 302.2105.00 1 meg ½ w R401 302.20485.00 6.8 meg ½ w R401 302.20485.00 33 k 2 w R402 304.0101.00 100 Ω ¼ w R421 316.0101.00 100 Ω ¼ w R422 335.015.300 15 k 2 w 5% R425 Use 305.033.00 101 104 39 k 1 w 5% R426 305.033.00 101 104 39 k 1 w 5% R426 305.033.00 105 39 k 2 w 5% R426 302.033.00 101 104 39 k 1 w 5% R432 302.033.400 330 k ½ w 5% R433 301.0105.00 1 meg ½ w 5% R440 302.0225.400 101 136 22 meg ½ w 5%	R398	301-0562-00			5.6 k	1/2 W		5%
R399 302.2101.00 100 Ω ½ w R400 302.2105.00 1 meg ½ w R401 302.20585.00 6.8 meg ½ w R405 306.4333.40 33 k 2 w R420 31.6.101.00 100 Ω ¼ w R421 31.6.2015.30 15 k 2 w R422 305.20153.30 15 k 2 w 5% R425 Use 302.623.300 101 104 34 k 1 w 5% R425 Use 302.633.300 105 39 k 2 w 5% R426 305.633.30 105 39 k 2 w 5% R426 305.633.30 105 39 k 2 w 5% R432 302.015.00 1 meg ½ w 5% R433 301.0105.00 1 meg ½ w 5% R440 302.025.00 131 136 22 meg ½ w								
R400 302-20105-00 1 meg V/2 w R401 302-2685-00 6.8 meg V/2 w R406 306-0333-00 33 k 2 w R400 316-101-00 100 Ω V/2 w R421 316-0101-00 100 Ω V/2 w R422 305-0153-00 15 k 2 w 5% R425 Use 305-0373-300 101 104 39 k 1 w 5% R426 305-03793-300 105 39 k 2 w 5% R432 302-0332-00 105 39 k 2 w 5% R433 301-0105-00 1 meg V/2 w 5% R433 301-0105-00 1 meg V/2 w 5% R440 302-02226-00 101 136 22 meg V/2 w R440 302-012-00 136 22 meg V/2 w 5%	R399	302-0101-00			100 Ω	1∕2 w		
K401 302-02685.00 6.8 meg V/2 w R406 306-0333.00 33 k 2 w R420 316-0101-00 100 Ω V/2 w R421 316-0101-00 100 Ω V/2 w R422 305-0153.00 15 k 2 w 5% R425 Use 305-0273.00 101 104 39 k 1 w 5% R426 305-0393.00 101 104 39 k 2 w 5% R426 305-0393.00 105 39 k 2 w 5% R432 302-0334.00 330 k V/2 w 5% R433 301-0105.00 1 meg V/2 w 5% R433 301-0105.00 1 meg V/2 w 5% R440 302-02226.00 101 136 22 meg V/2 w R440 302-02226.00 137 15 meg V/2 w 5%	R400	302-0105-00			1 meg	½ ₩		
R406 306-0333-00 33 k 2 w R420 316-0101-00 100 n 1/4 w R421 316-0101-00 100 n 1/4 w R421 316-0101-00 100 n 1/4 w R422 305-0153-00 15 k 2 w 5% R425 Use 305-0373-300 101 104 39 k 1 w 5% R426 305-03973-00 105 39 k 2 w 5% R432 302-0334-00 330 k ½ w 5% R433 301-0105-00 1 meg ½ w 5% R434 302-015-00 1 meg ½ w 5% R440 302-02226-00 101 136 22 meg ½ w R440 302-015-00 15 meg ½ w 5%	R401	302-0685-00			6.8 meg	½ ₩		
R420 316-0101-00 100 Ω ½ w R421 316-0101-00 100 Ω ½ w R422 305-0153-00 15 k 2 w 5% R422 305-0153-00 15 k 2 w 5% R425 Use 303-023-00 101 104 39 k 1 w 5% R426 305-0393-00 105 39 k 2 w 5% R426 305-0393-00 105 39 k 2 w 5% R426 305-0393-00 105 39 k 2 w 5% R426 302-0105-00 1 meg ½ w 5% R433 301-0105-00 1 meg ½ w 5% R440 302-02226-00 101 136 22 meg ½ w R440 302-02226-00 137 15 meg ½ w 5%	R406	306-0333-00			33 k	2 w		
R421 316-0101-00 100 Ω ½, w R422 305-0153-00 15 k 2w 5%, R425 Use 303-023-00 10 L 1/4 5%, R426 Use 305-0393-00 101 104 39 k 1 w 5%, R426 305-0393-00 105 39 k 2 w 5%, R426 302-0334-00 330 k ½ w 5%, R433 301-0105-00 1 meg ½ w 5%, R435 301-0105-00 1 meg ½ w 5%, R440 302-0225-00 101 136 22 meg ½ w R440 302-0225-00 15 15 meg ½ w 5%,	R420	316-0101-00			100 Ω	1/4 w		
KA22 SOG (33) (243,00 SOG (34) (24) (24) (24) (24) (24) (24) (24) (2	P/21	316-0101-00			100 0	1/. w		
Mark Subset Subse	R421	205 0152 00			15 6	74 W		Eal
KA22 Use 305-303-300 101 104 37 k 1 w 57 k K426 305-0393-00 105 37 k 2 w 5 k K426 305-0393-00 105 37 k 2 w 5 k K426 302-0334-00 30 k 2 w 5 k 5 k K433 301-0105-00 1 meg 1 k w 5 k K435 301-0105-00 1 meg 1 k w 5 k K440 302-0225-00 101 136 22 meg 1 k w 5 k K440 302-015-00 13 15 meg 1 k w 5 k	R422	303-0133-00			13 K	2 W		5%
κα∠ο Use Use Use Use UN S% I w S% R426 305.0393.00 105 39 k 2 w S% R426 302.0393.00 105 39 k 2 w S% R433 301.0105.00 1 meg ½ w S% R435 301.0105.00 1 meg ½ w S% R440 302.0252.60 101 136 22 meg ½ w R440 302.0256.00 137 15 meg ½ w 5%	R420 R407	Use 303-0243-00	101	104	24 K 20 L	1		5%
κα∠ρ 302-0334-00 105 37 k 2 w 5% R432 302-0334-00 330 k ½ w 5% R433 301-0105-00 1 meg ½ w 5% R435 301-0105-00 1 meg ½ w 5% R440 302-0222-60 101 136 22 meg ½ w R440 302-015-00 13 15 meg ½ w 5%	K420	Use 305-0393-00	101	104	37 K	1.0		5%
R432 302-0334-00 330 k ½ w R433 301-0105-00 1 meg ½ w 5% R435 301-0105-00 1 meg ½ w 5% R436 301-0105-00 1 meg ½ w 5% R440 302-0225-00 101 136 22 meg ½ w R440 302-015-00 137 15 meg ½ w 5%	K426	302-0393-00	105		37 K	2 W		5%
R433 301-0105-00 Ì meg ¼'w 5% R435 301-0105-00 Ì meg ¼'w 5% R440 302-0225-00 101 136 22 meg ¼'w 5% R440 302-015-00 137 Ì5 meg ¼'w	R432	302-0334-00			330 k	1/2 w		
R435 301-0105-00 1 meg ½ w 5% R440 302-0225-00 101 136 22 meg ½ w R440 302-015-00 137 15 meg ½ w	R433	301-0105-00			1 mea	1/2 w		5%
R440 302-0226-00 101 136 22 meg ½ w R440 302-0156-00 137 15 meg ½ w	R435	301-0105-00			1 meg	1/2 w		5%
R440 302-0156-00 137 15 meg 1/2 w	R440	302-0226-00	101	136	22 meg	1/2 W		- 70
	R440	302-0156-00	137		15 meg	1/2 W		

Ckt. No.	Tektronix Part No.	Serial/M Eff	odel No. Disc		Descrip	otion	
P//1	211 0152 00			0 5001 14			
R441	302 0224 00	101	12/	2 x 500 k, Var	1/		
P442	202 01 54 00	107	130	22 meg	<i>Y</i> ₂ ₩		
P445	302-0138-00	13/		15 meg	1/2 ₩		
D442	302-0101-00			100 Ω	%2 ₩		
K440	302-0101-00			100 Ω	1∕2 ₩		
R602	306-0180-00			18 Ω	2 w		
R603	311-0055-00			50 Ω. Var		ww	
R606	311-0055-00			50 Ω. Var		ww	
R610	304-0100-00			10 \Q	1 w		
R613	308-0027-00			30 k	10 w	ww	5%
R620	306-0104-00			100 k	2 w		
R621	301-0204-00			200 k	1/		E o/
R623	302-0105-00			1 meg	1/2 W		3%
R624	302-0102-00			1 k	1/2 W		
R625	Use 304-0123-00			12 k	1 w		
R626	306-0223-00			22 k	2		
R627	Use 308-0018-00			254	10	11011	
R628	306-0101-00	¥611		100 0	10 w	VV VV	
R633	306-0333-00	AGTT		22 6	2 w		
R634	302-0224-00			33 K	2 W		
1004	502-022-00			220 K	72 W		
R641	302-0224-00			220 k	½ w		
R642	304-0823-00			82 k	1 w		
R644	302-0102-00			1 k	½ ₩		
R645	302-0224-00			220 k	½ ₩		
K646	309-0045-00			100 k	½ w	Prec	1%
R647	311-0015-00			10 k. Var		ww	
R648	309-0090-00			50 k	1/2 W	Prec	1%
R650	308-0171-00			0.5 Ω	20 w	ww	5%
R651	308-0165-00	X403		0.5 Ω	5 w	ww	5%
R655	308-0179-00			5Ω	5 w	ww	5%
R660	306-0100-00			10.0	2		
R661	306.0100.00			10.0	2 W		
2662	308-0100-00			201	2 W	1017	
P470	202 0224 00			30 K	10 w	W W	5%
R671	302-0224-00			220 k 27 k	1/2 ₩ 1/2 ₩		
0/70	00/ 010/ 55						
K0/2	306-0184-00			180 k	2 w		
K0/3	302-0105-00			1 meg	½ w		
K6/4	302-0102-00			1 k	½ w		
R677	308-0017-00			2 k	10 w	ww	5%
R680	302-0105-00			1 meg	½ w		
R681	302-0395-00			39 meg	1/- 14		
R682	302-0104-00			100 k	1/2 W		
R683	302-0823-00			82 4	1/2 **		
R684	Use 301-0274-00			270 k	1/2 W		50/
R685	302-0102-00			ĩk	1/2 **		3%
					/2 **		

	Tektronix	Serial/M	odel No.				
Ckt. No.	Part No.	Eff	Disc		Descrip	tion	
				0001			
R688	309-0261-00			300 k	1/2 W	Prec	1%
R689	309-0101-00			330 k	1/2 W	Prec	1%
R693	302-0474-00			470 k	½ ₩		
R694	302-0102-00			1 k	1/2 W		
R695	302-0102-00			1 k	1/2 w		
R697	308-0033-00			4.5 k	20 w	WW	5%
R698	309-0046-00	101	1059	111 k	1/2 W	Prec	1%
R698	323-0610-00	1060		111 k	1/2 W	Prec	1%
P699	309-0276-00	101	1059	267 k	1/2 W	Prec	1%
R699	323-0426-00	1060	1007	267 k	1/2 W	Prec	1%
R700	306-0100-00			10 Ω	2 w		
R701	306-0100-00			10 Ω	2 w		
R702	308-0027-00			30 k	10 w	ww	5%
R704	308-0162-00			60 Ω	5 w	ww	5%
R710	302-0184-00			180 k	½ w		
R711	302-0273-00			27 k	½ w		
R712	306-0334-00			330 k	2 w		
R713	302-0105-00			1 mea	1/2 W		
R714	302-0102-00			1 k	1/2 W		
P715	302.0102.00			16	1/2 11/2		
K/15	302-0102-00				72 **		
P717	308-0093-00			12 k	8 w	ww.	5%
P718	309-0170-00			429 k	1/2 W	Prec	1%
0710	210 0007 00			900 k	1	Proc	1 %
N/ 17	310-0077-00			2201	2	riec	1 /0
R720	306-0334-00			330 K	2 W		
R/21	302-0683-00			08 K	72 W		
D700	201 0244 00			240 1	1/		E %/
N/ 22	200 0474 00			470 L	1/		J /0
K/23	302-04/4-00			470 K	72 W		
K/24	302-04/0-00			4/ 12	1/2 ₩		
R725	302-0102-00			l k	/₂ ₩		
R726	304-0151-00	X575		150 Ω	1 w		
·							
R727	304-0151-00	X575		150 Ω	1 w	-	
R728	310-0064-00			500 k	1 w	Prec	1%
R729	310-0119-00			332 k	1 w	Prec	1%
R730	302-0684-00			680 k	1/2 W		
R731	304-0151-00	X575		150 Ω	1 w		
R732A-J (9)	311-0039-00			1 meg, Var		2 nsec-1 µsec	
R734	302-0470-00			47 Ω	1/2 W		
R735	302-0470-00			47 Ω	1/2 w		
R740	302-0224-00			220 k	1/2 w		
P741	202 0224 00			220 k	1/2 w		
D740	202 0100 00			10.0	1/2 "		
N/ 4Z	302-0100-00			10.12	72 W		
07.0	20/ 012/ 00			120 k	2		
N/ 40	300-0124-00			20 1	1/		
K/51	302-0393-00			37 K	1/2 W		
R752	304-0154-00			150 k	1 w		
R753	302-0105-00			1 meg	1/2 W		
R754	302-0102-00			1 k	½ ₩		
R755	302-0102-00			1 k	1/2 w		

Ckt. No.	Tektronix Part No.	Serial/Me Eff	odel No. Disc		Descrip	otion	
				007 -	1/	Pres	1.0/
R758	Use 309-0328-00			287 K	72 W	Prec	1%
K/59	Ose 310-0100-00	101	1000	11 74 L	14. 14	Prec	1/. %
R/OU	309-0200-00	1240	1237	11.84	1/2 W	Prec	1%
R/OU P741	323-0276-00	1240		100 k	1/2 W	Prec	1%
101	036 323-0303-00			100 1	/2		. ,.
R763	302-0682-00	101	742	6.8 k	½ w		
R763	302-0272-00	743		2.7 k	1/2 W		
R764	302-0332-00			3.3 k	1/2 ₩		50/
R/65	301-0101-00			100 12	1/2 W		5%
K/66	302-0154-00			150 K	72 W		
R767	*310-0535-00			1 Ω	4 w	Prec	1/2 %
R768	308-0075-00			100 Ω	3 w	ww	
R769	308-0075-00			100 Ω	3 w	WW	
R770	308-0242-00	X403		0.25 Ω	5 w	ww	
R771	304-0104-00	101	/42	100 k	l w		
R771	306-0393-00	743		39 k	2 w		
R773	306-0102-00			1 k	2 w		
R777	308-0184-00			7.5 k	25 w	ww	
R778	308-0162-00			60 Ω	5 w	ww	5%
R780	302-0154-00			150 k	1∕2 w		
R781	302-0104-00			100 k	1/2 w		
R783	302-0104-00			100 k	1/2 W		
R785	302-0104-00			100 k	1/2 w		
R801	304-0391-00			390 Ω	1 w		
R803	306-0124-00	101	383	120 k	2 w		
R803	306-0104-00	384		100 k	2 w		
R804	306-0124-00	101	383	120 k	2 w		
R804	306-0104-00	384		100 k	2 w		
R806	302-0104-00			100 k	½ ₩		
R807	302-0152-00			1.5 k	½ w		
R815	302-0105-00			1 meg	1/2 w		
R834	302-0474-00			470 k	1/2 W		
R835	302-0474-00			470 k	1/2 w		
R840	304-0185-00			1.8 meg	1 w		
R841	311-0039-00			1 meg, Var			
R842	306-0475-00	101	521	4.7 meg	2 w		
R842	306-0275-00	522		2.7 meg	2 w		
R843	306-0475-00	101	521	4.7 meg	2 w		
R843	306-0275-00	522		2.7 meg	2 w		
R844	306-0475-00	101	521	4.7 meg	2 w		
R844	306-0225-00	522		22 meg	2 w		
R845	306-0225-00	X522		2.2 meg	2 w		
R846	306-0225-00	X522		2.2 meg	2 w		
R847	306-0225-00	X522		2.2 meg	2 w		
R848	316-0102-00	X914		1 k	1/4 w		
R849	306-0156-00	X158		15 meg	2 w		

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(B)

	Tektronix	Serial/M	odel No.				
Ckt. No.	Part No.	Eff	Disc		Descrip	tion	
R850	306-0275-00	101	157	27 meg	2 w		
R850	306-0225-00	158	10/	2.2 meg	2 w		
R851	306-0275-00	101	157	27 meg	2		
R851	306-0225-00	158	107	2.2 meg	2 w		
R852	306-0225-00	101	157	2.2 meg	2 w		
1002	000 0110 00	101	10,	Liz mog			
P852	306.0155.00	158		15 meg	2		
R853	311.0121.00	101	1239	5 meg Var	2.4		
P853	311.0121.01	1240	1257	5 meg, Var			
R055	306-0155-00	101	157	15 meg	2		
R854	306-0105-00	158	10/	1 mer	2		
R855	302-0274-00	101	383	270 k	1/2 w		
R855	311-0126-00	384		1 meg, Var			
R856	311-0041-00	101	1239	1 meg, Var			
R856	311-0041-02	1240		1 meg, Var			
R857	316-0274-00			270 k	1/4 w		
R858	316-0105-00			1 meg	1/4 w		
R859	316-0223-00			22 k	1/4 w		
R860	311-0141-00			2 k, Var		ww	
R861	311-0032-00			250 k, Var			
R862	302-0102-00			1 k	1/2 W		
R863	302-0102-00			1 k	½ w		
R864	311-0032-00			250 k, Var			
00/5	011 0000 00			0501 1/			
K865	311-0032-00			250 K, Var	17		
K800	302-0102-00			1001	72 W		
K00/	304-0124-00	101	004	120 K	1/		E 0/
K868	OSE 301-0183-00	101	224	18 K	1/2 W		5%
K000	301-0163-00	225		IOK	72 W		5,
R869	302-0103-00			10 k	1/2 W		
R870	309-0329-00			2.87 meg	1/2 w	Prec	1%
R871	309-0330-00			319 k	1/2 W	Prec	1%
R872	309-0328-00			287 k	1/2 W	Prec	1%
R875	311-0032-00			250 k, Var			
R876	309-0331-00	101	1059	20.2 k	% ₩		
R876	309-0392-00	1060	,	20 k	1/2 ₩	Selected	(nominal value)
R8/8	****		1	Selected w/K8/9	to 28 k		
P970	*312-0016-00		1	30 k Var	шт.		
K0/ /			,				
P992	316.0473.00			17 k	1/		
P883	319,0034,00			125 0	14. w		1%
R8858	311-0237-00			10 k. Var	/4	ww	• •
R886	308-0092-00			4.5 k	5 w	ww	
R888	302-0472-00			4.7 k	1/2 W		
R890	302-0152-00			1.5 k	½ w		
R891	302-0274-00			270 k	½ ₩		
R892A,B	311-0216-00			2 x 2 meg, Var			
R893	302-0474-00			470 k	½ ₩		
R895	302-0474-00			470 k	1∕₂ w		
⁸ Furnished as a	unit with SW885.						

Ckt. No.	Part No.	Serial/M	odel No.				
	Tektronix	Eff	Disc		Descrip	otion	
P898	302 0543 00			54 L	1/		
R910	302.0470.00			47.0	1/		
D011	201 0005 00			4/ 12	72 W		5.01
0010	*010.0225-00			2.2 meg	% ₩		5%
K913	*310-0503-00			25 k	8 w	ww	1%
K915	302-04/0-00			47 Ω	1∕2 w		
R916	302-0102-00			1 k	1/2 w		
R918	302-0564-00	101	481	560 k	1/2 W		
R918	306-0124-00	482		120 k	2 w		
R919	306-0274-00	101	481	270 k	2 w		
R919	301-0473-00	482		47 k	½ w		5%
R921	306-0473-00	101	481	47 k	2 w		
R921	304-0334-00	482		330 k	1 w		
R923	311-0218-00	101	481	50 k. Var		ww	
R923	311-0039-00	482		1 mea. Var			
R924	311-0159-00	101	481	20 k, Var		ww	
R924	311-0329-00	482	545	50 k Var			
R924	311-0326-00	566	505	10 k Var			
R925	309-0014-00	101	401	1	1/	Beer	1.0/
P925	200 0017 00	400	401	1 meg	72 W	Frec	1%
R925	309-0016-00	566	505	1.23 meg	1/2 ₩ 1/2 ₩	Prec	1%
K926	302-04/0-00			47 Ω	1∕2 ₩		
K928	302-0223-00			22 k	1∕2 w		
K929	305-0473-00			47 k	2 w		5%
R930	316-0332-00	X482		3.3 k	¼ w		
R931	311-0159-00	101	481	20 k, Var			
R931	311-0329-00	482		50 k. Var			
R932	302-0103-00	101	481	10 k	1/2 W		
R932	302-0472-00	482		4.7 k	1/2 W		
R933	302-0562-00			56k	1/2		
R934	305-0273-00	X482		27 k	2 w		5%
R937	315-0121-00	101	481	120 k	1/		E 9/
R937	315-0270-00	482		27.0	74 W	Neminal Val	5%
R938	315.0121.00	101	4002	2/ 14 100 L	74 W	Nominal value	5%
R939	315-0103-00	101	4028	10 k	% w		5% 5%

Switches

SW10	Wired *262-0378-00	Rotary	TRIGGER SOURCE
SW10	*260-0359-00	Rotary	TRIGGER SOURCE
SW20	Wired *262-0377-00	Rotary	GAIN
SW20	*260-0358-00	Rotary	GAIN
SW50	Wired *262-0376-00	Rotary	FUNCTION

Wired or Unwired

-		Tektronix	Serial/Model	No.		
Ckt. N	o.	Part No.	Eff	Disc	De	scription
	Wired	or Unwired				
SW50		*260-0357-00			Rotary	FUNCTION
SW160	Use	*260-0447-00			Slide	SWEEP DIODE
SW168		*260-0247-00			Push Button	RESET
SW336	Wired	*050-0058-00	101	161	Rotary	NANOSEC/CM
SW336		*260.0356.00	101	161	Rotary	NANOSEC/CM
0000		200 0000 00	101	101	Rolary	TARTAOSEC/CM
014004		+050 0050 00				
577336	Wired	*050-0058-00	162	302	Rotary	NANOSEC/CM
SW336		*260-0385-00	162	302	Rotary	NANOSEC/CM
SW336	Wired Use	*262-0500-00	303		Rotary	NANOSEC/CM
SW336	Use	*260-0385-00	303		Rotary	NANOSEC/CM
SW601	Use	*050-0194-00	101	402	Toggle	POWER ON
\$14/201		0/0 050/ 00	(02		T I.	
SW001		200-0306-00	403		loggie	POWER ON
SVV000	M.C. 1.11	136-0094-00	101	5.00	Tini-Jax	6.3 V CAMERA
511/070	vvired Use	*262-0611-00	101	549	Rotary	RANGE
51/18/0	Use	*260-0361-00	101	549	Rotary	RANGE
5///8/0	Wired	*262-0611-00	550		Rotary	RANGE
SW870		*260.0578.00	550		Poterv	PANGE
\$\A/971	He	260-03/ 0-00	550		Slide	BOLADITY
C14/0058	036	211 0227 00			SIDE	FOLARIT
C/4/002	1.1~~	*050 0242 00	101	(10	Brod	DEDI & CEMENIT, KIT
3 11 000	Ose	030-0243-00	101	610	Keed	KEPLACEMEINT KIT
SW886		260-0693-00	611		Mercury	SINGLE CLOSURE
SW920	Wired	*262-0379-00			Rotary	MULTIPLIER
SW920		*260-0360-00			Rotary	MULTIPLIER
				Transform	iers	
T4		*660-0402-00			Trigger Takeoff Assy	na body (Ferrite Core 276-0517-00)
T20		276-0519-00			Toroid Core	
T70		*120-0195-00	101	1114	Toroid TD28	
T70		*120-0195-01	1115	1114	Toroid 3 turns guadfila	r
T160		*120.0178.00	1113		Toroid TD19	
1100		120-017 0-00				
T180		*120-0194-00	101	1114	Toroid TD27	
T180		*120-0194-01	1115		Toroid, 4 turns, trifilar	
T198		*120-0193-00			Toroid TD26	
T601		*120-0186-00			L. V. Power	
T602		*120-0200-00			AC Line Filter Toroid, TI	030
T801		*120-0188-00			High Voltage	
T885		*660.0403.00			Trigger Takeoff Assy sh	ort body (Ferrite Core 276-0517-00)
T930		*120-0325-00	X482		Toroid 5T TD109	
.,		0 0010 00				

Switches (Cont)

Thermal	Cutout
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TK602			Use	260-	0336-00	
⁹ Furnished	as	a	unit	with	R885.	

150°

Electron Tubes (Cont)

Ckt. No.	Tektronix Part No.	Serial/A Eff	Nodel No. Disc		Description	
V114	154 0015 00			(/00		
V102	154-0213-00			0000		
V123	154-016/-00			6018		
V134	154-0215-00			6668		
V143	154-0187-00			6D18		
V184	Use 154-0187-00			6DJ8		
V194	Use 154-0187-00			6D.18		
V214	Use *050-0258-00	101	742	Replacement Kit		
V214	154-0496-00	743		12HG7		
V223	154-0187-00	101	742X	6DJ8		
V244	Use *050-0258-00	101	742	Replacement Kit		
V244	154-0496-00	743		12HG7		
V264	154-0022-00			6AU6		
V274	154-0202-00			6CW5/FL86		
V283	154-0187-00			6DJ8		
V312	154-0038-00			12415		
V322	154-0038-00			12415		
V331	154-0300-00			4CX250F		
V332	154-0301-00			64F3		
V343	154-0187-00			4D 18		
				0000		
V353	154-0187-00			6DJ8		
V363	154-0187-00			6DJ8		
V3/4	154-018/-00			6DJ8		
V388	154-0047-00			12BY7		
V393	154-0187-00			6DJ8		
V394	154-0040-00			12AU6		
V403	154-0047-00			12BY7		
V424	154-0187-00			6DJ8		
V624	154-0022-00			6AU6		
V627	154-0056-00			6080		
V639	154-0291-00			0G3/85A2		
V646	154-0043-00			12AX7		
V674	154-0040-00			12AU6		
V686	154-0043-00			12AX7		
V694	154-0040-00			12AU6		
V697	154-0056-00			6080		
V717	154-0260-00			7734		
V724	154-0040-00			12AU6		
V737	154-0056-00			6080		
V757	154-0260-00			7734		
V800	154-0021-00			64115		
V802	150-0005-01			11/28		
V812	150-0005-01			1 Y 2 B		
V814	154-0041-00			124117		
V822	150-0005-01			1X2B		

Electrical Parts List—Type 519

Ckt. No.	Tektronix Part No.	Serial/Model No. Eff Disc	Description	
V832 V859 ¹⁰ V885 V895 V915	154-0051-00 *154-0308-00 154-0017-00 154-0187-00 154-0187-00		5642 T5191-11 CRT with shield 6AQ5 6DJ8 6DJ8	

Electron Tubes

¹⁰When ordering a replacement CRT be sure to check the serial number of the CRT in your instrument (on the bezel cover over the front of the CRT). If the CRT serial number is 1016 or below, order replacement by Tektronix Part No. 154:0356:00. If it is 1017 or obsev, order by Tektronix Part No. 154:0306:00.

SECTION 8

PARTS LIST AND DIAGRAMS

PARTS ORDERING INFORMATION

Replacement parts are available from or through your local Tektronix Field Office.

Changes to Tektronix instruments are sometimes made to accommodate improved components as they become available, and to give you the benefit of the latest circuit improvements developed in our engineering department. It is therefore important, when ordering parts, to include the following information in your order: Part number including any suffix, instrument type, serial number, and modification number if applicable.

If a part you have ordered has been replaced with a new or improved part, your local Tektronix Field Office will contact you concerning any change in part number.

a or amp	amperes	mm	millimeter
BHS	binding head steel	meg or M	megohms or mega (10 ⁶)
с	carbon	met.	metal
cer	ceramic	μ	micro, or 10 ⁻⁶
cm	centimeter	'n	nano, or 10 ⁻⁹
comp	composition	Ω	ohm
cps	cycles per second	OD	outside diameter
crt	cathode-ray tube	OHS	oval head steel
CSK	counter sunk	р	pico, or 10 ⁻¹²
dia	diameter	PHS	pan head steel
div	division	piv	peak inverse voltage
EMC	electrolytic, metal cased	plstc	plastic
EMT	electroyltic, metal tubular	PMC	paper, metal cased
ext	external	poly	polystyrene
f	farad	Prec	precision
F & I	focus and intensity	PT	paper tubular
FHS	flat head steel	PTM	paper or plastic, tubular, molded
Fil HS	fillister head steel	RHS	round head steel
g or G	giga, or 10°	rms	root mean square
Ğe	germanium	sec	second
GMV	guaranteed minimum value	Si	silicon
h	henry	S/N	serial number
hex	hexagonal	t or T	tera, or 1012
HHS	hex head steel	TD	toroid
HSS	hex socket steel	THS	truss head steel
HV	high voltage	tub.	tubular
ID	inside diameter	v or V	volt
incd	incandescent	Var	variable
int	internal	w	watt
k or K	kilohms or kilo (10 ³)	w/	with
kc	kilocycle	w/o	without
m	milli, or 10 ⁻³	ŴŴ	wire-wound
mc	megacycle		

ABBREVIATIONS AND SYMBOLS

SPECIAL NOTES AND SYMBOLS

 000X
 Part removed after this serial number.

 *000-000
 Asterisk preceding Tektronix Part Number indicates manufactured by or for Tektronix, or reworked or checked components.

 Use
 000-000

 Part number indicated is direct replacement.

 Internol screwdriver adjustment.

Front-panel adjustment or connector.

Part first added at this serial number.

X000

Parts List—Type 519

FRONT & REAR



FRONT & REAR

REF.	PART NO	SERIAL//	NODEL NO.	Q	DESCRIPTION
NO.	1441 110.	EFF.	DISC.	Ŷ.	
1 2 3 4	333-0611-00 331-0065-00 337-0396-00 122-0721-00	101 101 101	789 789 789	1 1 1 1	PANEL, front GRATICULE, plexiglass SHIELD, graticule masking ASSEMBLY, camera mount (see ref. 5, 6, 7)
	426-0103-00 354-0110-00 354-0170-00 214-0166-00 211-0009-00 016-0240-00	101 583 790	582	1 1 1 2 1	motion includes: CSTINKS, light seal RNKS, light seal LATCH, strike SCREW, 4 Ado XJ, inch OHS phillips SCREW, 4 Ado XJ, inch OHS phillips mounting hardware. (not included w/assembly)
6 7 8	210-0531-00 210-0527-00 214-0149-00 200-0269-00			4 4 2	NUT, cr. shield retaining BOLT, graticule cover securing COVER, pot
9	210-0013-00 210-0840-00 210-0413-00			7 - 1 1 1	POT mounting hardware for each: (not included w/pot) LOCKWASHER, internal, $\frac{y}{4x} \times \frac{1}{1/4}$, inch WASHER, 390 ID x $\frac{y}{14}$, inch OD NUT, hex, $\frac{y}{6}$ -32 x $\frac{1}{2}$ inch
10	366-0113-00			1	KNOB, small_charcoal—FOCUS knob_includes:
11	175-0117-00		-	1	CABLE, coaxial, plus gate, assembly cable assembly includes:
12	131-0007-00 132-0001-00 132-0007-00 132-007-00 132-007-00 132-0028-00 132-0028-00 132-0028-00 132-0028-00 132-0032-00 132-0116-00 132-0117-00	101 431 101 433 101 431 431 X431	430 430 430		CONNECTOR, coble end, w/hardware NUT, coupling SLEEVE, conductor, outer RINO, snap TRANSITION, outer INSULATOR CONDUCTOR, inner TRANSITION, inner TRANSITION, inner FERRULE FERRULE DISC, tefion
14	366-0113-00 213-0004-00			1	KNOB, small charcoal—INTENSITY knob includes: SCREW, set, 6-32 x ³ / ₁₆ inch HSS
15	213-0004-00				knob includes: SCREW, set, 6-32 x 3/16 inch HSS
16	210-0012-00 210-0207-00 210-0840-00 210-0413-00			3	VOI mounting hardware for each: (not included w/pot) LOCKWASHER, internal, ½ x ½ inch LUC, solder, ¾ inch WASHER, 390 ID x ¼, inch OD NUT, hex, ¾-32 x ½ inch

REF.	PART NO.	SERIAL/I	MODEL NO.	Q	DESCRIPTION
NO.		EFF.	DISC.	Υ.	
17 18 19	385-0041-00 211-0538-00 213-0068-00 366-0113-00 213-0004-00 210-0202-00 210-0202-00	101 225	224	1 1 1 1 1	ROD, nylon mounting hardware: (not included w/rod) SCREW, 6-32 x ¼ ₁ inch, 100° csk, FHS SCREW, thread cutting, 6-32 x ¼ ₁ inch FHS phillips KNOB, small charcoal–SCALE ILUUM. knob includes: SCREW, set, 6-32 x ¼ ₁ inch HSS LUG, solder, SE # 6 mounting hardware: (not included w/lug) NUT, hex, 6-32 x ¼ inch
20 21 22 23	175-0116-00 131-0007-00 132-0007-00 132-0007-00 132-0026-00 132-015-00 132-012-0028-00 132-012-0028-00 132-0118-00 132-0117-00 132-0117-00 132-0117-00 132-0117-00 132-0119-00 211-0038-00	101 431 101 431 101 431 X431	430 430 430	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	CABLE, coaxial, plus gate delay, assembly cable assembly includes: CONINECTOR, cable and, w/hardware NUT, coupling SLEEVE, conductor, outer RINGS, snap TRANSITION, outer TRANSITION, outer INSULATOR CONDUCTOR, inner TRANSITION, inner SCREW, 440 x %, inch FHS phillips
24 25 26	175-0115-00 - 1.0007.00 132.0001.00 132.0001.00 132.0001.00 132.0012.00 132	101 431 101 431 101 431 101 431 101 X431 X431	430 430 430 430	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	CABLE, coaxial, rate generator output, assembly cobie assembly includes: CONNECTOR, cobie end, w/hardware NUT, roupling SLEPYE, conductor, outer RING, snap NUT, retaining RANSITION, outer TRANSITION, outer INSULATOR CONDUCTOR, inner TRANSITION, inner TRANSITION, inner FERRULE FERRULE FERRULE FERRULE

REF.	PART NO.	SERIAL/I	NODEL NO.	Q	DESCRIPTION
			bise.	Υ.	
27 28 29	366-0113-00 213-0004-00 331-0022-00 331-0085-00 132-0041-00 211-0011-00	101 403	402	1 1 1 2 4	KNOB, small charcool—VARIABLE knob includes: SCREW, set, 6-32 x 3/ ₁₆ inch HSS DIAL DIAL, w/brake and charcool knob—VOLTS ADAPTER, panel mounting hardware for each: (not included w/adapter) SCREW, 4-40 x 3/ ₁₆ inch BHS
30	366-0113-00			1	KNOB, small charcoal—FREQUENCY
31	213-0004-00 358-0010-00			1	knob includes: SCREW, set, 6-32 × 3/ ₁₆ inch HSS BUSHING, 3/ ₈ -32 × 7/ ₁₆ inch mounting hardware: (not included w/bushing)
	210-0013-00 210-0429-00			1	LOCKWASHER, internal, $\frac{3}{4} \times \frac{1}{16}$ inch NUT, hex, bushing, $\frac{3}{6}$ -32 x $\frac{1}{2} \times \frac{1}{16}$ inches
32 33	378-0513-00 366-0113-00			1	JEWEL, light, green KNOB, small charcoal—DRIVE knob includes:
34	213-0004-00 366-0117-00			1	SCREW, set, 6-32 x 3/16 inch HSS KNOB, large charcoal—TRIGGER SOURCE
35	213-0004-00 262-0378-00			1	SCREW, set, 6-32 x 3/16 inch HSS SWITCH, wired—TRIGGER SOURCE
	260-0359-00			1	SWITCH, unwired—TRIGGER SOURCE mounting hardware: (not included w/switch)
	210-0013-00 210-0413-00 210-0801-00 210-0004-00 210-0406-00			1 2 2 2	LOCKWASHER, internal, $\frac{3}{2} \times 1^{1}/i_{4}$ inch NUT, hex, $\frac{3}{2} \times 2^{1}$, inch LOCKWASHER, S5 x $\frac{3}{2}$; inch LOCKWASHER, internal, #4 NUT, hex, 440 x $\frac{3}{4}$ inch
36	366-0117-00			1	KNOB, large charcoal—RANGE knob includes:
37	213-0004-00 262-0380-00	101	549	1	SCREW, set, 6-32 x ¾ inch HSS SWITCH, wired—RANGE (see ref. #38) switch includes:
	260-0361-00 262-0611-00	101 550	549	1	SWITCH, unwired—RANGE SWITCH, wired—RANGE (see ref. #38)
38	260-0578-00	550		1	SWITCH, unwired—RANGE mounting hardware: (not included w/switch)
30	210-0013-00 210-0413-00			1	LOCKWASHER, internal, $\frac{1}{2}$ x 11 / ₁₆ inch NUT, hex, $\frac{1}{2}$ -32 x $\frac{1}{2}$ inch

REF.	PART NO	SERIAL/	MODEL NO.	Q	DESCRIPTION
NO.		EFF.	DISC.	Ý.	Descention.
39 40	366-0117-00 213-0004-00 260-0379-00 260-0360-00 179-0486-00			1 1 1	KNOB, large charcoal—MULTIPLIER knob includes: SCREW, set 6-32 x 3/4, inch HSS SWITCH, wired—MULTIPLIER (see ref. #41) switch includes: SWITCH, unwired—MULTIPLIER CABEL HARNESS, multiplier switch
41	210-0013-00 210-0840-00 210-0413-00			1 1 1 1 1 1	mounting hardware: (not included w/switch) LOCKWASHER, internal, ½ x ¹ / ₁ / ₄ inch WASHER, 390 ID x ¹ / ₁ inch OD NUT, hex, ½ 32 x ½ inch KNOR, small charconI—EUNCTION
13	213-0004-00			1	knob includes: SCREW, set, 6-32 x ³ / ₁₆ inch HSS SWITCH wired—FUNCTION (see ref. #44)
44 45	260-0357-00 179-0483-00 210-0013-00 210-0840-00 210-0840-00 210-0801-00 210-0801-00 210-0801-00 210-004-00			1 1 1 1 2 2 2	switch includes: SWITCH, unwirdd-FUNCTION CABE HARNESS, function switch mounting hardware: (not included w/switch) LOCKWASHER, internal, ½ x 1½, inch WASHER, 390 DL x ½, inch WASHER, 350 DL x ½, inch WASHER, 35 ½, inch LOCKWASHER, internal, #4 NUT, hex, 440 x ½, inch
46	366-017-00 213-0004-00 366-017-00 213-0004-00			1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	KNOB, large charcool—CYCLES/SEC knob includes: SCREW, set, 6-32 x 3/μ inch HSS KNOB, large charcool—GAIN knob includes: SCREW, set, 6-32 x 3/μ inch HSS

REF. NO.	PART NO.	SERIAL/I	MODEL NO.	QT	DESCRIPTION
48	262-0377-00			Υ. 1	SWITCH wired—GAIN (see ref #51)
49	260-0358-00 406-0626-00			1	switch includes: SWITCH, unwired—GAIN BRACKET, connector mounting mounting hordware. (est included w/bracket alone)
	211-0008-00 210-0004-00 210-0406-00			2 4 4	SCREW, 4-40 x / ₄ inch BHS LOCKWASHER, internal, #4 NUT, hex, 4-40 x 3/ ₁₆ inch
50	131-0159-00			1	CONNECTOR, printed board, 14 contact, female
51	211-0061-00 210-0004-00 210-0406-00 210-0013-00 210-0413-00 210-0801-00			2 2 2 1 1 2	SCREW, 4-40 x ½, inch 1H B SCREW, 4-40 x ½, inch 1H B LOCKWASHER, internal, #4 NUT, hex, 4-40 x ¼, inch mounting hardware: (not included w/switch) LOCKWASHER, internal, ½ x 1¼, inch NUT, hex, ¼-32 x ½, inch
	210-0004-00 210-0406-00			22	NUT, hex, 4-40 x ³ / ₁₆ inch
52	366-0113-00			1	KNOB, small charcoal—VERNIER SYNC
53	213-0004-00 366-0117-00			1	SCREW, set, 6-32 x 3/16 inch HSS KNOB, large charcoal—HORIZONTAL knob includes:
54	213-0004-00 366-0117-00			1	SCREW, set, 6-32 x 3/16 inch HSS KNOB, large charcoal—DELAY knob includes:
55	213-0004-00 366-0117-00			1 1	SCREW, set, 6-32 x 3/16 inch HSS KNOB, large charcoal—PULSE AMPLITUDE knob includes:
56	213-0004-00 358-0010-00			1 1	SCREW, set, 6-32 x 3/16 inch HSS BUSHING, 3/6-32 x 3/16 inch mounting berdware, loot included w/bushing)
57	219-0013-00 210-0429-00 210-0012-00 210-0207-00			1 1 1 1	LOCKWASHER, internal, $\frac{3}{4} \times \frac{3}{16}$ into LockWaSHER, internal, $\frac{3}{4} \times \frac{3}{16}$ into LockWaSHER, internal, $\frac{3}{4} \times \frac{3}{2} \times \frac{3}{2} \times \frac{3}{2} \times \frac{3}{2}$ inches LOCKWASHER, internal, $\frac{3}{4} \times \frac{3}{2}$ inches LUG, solder, $\frac{3}{4}$ inch
58	260-0212-00 260-0447-00	101 403	402	1	SWITCH, slide—SWEEP DIODE SWITCH, slide—SWEEP DIODE mounting hardware: (not included w/switch)
	210-0406-00			2	NUT, hex, 4-40 x ³ / ₁₆ inch
59	366-0115-00			1	KNOB, large charcoal—NANOSEC/CM knob includes:
	213-0004-00			1	SCREW, set, 6-32 x ¾, inch HSS

REF.	PART NO.	SERIAL/I	NODEL NO.	Q	DESCRIPTION
1.0	0.40.0075.00	101	1/1	Y.	
60	262-0375-00 262-0460-00 262-0453-00 262-0500-00	101 162 278 303	161 277 302	1 1 1	SWITCH, wired—NANOSEC/CM (see ref. #67) SWITCH, wired—NANOSEC/CM (see ref. #67) SWITCH, wired—NANOSEC/CM (see ref. #67)
61	260-0356-00 260-0385-00 406-0631-00	101 162	161	1 1 1	SWITCH includes: SWITCH, unwired—NANOSEC/CM SWITCH, unwired—NANOSEC/CM BRACKET, sweep switch, front moutine, bactware, load included w/bracket)
62 63	210-0413-00 210-0012-00 387-0313-00			1	NUT, hex, ³ / ₈ -32 x ¹ / ₂ inch LOCKWASHER, internal, ³ / ₈ x ¹ / ₂ inch PLATE, coaxial, clamping
	211-0012-00 210-0004-00 210-0406-00			- 4 4 4	mounting hardware: (not included w/plate) SCREW, 4.40 x 1/6 inch BHS LOCKWASHER, internal, #4 NUT, hex, 4.40 x 3/16 inch
64	210-0202-00 211-0504-00 210-0407-00			1 - 1 1	LUG, solder, SE #6 mounting hardware: (not included w/lug) SCREW, 632x V/, inch BHS NUT, hex, 6-32x V, inch
65	387-0336-00 387-0389-00	101 162	161	1	PLATE, circuit board PLATE, circuit board mounting hardware. (not included w/plate) LOCKWASHER internal #4
	210-0406-00			2	NUT, hex, 4-40 x $^{3}/_{16}$ inch
66	131-0172-00 131-0507-00 210-0202-00 210-0407-00	X162 1080	1079	4 - 1 1	CONNECTOR, jack JACK, inj mouning hardware for each: (not included w/connector) LUG, solder, SE # 6 NUT, hex, 6-32 x 1/4, inch
67	211-0507-00 210-0803-00			- 4 4	mounting hardware: (not included w/switch) SCREW, 6-32 x 5/16 inch BHS WASHER, 6L x 3/6 inch
68	376-0007-00			1	COUPLING, aluminum coupling includes:
69 70	384-0225-00 376-0005-00			1	ROD, extension COUPLING, aluminum mounting hardware: (not included w/coupling)
	213-0005-00			2	SCREW, set, 8-32 x 1/8 inch HSS
71 72	384-0574-00 358-0029-00 210-0013-00 210-0413-00			1 1 1 1	ROD, coupling BUSHING, hex 1/4-32 x ¹³ / ₃₂ inch mounting hardware: (nat included w/bushing) LOCKWASHER, internal, ¹ / ₄ x ¹ / ₁₄ inch NUT, hex, ³ / ₄ -32 x ¹ / ₂ inch

FRONT & REAR (Cont'd)

REF.	PART NO.	SERIAL/M	ODEL NO.	Q	DESCRIPTION
NO.		EFF.	DISC.	Y.	
73	260-0247-00 210-0583-00 210-0940-00			1	SWITCH, push button—RESET mounting hardware: (not included w/switch) NUT, hex, $1_{4-32} \approx 1_{16}$ inch WASHER, 1_{4} ID $\approx 3_{6}$ inch OD
74	366-0117-00			1	KNOB, large charcoal—VERTICAL knob includes:
75	213-0004-00 352-0008-00 352-0067-00 378-0541-00	101 820 X820	819	1 1 1	SCREW, set, 6-32 x 3/16 inch HSS HOLDER, neon bulb, single, black HOLDER, neon bulb, single, gray FILTER, lens, neon
	211-0031-00 211-0109-00 210-0406-00	101 820	819	1 1 2	mouning hardware: (not included w/holder) SCREW, 440 x 1 inch FHS SCREW, 440 x 7 ₆ inch FHS NUT, hex, 440 x ³ / ₆ inch
76	616-0413-00			1	ASSEMBLY, circuit board, trigger amplifier assembly includes:
77	387-0316-00 136-0095-00			1 2	PLATE, printed circuit board SOCKET, 4 pin transistor
	213-0113-00			2	mounting hardware for each: (not included w/socket alone) SCREW, thread forming, 2-32 x 5/16 inch RHS phillips
78 79	136-0047-00 260-0145-00 260-0449-00	101 403	402	1 1 1 - 2	SOCKET, light, w/red jewel and mounting nut SWTCH, slide—POLARITY SWTCH, slide—POLARITY mounting hardware: (not included w/switch NUT, hex 4-40 s-V ₀ , incl
80 81	136-0026-00 132-0052-00 211-0038-00			1	SOCKET, light FLANGE, outer tube mounting hardware: (not included w/flange) SCREW, 4-40 x % ₁₆ inch FHS phillips
82 83	260-0199-00 260-0506-00 210-0414-00 354-0055-00 210-0902-00 210-0473-00	101 403	402	1 1 1 1 1	SWITCH, toggle—POWER ON SWITCH, toggle—POWER ON mounting hardwate: (not included w/switch) NUT, hex. ¹ /g3-32 x/ _k , inch RING, locking, switch WASHER, A/70 ID x ³ / ₂ , inch OD NUT, switch, ¹⁹ / ₃₂ -32 x 4/ ₄₄ inch, 12 sided
84 85	337-0428-00 337-0563-00 175-0119-00	101 403	402	1	SHIELD, toggle switch SHIELD, toggle switch CABLE, coaxial, 125 Ω, twin lead, assembly
86	132-0001-00 132-0002-00 132-001-00 132-0014-00 132-0015-00 132-0015-00 132-015-00 132-015-00	101 431	430	1 1 2 1 1 1 1	coble assembly includes: NUT, coupling SLEVE, conductor, outer RING, snap INSULATOR SLEVE, transition CONTACT, assembly, 125 Ω TRANSITION, outer TRANSITION, outer

REF

NO

Ť Y. 87 132-0016-00 504 3 NUT, retaining 101 132-0121-00 505 3 NUT, retaining 88 660-0402-00 i ASSEMBLY, trigger take-off, long assembly includes: 132-0067-00 1 BARREL, trigger take-off, long 132-0068-00 i BARREL, trigger take-off, short TUBE, spacer, nylon 166-0182-00 1 354-0108-00 1 RING, housing 89 387-0330-00 2 PLATE, end, housing mounting hardware: (not included w/plate alone) 211-0020-00 4 SCREW, 4-40 x 11/a inches RHS 210-0004-00 4 LOCKWASHER, internal, #4 210-0406-00 NUT, hex, 4-40 x 3/16 inch 4 90 132-0001-00 2 NUT, coupling 91 132-0002-00 2 SLEEVE, conductor, outer 132-0007-00 2 RING, snap CONTACT, 125 Q, assembly 92 132-0015-00 2 132-0014-00 2 SLEEVE, transition CONDUCTOR, inner, long 132-0059-00 93 132-0011-00 4 INSULATOR 94 132-0016-00 101 504 1 NUT, retaining 132-0121-00 505 1 NUT, retaining 95 1 GROMMET, 3/8 inch 348-0004-00 96 387-0312-00 1 PLATE, bottom mounting hardware: (not included w/plate) 212-0039-00 13 SCREW, 8-32 x 3/ inch THS phillips 97 1 PLATE, cabinet side, right 387-0302-00 plate includes: 134-0058-00 1 PLUG, steel 98 214-0057-00 6 FASTENER, cabinet latch, assembly each fastener assembly includes: 213-0033-00 1 SCREW, fastening 210-0847-00 1 WASHER, nylon, .164 ID x .500 inch OD 105-0007-00 i STOP NUT, latch, nylon 99 210-0480-00 1 100 I 122-0090-00 1 ANGLE, frame, right mounting hardware: (not included w/angle) 212-0039-00 4 SCREW, 8-32 x 3/8 inch THS phillips 210-0458-00 4 NUT, keps, 8-32 x 11/32 inch

FRONT & REAR (Cont'd)

DESCRIPTION

SERIAL/MODEL NO

DISC

EFF.

REF.	PART NO.	SERIAL/	NODEL NO.	Q	DESCRIPTION
NO.		EFF.	DISC.	Ý.	
101	660-0401-00			1	ASSEMBLY, 125 Ω pulser (see ref. #114 and 115) assembly includes: TUBE, outer, 125 Ω line
102 103	166-0197-00 132-0002-00 132-0011-00			1 4 8	TUBE, spacer, assembly SLEEVE, conductor, outer INSULATOR
104 105	132-0014-00 132-0015-00 132-0041-00 132-0061-00 166-0195-00 260-0362-00 260-0693-00	101	610	2 4 1 18 18 1	SLEEVE, transition CONTACT, 125 D, assembly ADAPTER, panel CONDUCTOR, center TUBE, spacer, styrofoam SWITCH, reed
106 107 108 109 110 111 112 113 114	132.0047.00 132.0014.00 132.0016.00 332.0016.00 343.0063.00 132.0048.00 132.0048.00 132.0048.00 132.0050.00 132.0050.00 211.0510.00 211.0510.00 211.0012.00	101 611	610	2 2 3 1 2 4 1 1 1 1 1 2 2 4	CONNECTOR, front switch CONNECTOR, front switch NUT, retraining CLAMP, tube, outer RING, nap EBOW, inner EBL, outer COLLAR, hold down NUT, elbow mounting mounting, hardware: (not included w/assembly) SCREW, 402 x ½; inch SCREW, 402 x ½; inch SCREW, 404 y ½; inch SCREW, 404 y ½; inch
114	244 0049 00				
	213-0012-00			2	mounting hardware: (not included w/clip) SCREW, thread cutting 4-40 x % inch FHS
117 118 119	650-0440-00 387-0025-00 331-0015-00 131-0102-00 129-0041-00 200-0185-00 210-0053-00 210-00551-00 211-0078-00 377-0041-00 386-0933-00			1 2 1 1 1 2 2 1 2 1 2 1 2	ASSEMBLY, AC line filter assembly includes: PLATE, line filter cap SPACER, line filter CONNECTOR, chassis mounted, motor base (see ref. #120-122 connector includes: POST, ground COVER, motor base LOCKWASHER, external, #4 NUT, hex, 440 x ½, inch SCREW, 440 x ½, inch SCREW, 440 x ½, inch SCREW, Loback urea PLATE, motor base mounting hardware: (not included w/assembly)
120 121 122	211-0552-00 385-0146-00 211-0507-00			2 2 2	SCREW, 6-32 x 2 inches BHS ROD, hex SCREW, 6-32 x \$/16 inch BHS
123	122-0089-00 212-0039-00 210-0458-00			1 -4 4	ANGLE, frame, left mounting hardware: (not included w/angle) SCREW, 8-32 x ½; inch THS NUT, keps, 8-32 x ¹¹ / ₅₂ inch

REF.	PART NO	SERIAL/N	NODEL NO.	Q	DESCRIPTION
NO.	PARI NO.	EFF.	DISC.	Ŷ.	,
124 125 126	387-0306-00 387-0306-01 200-0237-00 387-0304-00 387-0741-00	101 101 403	1190 574X 402	1 1 1 1	PLATE, rear sub-panel PLATE, rear sub-panel COVER, insulation, fuse holder PLATE, rear overlagy PLATE, rear overlagy
127	213-0041-00			4	SCREW, thread cutting, $6-32 \times \frac{3}{8}$ inch THS
128	334-0649-00 213-0088-00			1 - 2	TAG, voltage rating mounting hardware: (not included w/tag) SCREW, thread forming, 4-40 x ¼ inch PHS
129 130 131 132 133 134 135	352-0002-00 200-0582-00 210-0873-00 378-0011-00 380-0018-00 212-0031-00 210-0458-00 210-0458-00 210-0402-00			1 1 1 1 1 2 2 2	HOLDER, fuse, assembly holder assembly includes: HOLDER, fuse CAP, fuse CAP, fuse HUER, rubber, 1/2 ID x ¹ /1 ₄ inch OD HUER, and the second second second HOLSE, air filter mounting hardware: (not included w/housing) SCEEW, 8-32 x ¹ /2 ₄ inches RHS NUT, keps, 8-32 x ¹ /2 ₅ inch NUT, cap, hex, 8-32 x ⁴ / ₁₄ inch
136	387-0333-00			1	PLATE, air intake
137	212-0023-00 131-0267-00	X403		6	mounting hardware: (not included w/plate) SCREW, 8-32 × ³ / ₈ inch BHS CONNECTOR, conductor, phone jack connector includes:
138	210-0413-00 210-0012-00 343-0002-00 211-0511-00 210-0803-00 210-0457-00			1 1 1 1 1 1	NUT, hex, ½-32 x ½, inch LOCKWASHER, internal, ½ x ½ inch CLAMP, cable, ¾ inch mounting hardware: (not included w/clamp) SCREW, 6-32 x ½ inch BHS WASHER, dx ¾ inch NUT, keps, 6-32 x ¾ inch
139	406-0663-00 210-0803-00 210-0457-00			2 - 2 2	BRACKET, crt shield mounting mounting hardware for each: (not included w/bracket) WASHER, dx ½, inch NUT, keps, 6-32 x 1/4, inch
140	124-0138-00 210-0457-00 210-0202-00			2 - 1 1	STRIP, crt, ground mounting hardware for each: (not included w/strip) NUT, keps, d32x \$/4, inch LUG, solder, SE #6
141	136-0100-00 211-0012-00 210-0201-00 210-0004-00 210-0406-00			1 2 2 2 2 2	SOCKET, crt, 14 pin mounting hardware: (not included w/socket) SCREW, 4-40 x-74; inch BHS LUG, solder, SE #4 LOCKWASHER; internal, #4 NUT, hex, 4-40 x % inch

REF.	PART NO	SERIAL/	AODEL NO.	9	DESCRIPTION
NO.	TAKI NO.	EFF.	DISC.	Y.	DESCRIPTION
142	426-0097-00 211-0507-00 210-0803-00			1 - 2 2	MOUNT, fan motor and blower housing mounting hardware: (not included w/mount) SCREW, 632x 5%, inch BHS WASHER, 6L x 3% inch
143 144 145 146 147 148 149	179-0438-00 369-0019-00 650-0439-00 384-0567-00 384-0568-00 384-0572-00 384-0572-00 131-0007-00 212-0029-00			1 1 5 2 2 1 3 1	CABLE HARNESS, focus and intensity FAN, blower wheel ASSEMBLY, clipping line drum (see ref. #150) assembly includes: PLATE, clipping line drum ROD, spocing, J ₄ , inch ROD, spocing, J ₄ , inch CONNECTOR, Robe end SCREW, 8-32 ×3 inches HHS
150	210-0458-00			1	NUT, keps, 8-32 x ¹ / ₃₂ inch mounting hardware: (not included w/assembly) NUT, keps, 8-32 x ¹ / ₃₂ inch
151 152	147-0024-00 212-0004-00 136-0094-00			1 - 6 1	MOTOR, fan mounting, hardware: (not included w/motor) SCREW, 8-32 %, inch BHS SOCKET, miniature jack w/hardware
153 154	406-0627-00 136-0035-00 211-0507-00 210-0803-00 210-0006-00 210-0407-00			1 2 1 1 1 1	BRACKET, graticule lamp holding SOCKET, graticule light mounting hardware for each: (not included w/socket) SCREW, 6-32 x $\frac{9}{16}$ inch BHS WASHER, 64 x $\frac{9}{6}$ inch LOCKWASHER, internal, #6 NUT, hex, 6-32 x $\frac{1}{6}$ inch
155	387-0311-00			1	PLATE, top assembly
156	367-0040-00 367-0011-00			2	plare includes: ASSEMBLY, handle each assembly includes: HANDLE
157	343-0073-00 211-0507-00 210-0457-00 212-0039-00			1 6 6 10	CLAMP, cover, handle SCREW, 632 */4, inch BHS NUT, keps, 632 */3, inch mounting hardware: (not included w/plate) SCREW, 8-32 x % inch THS
158	387-0303-00			1	PLATE, cabinet side, right
159	214-0057-00			6	Plate includes: FASTENER, cabinet latch, assembly
160	213-0033-00 210-0847-00 105-0007-00 210-0480-00 387-0308-00 387-0308-01	101 1191	1190	1 1 1 1 1	each tastening assembly includes: SCREW, fastening WASHER, nylon, .164 ID x.500 inch OD STOP NUT, latch, nylon PLATE, front sub-panel PLATE, front sub-panel

CHASSIS



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REF. NO.	PART NO.	SERIAL/I EFF.	NODEL NO. DISC.	Q T Y	DESCRIPTION
1 2 3 4	387-0346-00 124-0128-00 348-0003-00 			1 2 4 2 1 1 1	PLATE, ground STRIP, 50 0; bronze GROMMET, %i, sinch POT Wommen bardware for each: (not included w/pot) LUG, solder, 1/, inch LOCKWASHER, internal, 400 OD x 261 inch ID NUT, hex, 1/, 23 x %i, inch
5	136-0095-00 211-0062-00 210-0001-00 210-0405-00			2 - 2 2 2	SOCKET, 4 pin transistor mounting hardware for each: (not included w/socket) SCREW, 2-56 x ¾, inch RHS LOCKWASHER, internal, #2 NUT, hex, 2-56 x ¾, inch
6	386-0253-00 211-0534-00 210-0006-00 210-0407-00			1 2 2 2	PLATE, metal, small capacitor mounting hardware for each: (not included w/plate) SCREW, 6-32 x ½, inch PHS w/lockwasher LOCKWASHER, internal, #6 NUT, hex, 6-32 x ½ inch
7	348-0006-00			2	GROMMET, ¾ inch
8	386-0252-00 211-0534-00 210-0006-00 210-0407-00	T.		2 - 2 2 2	PLATE, fiber, small capacitor mounting hardware for each: (not included w/plate) SCREW, 6-32 x %, inch LOCKWASHER, internal, #6 NUT, hex, 6-32 x ½ inch
9 10	200-0256-00 210-0006-00 210-0407-00			1 1 2 2	COVER, capacitor CAPACITOR, timing mounting hardware: (not included w/capacitor) LOCKWASHER, internal, #6 NUT, hex, 6-32 x 1/4 inch
11 12	212-0524-00 210-0812-00 210-0010-00 210-0564-00 210-0201-00			4 4 4 10	SCREW, 10-32 x 31/c inches HHS WASHER, fiber, #10 LOCKMASHER, internol, #10 NUT, hex, 10-32 x 1/g inch LUG, solder, SE #4 mounting hardware for each: (not included w/lug)
13	213-0044-00 260-0246-00 260-0336-00 213-0044-00	101 263	262	1	SURTER, intered cutting, 5-32 x 7/ ₁₆ inch PhS phillips SWITCH, thermal cutout SWITCH, thermal cutout mounting hardware: (not included w/switch) SCREW, thread cutting, 5-32 x 3/ ₁₆ inch PHS phillips

CHASSIS (Cont'd)

REF. NO.	PART NO.	SERIAL/M EFF.	ODEL NO. DISC.	Q T	DESCRIPTION
14 15	348-0004-00 211-0553-00 210-0601-00 210-0478-00 211-0507-00			5 5 1 1 1	GROMMET, 1/6 inch RESISTOR, 10 watt mounting hardware for each: (not included w/resistor) SCREW, 632 x 1/5, inches RHS phillips EYELET NUT, hex, resistor mounting SCREW, 6-32 x 1/6, inch BHS
16	212-0037-00 210-0008-00 210-0809-00 210-0462-00 212-0004-00			1 1 1 1 1	RESISTOR, 25 watt mounting hardware: (not included w/resistor) SCREW, 832 11/2, inches Fil HS LOCKWASHER, internal, #8 WASHER, resistor cantering NUT, hex, resistor mounting SCREW, 8-32 x %, inch BHS
17	212-0037-00 210-0808-00 210-0462-00 212-0004-00			2 1 1 1 1	RESISTOR, 20 watt mounting hardware for each: (not included w/resistor) SCREW, 832 11/, inches Fil HS WASHER, resistor centering NUT, hex, resistor mounting SCREW, 8-32 x ⁴ / ₁₄ , inch BHS
18	441-0341-00 212-0004-00 212-0040-00 210-0458-00			1 - 4 6 3	CHASSIS, power mounting hardware: (not included w/chassis) SCREW. 8-32 × 1/4, inch BHS SCREW, 8-32 × 1/4, inch FHS phillips NUT, keps, 8-32 × 1/32 inch
19	210-0840-00 210-0413-00			1 1 1	POT mounting hardware for each: (not included w/pot) WASHER, .390 ID x %;s inch OD NUT, hex, ¾-32 x ½ inch
20 21 22	200-0261-00 200-0257-00 343-0001-00 211-0510-00 210-0803-00 210-0803-00 210-0006-00 210-0407-00			3 2 3 1 1 1 1	COVER, capacitor COVER, capacitor COVER, capacitor CLAMP, cable, J& inch mounting hardware for each: (not included w/clamp) SCREW, 6-32 x ½, inch BHS WASHER, 6L x ½, inch LOCKWASHER, internal, #6 NUT, hex, 6-32 x ¼, inch
CHASSIS (Cont'd)

REF.	PART NO.	SERIAL/I	NODEL NO. DISC.	Q T	DESCRIPTION
23	200.0293.00			1	COVER capacitor
24	406-0644-00 212-0004-00 166-0098-00 212-0008-00 212-0010-00 210-0804-00 166-0030-00	101 580 101 580	579 579	1 2 2 2 2 2 2 2 2 2 2	BRACKET, capacitor mounting mounting hardware: (not included w/bracket) SCREW, 8-32 x 1/s, inch BHS SCREW, 8-32 x 1/s, inch BHS SCREW, 8-32 x 1/s, inch BHS WASHER, 85 x 1/s, inch TUBE, spacer
25	386-0254-00 211-0534-00 210-0006-00 210-0407-00			1 - 2 2 2	PLATE, fiber, large capacitor mounting hardware: (not included w/plate) SCREW, 6-32 × Y ₁₆ inch, PHS w/lockwasher LOCKWASHER, internal, ²⁴ NUT, hex, 6-32 × Y ₄ inch
26	386-0255-00 211-0534-00 210-0006-00 210-0407-00			3 - 2 2 2	PLATE, metal, large capacitor mounting hardware for each: (not included w/plate) SCREW, 6-32 × 1/ ₆ , inch, PHS w/lockwasher LOCKWASHER, internal, #6 NUT, hex, 6-32 × 1/ ₄ inch
27	211-0504-00			2 - 2	RELAY mounting hardware for each: (not included w/relay) SCREW, 6-32 x $^{\prime}\!$
28 29	348-0005-00 386-0254-00 211-0543-00 210-0006-00 210-0407-00			4 2 2 2 2	GROMMET, V_2 inch PLATE, fiber, large capacitor mounting hardware for each: (not included w/plate) SCREW, 6-32 x V_{i4} inch, RHS LOCKWASHER, internal, # 6 NUT, hex, 6-32 x V_4 inch
30	136-0015-00 213-0044-00			15 - 2	SOCKET, STM9G mounting hardware for each: (not included w/socket) SCREW, thread cutting, 5-32 x ϑ_{14} inch, PHS phillips
31	136-0008-00 213-0044-00			8 - 2	SOCKET, STM7G mounting hardware for each: (not included w/socket) SCREW, thread cutting, 5-32 x $\frac{3}{16}$ inch, PHS phillips
32	136-0011-00 211-0538-00 210-0006-00 210-0407-00			3 2 2 2	SOCKET, STM8G mouning hardware for each: (not included w/socket) SCREW, 632 v 1/4, inch, FHS phillips LOCKWASHER, internal, #6 NUT, hex, 632 x 1/4, inch

CHASSIS (Cont'd)

REF.	F. PART NO.	SERIAL/MODEL NO.		Q	DESCRIPTION
NO.		EFF.	DISC.	Y.	
33	210-0202-00 211-0504-00 210-0407-00			2	LUG, solder, SE #6 mounting hardware for each: (not included w/lug) SCREW, 6-32 x /4, inch, BHS NUT, hex, 6-32 x 1/4, inch
34 35	406-0623-00 212-0004-00 212-0040-00			1 2 1	This hardware is a sub-part of the diode. BRACKET, transformer, front mounting hardware: (not included w/bracket) SCREW, 8-32 × ½, inch, BHS SCREW, 8-32 × ½, inch, FHS phillips
36	406-0622-00 212-0004-00 212-0040-00			1 - 2 1	BRACKET, transformer, rear mounting hardware: (not included wbracket) SCREW, 8-32 × ½, inch, BHS SCREW, 8-32 × ½, inch, FHS phillips
37	406-0632-00 211-0507-00 210-0803-00			1 2 2	BRACKET, switch mounting mounting hardware: (not included w/bracket) SCREW, 632 */4; inch, BHS WASHER, 8L * ½ inch
38	352-0031-00 211-0507-00 210-0457-00	X552		1 1 1	HOLDER, fuse, single mounting hardware: (not included w/holder) SCREW, $632 x M_{\rm pl}$ inch, BHS NUT, keps, $6.32 x M_{\rm pl}$ inch
39	387-0337-00 211-0507-00 210-0803-00			1 6 6	PLATE, air duct end mounting hardware: (not included w/plate) SCREW, 632 * 1/4; inch, BHS WASHER, 6L * 1/4; inch
40	352-0031-00 211-0025-00 210-0004-00 210-0406-00	X403		1 1 1 1	HOLDER, fuse, single mounting hardware: (not included w/holder) SCREW, 4-40 x 3/6 inch, FHS LOCKWASHER, internal, #4 NUT, hex, 4-40 x 3/6 inch
41	441-0343-00 211-0507-00 210-0803-00		×	1 - 4 2	CHASSIS, voltage regulator mounting hardware: (not included w/chassis) SCREW, 6-32 x ³ / ₄ inch, BHS WASHER, 6L x ³ / ₈ inch

Parts List—Type 519

CHASSIS (Cont'd)

REF.	PART NO.	SERIAL/M	NODEL NO.	Q T	DESCRIPTION
NO.		EFF.	DISC.	Ŷ.	
42	387-0345-00 129-0049-00 211-0511-00 210-0900-00 210-0080-00 210-0008-00 210-0006-00 210-0020-00 210-0409-00 210-0407-00	101 540 101 540 540 101 540	539 539 539	2 1 2 2 2 2 2 1 1 2 2 2 2 1 2 2 2 1 2 2 2 2 2 1 2	TRANSISTOR mounting hardware for each: (not included w/transistor) PLATE, insulator POST, terminal, transistor mounting SCREW, 632 x / ₂ inch, BHS WASHER, bakelite, transistor mounting WASHER, 64 x / ₂ inch LOCKWASHER, internal, #8 LOCKWASHER, internal, #6 LUCK, solder, SE #6 NUT, hex, 832 x / ₄ , inch NUT, hex, 632 x / ₄ inch
44 45	386-0689-00 210-0813-00 210-0805-00 210-0206-00 210-0410-00			2 1 1 1 1 1	TRANSISTOR mounting hardware for each: (not included w/transistor) PLATE, micz, transistor insulator WASHER, fiber, #10 WASHER, 108 × 1/4, IUG, solder, SE #10, long NUT, hex, 10-32 × 1/4, inch
46	210-0202-00 211-0507-00 210-0407-00			2 1 1	LUG, solder, SE #6 mounting hardware for each: (not included w/lug) SCREW, 6-32 x %is inch, BHS NUT, hex, 6-32 x ¼ inch
47 48 49 50	175-0118-00 132-0001-00 132-0007-00 132-001-00 132-0014-00 132-0015-00 132-015-00 132-015-00 132-015-00 132-012-00 132-012-00 132-0016-00 211-0008-00	101 431 101 505	430 504	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	CABLE, coaxial, 125 Ω, twin lead, assembly cable assembly includes: NUT, coupling SLEEVE, conductor, outer RINOS, nap INSULATOR SLEEVE, transition CONTACT, assembly TRANSITION, outer TRANSITION, outer NUT, retaining ADAPTER, panel mounting hardware: (not included w/adapter) SCREW, 4-40 × ¼, inch, BHS
51	387-0309-00 212-0040-00 210-0458-00			1 - 8 8	PLATE, bulkhead mounting hardware: (not included w/plate) SCREW, 823-2% inch, FHS phillips NUT, keps, 8-32 x ¹¹ / ₃₂ inch

REF.	PART NO	SERIAL/MODEL NO.		Q	DESCRIPTION	
NO.	TAKT NO.	EFF.	DISC.	Ŷ.	DESCRIPTION	
52	660-0403-00 354-0108-00 132-0058-00 132-0058-00 166-0182-00 166-0182-00 387-0330-00 387-0330-00 211-0020-00 210-0004-00 210-0406-00			1 1 2 1 2 1 2 4 4 4	ASSEMBLY, trigger take-off short assembly includes: RING, housing CONDUCTOR, inner, short BARREL, trigger take-off TUBE, compression sleeve PLATE, end, housing mounting hardware for each (not included w/plate alone) SCREW, 4-40 x 1/4, inches, RHS LOCKWASHER, internal, #4 NUT, hex, 4-40 x ¹ /4, inch	
55 56 57 58 59	132-0001-00 132-0002-00 132-0011-00 132-0011-00 132-0015-00 204-0059-00 210-0530-00 132-0043-00 132-0043-00 132-0007-00 132-0007-00 132-0007-00 132-0011-00 132-0012-00 132-0012-00			2 2 2 4 2 2 4 2 2 1 1 1 1 1 1 1 1 1 1 2 1 1	NUT, coupling SLEEVE, conductor, outer RING, snap INSULATOR SLEEVE, transition CONTACT, 125 Q, assembly BODY, tubing union NUT, 7/4z/24 x ⁷ /4, inch NUT, coupling GUARD, cord, rubber NUT, coupling SLEEVE, conductor, outer RING, snap INSULATOR CONTACT, 125 Q, assembly CONTACT, 125 Q, assem	
60	132-0038-00 343-0014-00 211-0507-00 210-0803-00			1 1 1 1	FERRULE CLAMP, coble, 1 inch mounting hardware: (not included w/clamp) SCREW, 6-32 x \mathscr{Y}_{i_d} inch, BHS WASHER, 6L x \mathscr{Y}_{g} inch	
61 62 63	348-0012-00 017-0007-00 017-0051-00 385-0080-00 211-0507-00	101 550	549	2 1 7 1	GROMMET, $\frac{y_{\rm B}}{100}$ inch TERMINATION, 125 Ω TERMINATION, 125 Ω ROD, hex mounting hardware for each: (not included w/rod) SCREW, 6-32 x $\frac{y_{\rm H}}{100}$ inch, BHS	
64	441-0342-00 211-0504-00			1 5	CHASSIS, high voltage mounting hardware for each: (not included w/chassis) SCREW, 6-32 x ¼ inch, BHS	
65	384-0569-00 211-0507-00			4	ROD, spacer, high voltage box mounting hardware for each: (not included w/rod) SCREW, 6-32 x 1/14 inch, BHS	

CHASSIS (Cont'd)

REF	PART NO.	SERIAL/	NODEL NO.	Q T	DESCRIPTION
NO		EFF.	DISC.	Y.	
66	343-0064-00 211-0507-00 210-0803-00 212-0033-00 210-0008-00 210-0409-00			1 3 3 1 1 1	CLAMP, capacitor mounting mounting hardware: (not included w/clamp) SCREW, 632 ¥/4; inch, BHS WASHER, 6L x ½, inch SCREW, 832 x ½, inch LOCKWASHER, internal, #8 NUT, hex, 8-32 x ½; inch
67	352-0002-00 352-0010-00 200-0582-00 210-0873-00			1	HOLDER, fuse, assembly holder assembly includes: HOLDER, fuse CAP, fuse WASHER, rubber, ½ ID x ¹¹ / ₁₆ inch OD
68	210-0207-00 210-0012-00 210-0840-00 210-0413-00			1 1 1 1 1 1 1	POT not induce included w/pot) LUG, solder, ½ inch LUG, solder, ½ inch LOCKWASHER, internal, ½ × ½ inch WASHER, 390 ID × ½ inch OD NUT, hex, ½-32 × ½ inch
69	136-0011-00 211-0534-00 210-0006-00 210-0407-00			1 2 2 2	SOCKET, STM8G mounting hardware: (not included w/socket) SCREW, 6-32 × 1/4; inch, PHS w/lockwasher LOCKWASHER, internal, #6 NUT, hex, 6-32 × 1/4; inch
70	343-0008-00 211-0510-00 210-0803-00 210-0006-00 210-0407-00			1 1 1 1 1	CLAMP, cable, ¼ inch mounting hardware: (not included w/clamp) SCREW, 632 x¼, inch, BHS WASHER, 6L x¼, inch LOCKWASHER, internal, # 6 NUT, hex, 6-32 x ¼, inch
71	136-0015-00 211-0033-00 210-0004-00 210-0406-00			1 2 2 2	SOCKET, STM9G mounting hardware: (not included w/socket) SCREW, 4-40 x */1 ₄ inch, PHS w/lockwasher LOCKWASHER, internal, #4 NUT, hex, 4-40 x */1 ₄ inch
72	212-0001-00 210-0205-00		r.	1 - 2 2	CAPACITOR, ceramic, (not shown) mounting hardware: (not included w/capacitor) SCREW, 8-22 x /4, inch, BHS LUG, solder, SE #8
73	212-0001-00 211-0504-00 210-0205-00 210-0202-00	101 482 101 482	481 481	2 2 2 2 2 2	CAPACITOR, ceramic mounting hardware for each: (not included w/capacitor) SCREW, 8-32 × 1/, inch, BHS SCREW, 6-32 × 1/, inch, BHS LUG, solder, SE #8 LUG, solder, SE #6

CHASSIS (Cont'd)

REF.	PART NO	SERIAL/MODEL NO.		P P	DESCRIPTION	
NO.		EFF.	DISC.	Ý.	BESCHITTON	
74	210-0204-00 211-0507-00 210-0407-00			1 - 1 1	LUG, solder, DE #6 mounting hardware: (not included w/lug) SCREW. 632x 1/4; inch, BHS NUT, hex, 632x 1/4 inch	
75 76	136-0074-00 354-0109-00 211-0040-00			3 - 1 2	SOCKET, 9 pin, UHF mounting hardware for each: (not included w/socket) RING, socket retaining SCREW, 4-40 x 1/4, inch, BH nylon	
77	202-0059-00 211-0507-00			1 - 4	BOX, high voltage power supply, top mounting hardware: (not included w/box) SCREW, 6-32 x \$/16 inch, BHS	
78 79 80	131-0161-00 660-0404-00 132-0055-00 132-0062-00 211-0060-00			1 1 1 1 2	CONNECTOR, terminal feed-thru ASSEMBLY, copacitor charger network assembly includes: BODY, charging network PLATE, end, charging mounting hardware: (not included w/plate alone) SCREW, 2-56 x 7/16 inch, FHS	
81 82 83	131-0162-00 132-0001-00 132-002-00 132-007-00 132-001-00 132-0012-00 337-0395-00 211-0504-00			1 1 1 2 1 1 5	CONNECTOR, terminal test jack NUT, coupling SLEEVE, conductor, outer RINGS, snap INSULATOR CONTACT, 125 Ω, assembly SHIELD, high voltage mounting hardware: (not included w/shield) SCREW, 6-32 x ¼, inch, BHS	
84	346-0010-00 210-0004-00 210-0406-00			1 - 2 2	STRAP, mounting, high voltage transformer mounting hardware: (not included w/strap) LOCKWASHER, internal, #4 NUT, hex, 4-40 x 3/14 inch	
85	276-0060-00 202-0058-00			4 1	FORM, plastic (not shown) BOX, high voltage power supply	

REF.	PART NO.	SERIAL/I	NODEL NO.	Q	DESCRIPTION
NO.		EFF.	DISC.	Ŷ.	
86 87	119-0011-00 210-0522-00			1 - 2	DELAY LINE (see ref. #91) delay line includes: NUT, tapered sleeve and holding sleeve
88 89	132-0054-00 132-0001-00 132-0002-00			222	NUT, coaxial, delay line NUT, coupling SLEEVE, conductor, outer
	132-0007-00 132-0011-00 132-0014-00			2 4 2	RING, snap INSULATOR SLEEVE, transition
90 91	132-0013-00 132-0053-00 361-0023-00			2 2	CONDUCTOR, tapered, delay line SPACER, delay line, plastic mounting hardware: (not included w/delay line)
	212-0023-00 210-0804-00			6 6	SCREW, 8-32 x ¾ inch, BHS WASHER, 85 x ¾ inch
92	175-0098-00			1	CABLE, crt input, assembly
93	132-0001-00			2	NUT, coupling
	132-0002-00			2	SLEEVE, conductor, outer NUT, retaining
	132-0004-00			2	FERRULE
	132-0006-00			2	WASHER
	132-0007-00 132-0009-00			2	RING, snap BODY brass
	132-0010-00			2	CONDUCTOR, outside, tapered
	132-0011-00			4	INSULATOR CONTAC: 125 Ω; assembly
	132-0013-00			2	CONDUCTOR, inside, tapered
94 95	276-0519-00			3	CORE, powered iron
96	441-0344-00			1	CHASSIS, trigger-shockmount
97	211-0018-00			3	SCREW, 4-40 x 7/8 inch, RHS
98	210-0958-00 348-0007-00			6	WASHER, .115 ID x ¹⁵ / ₃₂ inch OD SHOCKMOUNT_rubber
99	166-0106-00			3	TUBE, spacing
	210-0004-00 210-0406-00			3	LOCKWASHER, internal, #4 NUT, hex, 4-40 x 3/14 inch
				-	110 110 110
100	136-0072-00	101	579	4	SOCKET, 9 pin, UHF
	136-0015-00	580		4	SOCKET, STM9G mounting hardware for each (not included w/socket)
	213-0044-00			2	SCREW, thread cutting, 5-32 x 3/16 inch, PHS phillips
101	136-0095-00 136-0181-00	101 640	639	32	SOCKET, 4 pin, transistor SOCKET, 3 pin, transistor
	136-0182-00	040		-	mounting hardware for each: (not included w/socket)
	213-0113-00 354-0234-00	101 640	639	2 1	SCREW, thread forming, 2-32 x ³ / ₁₆ inch, RHS phillips RING, locking, transistor socket

CHASSIS (Cont'd)

REF.	PART NO.	SERIAL/	NODEL NO.	Q	DESCRIPTION
NO.		EFF.	DISC.	Ŷ.	
102	386-0253-00 211-0534-00 210-0202-00 210-0407-00	-		4 - 2 2 2	PLATE, metal, small capacitor mounting hardware for each: (not included w/plate) SCREW, 6-32 x ½, inch, PHS w/lockwasher LUG, solder, SE #6 NUT, hex, 6-32 x ¼ inch
103	210-0204-00 210-0202-00 211-0504-00 210-0407-00			1 1 1 1 1	LUG, solder, DE #6 LUG, solder, SE #6 mounting hardware: CCREW, 632 x /4, inch NUT, hex, 6-32 x 1/4, inch
104	441-0346-00 211-0507-00			1 - 4	CHASSIS, trigger mounting hardware: (not included w/chassis) SCREW, 6-32 x \$/16 inch, BHS
105	616-0413-00			1	ASSEMBLY, circuit board, trigger amplifier assembly includes: PLATE, printed circuit board
106	136-0095-00 213-0113-00			2	SOCKET, 4 pin transistor mounting hardware for each: (not included w/socket alone) SCREW, thread forming, 2-32 x $\frac{5}{16}$ inch, RHS phillips
107 108 109	131-0159-00 211-0061-00 385-0150-00 210-0201-00 211-0011-00			1 2 2 2 2 2	CONNECTOR, printed board, 14 contact, female mounting hardware: (not included w/connector) SCREW, 4-40 x ½ inch, Fil HS ROD, spacer LUG, solder, SE #4 SCREW, 4-40 x ³ / ₁₄ inch, BHS
110 111	131-0157-00 210-0204-00 213-0044-00			6 1 - 1	CONNECTOR, terminal stand-off LUG, solder, DE #6 mounting hardware: (not included w/lug) SCREW, thread cutting, 5-32 x ½, inch, PHS phillips
112	387-0353-00 387-0471-00 211-0504-00 210-0006-00 210-0407-00	101 269	268	1 1 3 3 3	PLATE, circuit board PLATE, circuit board, trigger count down mounting bardware: (not included w/plate) SCREW, 632 x ½, inch BHS LOCKWASHER, internal, #6 NUT, hex, 6-32 x ½, inch
113	387-0334-00 211-0507-00			1 - 4	PLATE, air baffle maunting hardware: (not included w/plate) SCREW, 6-32 × 5/16 inch, BHS
114	406-0629-00 211-0507-00 210-0803-00			1 10 10	BRACKET, air duct mounting hardware: (not included w/bracket) SCREW, 6-32 x \mathscr{Y}_{tk} inch, BHS WASHER, 6L x \mathscr{Y}_k inch

REF.	PART NO.	SERIAL/A	NODEL NO.	Q	DESCRIPTION
NO.		EFF.	DISC.	Υ.	
115	387-0335-00 210-0202-00 213-0049-00			1 2 2	PLATE, ground strap, mounting mounting hardware: (not included w/plate) LUG, solder, SE #6 SCREW, 6-32 x \$/ ₁₆ inch, HHS
116	385-0060-00 211-0507-00			2 1	ROD, nylon mounting hardware for each: (not included w/rod) SCREW, 6-32 x ${}^{5}\!/_{16}$ inch, BHS
117	406-0621-00 211-0011-00			1 - 2	BRACKET, tube mounting mounting hardware: (not included w/bracket) SCREW, 4-40 x $\%_{14}$ inch, BHS
118	136-0097-00 211-0511-00 210-0006-00 210-0407-00			1 3 3 3	SOCKET, 7 pin mounting hardware: (not included w/socket) SCREW, 6-32 x ½ inch, BHS LOCKWASHER, internal, #6 NUT, hex, 6-32 x ¼ inch
119	214-0012-00 210-0202-00 210-0006-00 210-0407-00	X743		2 1 2 2	BOLT, spade, 6-32 x 3/ ₈ inch mounting hardware for each: (not included w/bolt) LUG, solder, SE6 LOCKWASHER, internal, #6 NUT, hex, 6-32 x 1/ ₄ inch
120	343-0015-00 343-0005-00 211-0504-00 211-0507-00 210-0803-00 210-0006-00 210-0407-00	101 743 101 743 X743	742 742	2 2 1 1 1 1 1	CLAMP, right angle to circle CLAMP, plastic, 7/ ₁₆ inch mounting hardware for each: (not included w/clamp) SCREW, 6-32 x 1/ ₂ inch, BHS WASHER, 6L x 1/ ₈ inch LOCKWASHER, internal, #6 NUT, hex, 6-32 x 1/ ₄ inch
121	441-0340-00 441-0340-01 212-0004-00 212-0040-00 210-0458-00	100 743	742	1 1 4 8 4	CHASSIS, sweep CHASSIS, sweep mounting hardware: (not included w/chassis) SCREW. 8-32 x ½, inch, BHS SCREW. 8-32 x ½, inch, FHS phillips NUT, keps, 8-32 x ½, inch
122	385-0075-00 385-0135-00 211-0507-00 213-0041-00	101 225 101 225	224 224	4 4 - 1 1	ROD, nylon ROD, delrin mounting hardware for each: (not included w/rod) SCREW, 6-32x 3/ ₁₆ inch BHS SCREW, thread cutting, 6-32 x 3/ ₈ inch, THS phillips
123	211-0507-00 210-0203-00 210-0006-00 210-0407-00			1 2 1 2 2	TRANSISTOR mounting hardware: (not included w/transistor) LUG, solder, SE #6, long SCREW, 632 x ½, inch, BHS LOCKWASHER, internal, #6 NUT, hex, 6-32 x ¼, inch
124	384-0519-00 211-0507-00			2	ROD, spacing, hex mounting hardware for each: (not included w/rod) SCREW, 6-32 x \$/16 inch, BHS

8-25

REF.	PART NO	SERIAL/	MODEL NO.	Q	DESCRIPTION
NO.	PART NO.	EFF.	DISC.	Y .	DESCRIPTION
125	210-0202-00 211-0504-00			2	LUG, solder, SE #6 mounting hardware for each: (not included w/lug) SCREW, 6-32 x 1/4 inch, BHS
126	406-0635-00 213-0034-00	X105		1 - 2	BRACKET, low capacity pot mounting hardware: (not included w/bracket) SCREW, thread cutting, 4-40 x ½,4 inch, RHS phillips
127	210-0438-00	X105		1 2	POT, miniature mounting hardware: (not included w/pot) NUT, hex, 1-72 x ⁶ / ₃₂ inch
128 129 130 131	348-0002-00 348-0055-00 131-0158-00 136-0074-00 354-0109-00 213-0034-00	101 750	749	1 2 5 1 2	GROMMET, 1/4 inch GROMMET, plastic, 1/4 inch CONNETCOR, terminal feed-thru SOCKET, 9 pin, UHF, w/o saddle mounting hardware for each: (not included w/socket) RING, socker tertaining SCREW, thread cutting, 4-40 x 3/16 inch, RHS phillips
132	136-0071-00 213-0044-00			2	SOCKET, 7 pin, UHF mounting hardware for each: (not included wsocket) SCREW, thread cutting, 5-32 x 3/16 inch, PHS phillips
133	343-0004-00 211-0510-00 210-0803-00			2 1 1	CLAMP, cable, $\%_6$ inch mounting hardware for each: (not included w/clamp) SCREW, $6.32 x_6$ inch, BHS WASHER, $6Lx\%_6$ inch
134	352-0031-00 211-0507-00 210-0006-00 210-0407-00	X323		1 - 1 1 1	HOLDER, fuse, single mounting hardware: (not included w/holder) SCREW, 6-32 x ½, inch, BHS LOCKWASHER, internal, #6 NUT, hex, 6-32 x ¼ inch
135	343-0042-00 211-0510-00 210-0803-00			1 - 1 1	CLAMP, cable, $\frac{y}{t_6}$ inch (half) mounting hardware: (not included w/clamp) SCREW, 6.32 x ½; inch, BHS WASHER, 6L x ¼ inch
136	406-0620-00			1 - 3	BRACKET, pot mounting mounting hardware: (not included w/bracket) SCREW, $6.32 \times \frac{9}{16}$ inch, FHS phillips
137	175-0585-00 175-0587-00 175-0588-00 175-0593-00 175-0598-00 175-0598-00 134-0031-00			1 1 1 1 2 1	WIRE, crt lead, .290 foot, striped brown, w/connector WIRE, crt lead, .500 foot, striped red, w/connector WIRE, crt lead, .833 foot, striped blue, w/connector WIRE, crt lead, .333 foot, striped gray, w/connector WIRE, crt lead, .334 foot, striped gray, w/connector WIRE, crt lead, .334 foot, striped gray, w/connector PLUG, crt contact

CHASSIS (Cont'd)

REF.	PART NO	SERIAL/	MODEL NO.	Q T	DESCRIPTION
NO.	TAKI NO.	EFF.	DISC.	Ý.	BESCHI HON
139	136-0072-00 211-0007-00 211-0008-00 385-0109-00			1 2 2 2	SOCKET, 9 pin, UHF mounting handware: (not included w/socket) SCREW 4-40 x 1/ ₄₆ inch, BHS SCREW, 4-40 x 1/ ₄ inch BHS KOD, nylon
140	211-0013-00 210-0004-00 210-0406-00			1 2 2 2	CAPACITOR mounting hardware: (not included w/capacitor) SCREW, 4-40 \times γ_4 inch, RHS LOCKWASHER, internal, #4 NUT, hex, 4-40 \times γ_{4e} inch
141	441-0345-00 211-0016-00 211-0017-00 210-0201-00 210-0204-00 166-0005-00			1 - 5 4 7 1 9	CHASSIS, sweep output mounting hardware: (not included w/chassis) SCREW, 440 x %; inch, RHS SCREW, 440 x %; inch, RHS LUG, solder, DE #6 SPACER, vobe
143	131-0154-00 211-0511-00 210-0803-00 210-0006-00 210-0407-00	101	466	1 2 2 2 2 2	CONNECTOR, high voltage, chassis mounted mounting hardware: (not included w/connector) SCREW, 632 v/5 inch, BHS WASHER, 6L x ½, inch LOCKWASHER, internal, #6 NUT, hex, 6-32 x ¼, inch
144 145	131-0200-00 131-0164-00 131-0334-00 131-0334-01 166-0176-00 214-0210-00 361-0007-00	467 101 522 852 X229	521X 521 851	1 1 1 1 1 1	CONNECTOR, high voltage, female CONNECTOR, anade CONNECTOR, anade TUBE, air transfer SPOOL, solder, assembly mounting hardware: (not included w/spool) SPACER, nylon, .063 inch
146	136-0072-00 213-0044-00			2 - 2	SOCKET, 9 pin UHF mounting hardware for each: (not included w/socket) SCREW, thread cutting, 5-32 x $\frac{3}{16}$ inch PHS phillips
147	348-0003-00	101	749	4	GROMMET, rubber, 5/16 inch
148	348-0067-00 348-0006-00	750 101	749	4	GROMMET, plastic, ³ /16 inch GROMMET, rubber, ³ /4 inch
149	348-0050-00 348-0004-00	750 101	749	2	GROMMET, plastic, ³ / ₄ inch GROMMET, rubber, ³ / ₈ inch
150	348-0056-00 348-0005-00	750 101	749	2	GROMMET, plastic, $\frac{9}{8}$ inch GROMMET, rubber, $\frac{1}{2}$ inch
151	348-0063-00	750	749	1	GROMMET, plastic, 1/2 inch
	348-0064-00	750		i	GROMMET, plastic, 5% inch
152	136-0015-00 136-0181-00 213-0044-00 354-0234-00	101 X743 101 X743	742X 742X	1 2 2 1	SOCKET, STM9G SOCKET, 3 pin transistor mounting hardware for each: (not included w/socket) SCREW, thread forming, 5-32x 3/ ₁₆ inch PHS phillips RING, locking, transistor socket
153 154	252-0564-00 348-0004-00	X830 X297		FT 1	POLYETHYLENE, extruded channel, (2) $51_2''$ pieces GROMMET, rubber, 3_8 inch



CABLE HARNESS & CERAMIC STRIP DETAIL

REF. NO.	PART NO.	SERIAL/I EFF.	MODEL NO. DISC.	Q T Y.	DESCRIPTION
1 2 3 4 5 6 7 8 9	179-0431-00 179-0724-00 179-0724-01 179-0439-00 179-0430-00 179-0430-00 179-0430-00 179-0430-00 179-0430-00 179-0430-00 179-0440-00 124-0089-00 124-0089-00 365-0046-00 361-0009-00	101 403 743 101 403 477	402 742 402 476	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	CABLE HARNESS, sweep CABLE HARNESS, sweep CABLE HARNESS, sweep CABLE HARNESS, fornt panel CABLE HARNESS, power #1 CABLE HARNESS, power #1 CABLE HARNESS, power #1 CABLE HARNESS, power #1 CABLE HARNESS, progree chossis CABLE HARNESS, trigger chossis STUD, mylen mounting hardware for each: (not included w/strip) SPACER, mylen
10	124-0091-00 355-0046-00 361-0009-00			1 - 2 - 2	STRIP, ceramic, 3/2 inch x 11 notches strip includes: STUD, nylon mounting hardware for each: (not included w/strip) SPACER, nylon, 313 inch
11	124-0090-00 355-0046-00 361-0009-00			5 2 2	STRIP, ceramic, ½, inch x 9 notches each strip includes: STUD, nylon mounting hardware for each: (not included w/strip) SPACER, nylon, 313 inch
12	124-0088-00 355-0046-00 361-0009-00			6 - 2 - 2	STRIP, ceramic, ½, inch x 4 notches each strip includes: STUD, nyton mounting hardware for each: (not included w/strip) SPACER, nyton, 313 inch
13	124-0100-00 355-0046-00 361-0009-00			1	STRIP, ceramic, ¾ inch x 1 notch strip includes: STUD, nyton mounting hardware for each: (not included w/strip) SPACER, nyton, .313 inch
14	124-0094-00 355-0046-00 361-0009-00			1 2 2	STRIP, ceramic, 7/16 inch x 7 notches strip includes: STUD, nyton mounting hardware for each: (not included w/strip) SPACER, nyton, 313 inch
15	124-0092-00 355-0046-00 361-0009-00			1 1 1	STRIP, ceramic, 7/4 inch x 3 notches strip includes. STUD, nylon mounting hardware for each: (not included w/strip) SPACER, nylon, .313 inch

Parts List—Type 519

REF.	PART NO	SERIAL/	MODEL NO.	Q	DESCRIPTION
NO.	TAKI NO.	EFF.	DISC.	Y.	DESCRIPTION
16	124-0093-00			6	STRIP, ceramic, 7/16 inch x 5 notches each strip includes:
	355-0046-00			2	STUD, nyton mounting bardware for each: (not included w/strip)
	361-0009-00			2	SPACER, nylon, .313 inch
17	124-0092-00			1	STRIP, ceramic, 7/16 inch x 3 notches
	355-0046-00			1	STUD, nylon
	361-0008-00			i	mounting hardware for each: (not included w/strip) SPACER, nylon, .188 inch
18	124-0093-00			1	STRIP, ceramic, γ_{16} inch x 5 notches
	355-0046-00			2	STUD, nylon
					mounting hardware for each: (not included w/strip)
	361-0008-00				SPACER, nylon, 188 inch
19	124-0090-00			2	STRIP, ceramic, 3/4 tinch x 9 notches
	355-0046-00			2	each strip includes: STUD, nylon
				Ĩ.	mounting hardware for each: (not included w/strip)
	361-0008-00			2	SPACER, nylon, .188 inch
20	124-0089-00			22	STRIP, ceramic, 3/4 inch x 7 notches
	355 0046 00			-	each strip includes:
				1	mounting hardware for each: (not included w/strip)
	361-0008-00			2	SPACER, nylon, .188 inch
21	124-0091-00			10	STRIP, ceramic, 3/4 inch x 11 notches
	255 004/ 00			-	each strip includes:
	333-0048-00			2	mounting hardware for each: (not included w/strip)
	361-0008-00			2	SPACER, nylon, .188 inch
22	124-0095-00			4	STRIP, ceramic, $7/_{16}$ inch x 9 notches
	255 0044 00			-	each strip includes:
				-	mounting hardware for each: (not included w/strip)
	361-0009-00			2	SPACER, nylon, .313 inch
23	124-0086-00			2	STRIP, ceramic, 3/4 inch x 2 notches
	355-0046-00			1	each strip includes: STUD, nylon
				-	mounting hardware for each: (not included w/strip)
	361-0009-00			1	SPACER, nylon, .313 inch

CABLE HARNESS & CERAMIC STRIP DETAIL (Cont'd)

CABLE HAR	RNESS &	CERAMIC	STRIP	DETAIL	(Cont'd)
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REF.	PART NO	SERIAL/N	NODEL NO.	Q	DESCRIPTION
NO.	FART NO.	EFF.	DISC.	Ý.	DESCRIPTION
24	124-0095-00 355-0046-00 361-0008-00			3	STRIP, ceramic, 7/ ₁₆ inch x 9 notches each strip includes: STUD, nyton mounting hardware for each: (not included w/strip) SPACER, nyton, 188 inch
25	124-0106-00 355-0046-00 361-0008-00			4 - 2 - 2	STRIP, ceramic, 7/16 inch x 11 notches each strip includes: STUD, nylon mounting hardware for each: (not included w/strip) SPACER, nylon, 188 inch
26	124-0094-00 355-0046-00 361-0008-00			4 2 2	STRIP, ceramic, 7/16 inch x 7 notches each strip includes: aTLD, mytan maming hackware for each: (not included w/strip) SPACER, mytan, 188 inch

STANDARD ACCESSORIES



1

2

TERMINATION, 125 Ω

MANUAL, instruction (not shown)

017-0051-00

070-0243-00

550







TIME-BASE TRIGGER & HOLD-OFF

CIRCUIT NUMBERS

+

TIME-BASE GATE & UNBLANKING S/N 743 & UP

1066

CIRCUIT NUMBERS 200 THRU 287 ALSO V393B

+50

JUSEC/CM

50V/CM





D3

GATE EXTENDER CIRCUIT-

+225V +225V

ŽR270

SEE TIMING



+





с,

TYPE 519 OSCILLOSCOPE

TIMING SWITCH

аль 1263





TYPE 519 OSCILLOSCOPE

E HEATER WIRING DIAGRAM

+-



CIRCUIT NUMBERS



С

TYPE 519 OSCILLOSCOPE

+

CALIBRATION - STEP GENERATOR

GAB 567

CIRCUIT NUMBERS 870 THRU 898

SEE PARTS LIST FOR EARLIER VALUES AND SERIAL NUMBER RANGES OF PARTS MARKED



A2

CIRCUIT NUMBERS 910 THRU 939 ALSO V895B

S/N 101 - 481 RATE GENERATOR бав 1263





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MANUAL CHANGE INFORMATION

At Tektronix, we continually strive to keep up with latest electronic developments by adding circuit and component improvements to our instruments as soon as they are developed and tested.

Sometimes, due to printing and shipping requirements, we can't get these changes immediately into printed manuals. Hence, your manual may contain new change information on following pages. If it does not, your manual is correct as printed. TYPE 519

ELECTRICAL PARTS LIST CORRECTION

REMOVE :

	D1 80	152-0185-00	Silico	n .	
CHAN	GE TO:				
	C183	283-0001-00	.005 µf	Discap	500 V
	R336 B	310-0590-01	18 kΩ	8 w	1%
	R336F	310-0590-01	18 kΩ	8 w	1%
	R336G	310-0590-01	18 kΩ	8 W	1%
ADD:					
	R1 86	315-0101-00	100 N	1/4 W	5%

SCHEMATIC CORRECTION



TYPE 519 TENT SN 1280

ELECTRICAL PARTS LIST CORRECTION

CHANGE TO:

D344	153-0026-01	Zener 1	W	120 V	Selected	neir
D345	(1))-0020-02	Zener 1	W	120 V	,	Parr

ML3,776/868

PARTS LIST CORRECTION

CHANGE TO:

R156	301-0392-00	3.9 k 1/2 W 5%
ADD:		
D73	152-0185-00	Silicon
D180	152-0185-00	Silicon
R79	315-0220-00	22 Q 1/4 W 5%







PARTIAL TIME-BASE TRIGGER & HOLD-OFF

M13,537/268

- 1

ELECTRICAL PARTS LIST CORRECTION

CHANGE TO:

	R760	323-0296-00	11.8 kQ	1/2 W	Prec	1
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ELECTRICAL PARTS LIST CORRECTION

CHANGE TO:

r853	311-0121-01
R856	311-0041-02

TYPE 519 TENT SN 910

TEXT CORRECTION

SECTION 6 -- CALIBRATION PROCEDURE

Page 6-2, ADJUSTMENT PROCEDURE

Change step 4 to read as follows:

4. Set Position of Pulse Amplitude or Sync Knob

Set the FULSE AMPLITUDE OR SYNC control counterclockwise and the NANOSEC/CM switch to 2. Connect a DC voltmeter to the center terminal of R/G (located on the front ceramic strip on the Time-Base Trigger chassis) and with a screwdriver set the potentiometer for a reading of maximu megative volts. Turn the FULSE AMPLITUDE OR SYNC control slowly clockwise until the sweep free runs, then back the control off slightly. Adjust R/G until the sweep fust stops running. The dot on the FULSE AMPLITUDE OR SYNC knob should be just above the RECURRENT line on the front panel. If not, loosen the knob. The sweep should now free run any time that the knob is positioned clockwise past the line.

PARTS LIST CORRECTION

CHANGE TO:

C77	283-0114-00	.0015 µF	200V	
R78	302-0471-00	470 Ω	1/2 W	10%
Q70,Q180	151-0083-00	Similar to	2N964	

ADD:

R76 311-0480-00

500 Ω ±20%





M11,147/866


519 NO. /33/

09743

CRT. NO.



519 NO./33/ TRIGGERING ABILITY Long Duration & Short Impulse 2mm & 2cm Scoler /.'/