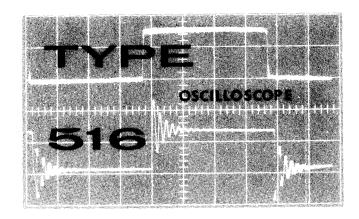
# INSTRUCTION

Serial Number \_\_\_\_\_



# WARRANTY

All Tektronix instruments are warranted against defective materials and workmanship for one year. Tektronix transformers, manufactured in our own plant, are warranted for the life of the instrument.

Any questions with respect to the warranty mentioned above should be taken up with your Tektronix Field Engineer.

Tektronix repair and replacement-part service is geared directly to the field, therefore all requests for repairs and replacement parts should be directed to the Tektronix Field Office or Representative in your area. This procedure will assure you the fastest possible service. Please include the instrument Type and Serial number with all requests for parts or service.

Specifications and price change privileges reserved.

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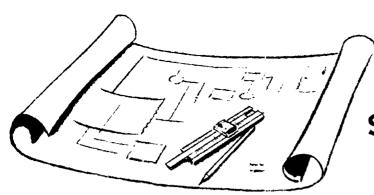
Section 8 Electrical Parts List

Diagrams

Abbreviations and symbols used in this manual are based on, or taken directly from IEEE Standard 260 "Standard Symbols for Units", MIL-STD-12B and other standards of the electronics industry. Change information, if any, is located at the rear of this manual.



# SECTION 1



# SPECIFICATIONS

#### Introduction

The Tektronix Type 516 Oscilloscope is a dual-trace general purpose laboratory oscilloscope with a bandpass of dc to 15 megahertz. Dual-trace operation is provided by two identical vertical input channels; signals may be applied to either or both input channels at the same time. Switching between the two channels takes place at the end of each sweep of the beam or at a free-running rate of approximately 150 kHz. Either channel may be used separately to provide single-trace operation when desired.

# **Vertical Deflection System, Both Channels**

Bandpass . . . DC to 15 MHz (response not more than 3 dB down within these limits).

Risetime...23 nanoseconds or less.

Sensitivity...0.05 volt/div. to 20 volts/div. in 9 calibrated steps; accuracy with 3%. Continuously variable from 0.05 volt/div. to at least 50 volts/div., uncalibrated.

Input Impedance...1 megohm paralleled by 20 picofarads.

Operating Modes . . . Channel A only, Channel B only, electronic switching at approximately 150 kHz (chopped), or electronic switching on alternate sweeps.

Maximum Voltage Input...600 volts, combined dc- and acvoltage.

#### Triggering

Type...Amplitude-level selection with preset or manual stability control.

Modes...Automatic, ac-coupled, dc-coupled, and high-frequency synchronized.

Source...Internal from vertical signal, external from triggering signal, or line frequency.

Slope...Plus (rising slope of triggering waveform), or minus (falling slope of triggering waveform).

Signal Requirements . . . Internal-AC: 2 mm of display at 1 kHz increasing to 5 mm at 2 MHz. Low frequency response is 3 dB down at approximately 16 Hz. AUTO: 5 mm of display from 50 Hz to 1 kHz increasing to 1 cm at 2 MHz. DC: 5 mm of display from DC to 1 kHz increasing to 2 cm at 2 MHz. AC LF REJECT: Attenuates frequencies below 16 kHz. HF SYNC: 2 cm of display at 20 MHz. External-AC: 0.5 V at 1 kHz increasing to 1.5 V at 2 MHz. Low frequency response is 3 dB down at approximately 16 Hz. AUTO: 1 V from 50 Hz to 1 kHz increasing to 3 V at 2 MHz. DC: 0.5 V from DC to 1 kHz

increasing to 1.5 V at 2 MHz. AC LF REJECT: Attenuates frequencies below 16 kHz. HF SYNC: 2 V at 20 MHz. Line-AC LF REJECT: Attenuates frequencies between 16 kHz.

## Sweep

Type...Miller Integrator.

Sweep Rates...0.2  $\mu$ sec/div. to 2 sec/div. in 22 calibrated steps. Accuracy typically within 1% of full scale; in all cases, within 3% of full scale.

Continuously variable sweep rates are available from 0.2 asec/div. to 6 sec/div., uncalibrated.

Magnifier...Expands center portion of sweep 5 times. Extends fastest sweep rate to 0.04  $\mu$ sec/div; accuracy within 5%.

## **External Horizontal Input**

Bandpass . . . DC to 500 kHz (response not more than 3 dB down within these limits).

Deflection Factor...1.5 volts/div.

# **Amplitude Calibrator**

Waveform...Square-waves at approximately 1,000 cycles.

Amplitude...0.05 volt to 100 volts, peak-to-peak, in eleven fixed steps; accuracy within 3% of indicated amplitude.

#### Cathode-Ray Tube

Type...T55P31

P1, P2, P7 and P11 phosphors optional; other phosphors furnished on special order.

Accelerating Potential...4,000 volts.

Unblanking...Cutoff type, dc-coupled.

Z-Axis Modulation... External terminal permits RC coupling to crt cathode.

Deflection Factors at plates

Vertical-6 to 71/2 volts per centimeter.

Horizontal-19 to 23 volts per centimeter.

## **Output Waveforms Available**

Positive gate, coincident with sweep, at least 20 volts peak-to-peak.

## Specifications — Type 516

Positive-going sawtooth, coincident with sweep, at least 150 volts peak-to-peak.

#### **Graticule**

Illumination... Variable edge lighting.

Display Area...Marked in 6 vertical and 10 horizontal divisions. Each major division divided into 5 parts on centerlines.

# **Power Supplies**

Electronically regulated for stable operation with widely varying line voltages and loads (see Section 2 of this manual).

Power...Approximately 310 watts.

## **Ventilation**

Forced filtered air. Thermal relay interrupts instrument power if the internal temperature rises above 137°F.

## **Mechanical Specifications**

Construction...Aluminum alloy chassis and cabinet.

Finish...Photo-etched anodized front panel, blue vinyl-finish cabinet.

Dimensions... $13\frac{1}{2}$ " high,  $9\frac{3}{4}$ " wide,  $21\frac{1}{2}$ " deep.

# **Accessories**

2... P6006 Probe, 010-0127-00

2 . . . Instruction Manuals, 070-0225-01

1 . . . 3- to 2-wire Adapter, 103-013-00

1 . . . BNC to BNC Patch Cord, 012-0087-00

1 . . . BNC to Banana Patch Cord, 012-0091-00

1 . . . BNC to Post Jack, 012-0092-00

1 . . . 3-conductor Power Cord, 161-0010-03

1 . . . Green Light Filter, 378-0567-00

# SECTION 2



## **Power Requirements**

The Type 516 Oscilloscope line transformer can be wired for proper operation of the instrument on nominal line voltages of 110, 117, 124, 220, 234, and 248 volts at line frequencies of 50 to 60 Hz. Fig. 2-1 shows the transformer primary connections to be made for each line voltage.

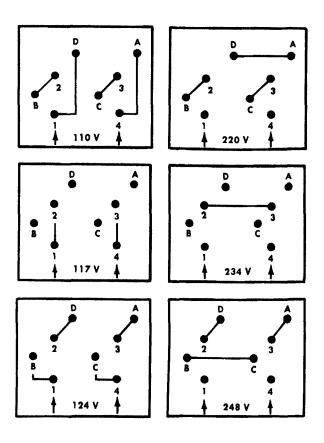


Fig. 2-1. Power Transformer primary connections.

Proper regulation of the oscilloscope power supplies will be maintained at line voltages between 105 and 125 volts when the instrument is wired for a nominal voltage of 117 volts, and within proportionate limits when it is wired for the other nominal line voltages.

When the Type 516 is supplied with a dc fan and the associated rectifiers (circuit D646 on the Power Supply circuit diagram), it can be operated at any line frequency

# PRELIMINARY INFORMATION

from 50 Hz to 400 Hz, although slightly higher line voltages are required at the higher line frequencies. Normally, the Type 516 Oscilloscope with the dc fan will operate satisfactorily on a 400-Hz line voltage of 117 volts when the primary of the line transformer is connected for 110-volt, 50-to-60 Hz operation, as shown in Fig. 2-1. For maximum dependability and longest life, is is recommended that the line voltage be kept at or slightly below the nominal,

## **Fuse Requirements**

When the Type 516 Oscilloscope is connected for 110-, 117-, or 124-volt operation, a 3.2-amp slow-blowing type fuse should be used. When the instrument is connected for 220-, 234-, or 248-volt operation, a 1.6-amp slow-blowing type fuse should be used.

#### Cooling

Your Type 516 Oscilloscope will last much longer if you keep it as cool as possible whenever it is being operated. A fan provides cooling by drawing air in through a filter at the rear of the instrument and blowing it over the internal components. The instrument must be placed such that the air intake is not blocked, and the air filter must be kept clean to permit adequate air circulation. Instructions for replacing and cleaning the air filter are given in Section 5 of this manual.

Furthermore, the side panels of the Type 516 Oscilloscope are designed to promote maximum air circulation over the internal components where the most heating takes place. For this reason, the instrument should not be operated for long periods of time with the side panels off. Also, there are differences between the right-hand side panel and the left-hand side panel, so they must be mounted on the proper sides. Fig. 2-2 shows the two sides of the oscilloscope with the side panels correctly mounted.

If the temperature inside the instrument should become so high that it might cause immediate damage to components, a thermal cutout switch will disconnect the power. When the temperature drops to a safe level, the switch will again close automatically; no manual reset is necessary.

#### Fan Connections

The manner in which the fan is wired in the Type 516 Oscilloscope depends upon the line voltage for which the instrument is wired. For 110-, 117-, and 124-volt operation, the black fan lead should be connected to the third notch of the 11-notch ceramic terminal strip at the right rear of the

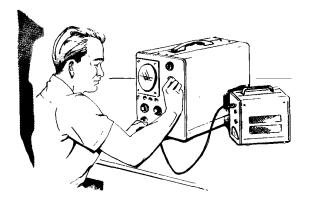


Fig. 2-2. Type 516 Oscillescope, showing side panels properly mounted.

bottom of the instrument. For 220-, 234-, and 248-volt operation, it should be connected to the second notch of the same ceramic strip.

Note that when a dc fan is used, no change in fan connections is necessary with changes in nominal operating voltage.

# SECTION 3



# **OPERATING** INSTRUCTIONS

## Introduction

This section of the manual is divided into three parts. The first part, Preliminary Instructions, is a step-by-step procedure designed to acquaint a "first-time" operator of a Tektronix oscilloscope with the basic operation of the Type

516. The second part, Operating Information, which starts on page 3-5, covers all phases of the operation of the Type 516 Oscilloscope in detail. The third part, Applications, which starts on page 3-8, describes techniques which can be used for certain basic applications of the Type 516 Oscilloscope.

# **Preliminary Instructions**

The front panel of the Type 516 Oscilloscope is shown in Fig. 3-1. Functions of all front panels controls, indicators, and connectors are described in Table 3-1.

#### TABLE 3-1

TRIGGER **SELECTOR** (black knob) Selects source of triggering signal and slope on which triggering occurs.

TRIGGER **SELECTOR** (red knob)

Selects triggering mode.

TRIGGER INPUT Coaxial connector for application of external triggering signal when black TRIG-GER SELECTOR control is in the EXT. position.

PRESET STABILITY Sets triggering stability level when red

TRIGGER SELECTOR control is in the AUTO. position, or when STABILITY control is in the PRESET position.

TRIGGERING

LEVEL

Selects the voltage point on the triggering waveform at which the horizontal sweep is

triggered.

STABILITY OR HORIZ.-INPUT

Sets Time-Base Generator for triggered or free-running operation. Also serves as ATTEN. (red knob) attenuation control for signals applied through EXT, HORIZ, INPUT when HORIZ.

DISPLAY switch is in the EXT. position.

HORIZ. DISPLAY Sets horizontal sweep for normal or magni-

fied (X5) presentation. Also provides for application of external signal to Horizontal Amplifier.

TIME/DIV.

Selects the desired horizontal sweep rate

from 22 calibrated steps.

VARIABLE TIME/DIV. (red knob) Provides a continuous range of sweep rates between the fixed steps selected by the TIME/DIV. control. (The sweep rates are

calibrated only when the VARIABLE control is in the CALIBRATED position.)

UNCALIBRATED lamp

Lights when VARIABLE TIME/DIV. control is not in the CALIBRATED position.

#### A VERTICAL

VOLTS/DIV.

Selects the sensitivity of the A VERTICAL Channel from nine calibrated steps.

VARIABLE VOLTS/DIV. (red knob)

Provides a continuous range of sensitivity values between the fixed calibrated steps selected by the VOLTS/DIV. control. (Sensitivity is calibrated only when the VARIABLE control is in the CALIBRATED

position.)

UNCALIBRATED lamp

Lights when VARIABLE VOLTS/DIV. control is not in the CALIBRATED position.

SHUNT and SERIES adjustments

Compensation adjustments for attenuator.

(See Calibration Section.)

**POLARITY** 

Selects ac or dc coupling of signal applied to A VERTICAL input connector, and normal or inverted presentation on the screen.

**POSITION** (red knob)

Controls vertical positioning of A VERTI-CAL Channel signal on the crt screen.

DC. BAL.

Provides for vertical stability of no-signal trace for all positions of the A VERTICAL

VOLTS/DIV. control.

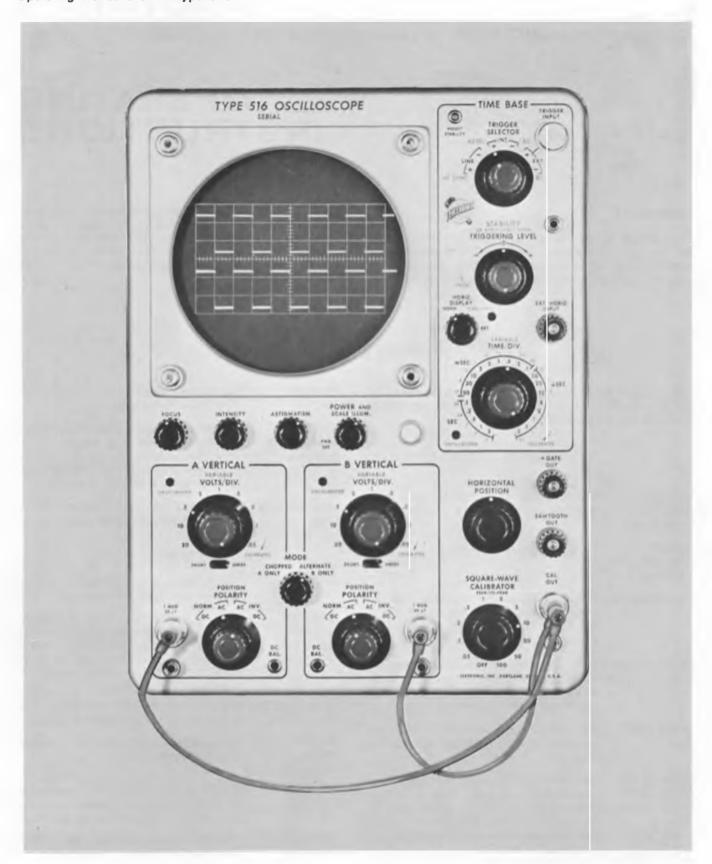


Fig. 3-1. Type 516 Oscilloscope front panel.

Input connector Input connector for signal through A VERTI-CAL Channel.

#### **B VERTICAL**

(All controls in the B VERTICAL Channel are the same as those in the A VERTICAL Channel except that they control the signal applied through the B VERTICAL Channel input connector.)

#### **OTHERS**

**FOCUS** Focuses the trace or spot on the screen. INTENSITY Controls the brightness of the trace or spot on the screen. **ASTIGMATISM** Controls the roundness of the spot on the **POWER AND** Turns the instrument on and off, and con-SCALE ILLUM. trols graticule illumination. MODE Provides for display of signal in A VERTI-CAL Channel or B VERTICAL Channel independently, or both channels either on alternate sweeps of the trace or alternately at 150-kHz rate. HORIZONTAL Controls horizontal positioning of both A **POSITION** and B VERTICAL Channel signals on the +GATE OUT Provides +25-volt (approximate) positive connector gate coincident with sweep time. SAWTOOTH Provides +150-volt (approximate) sawtooth waveform coincident with sweep time. OUT connector

\*On some instruments, this control is labeled SQUARE-WAYE CALIBRATOR.

Calibrator output.

Selects amplitude of square-wave available

at CAL. OUT connector from 11 calibrated

Coaxial connector for supplying Amplitude

## **INITIAL OPERATION**

The following paragraphs describe a simple procedure by which you can become acquainted, in a general way, with the controls of the Type 516 Oscilloscope and the effect they have on its operation.

#### Initial Setup

**AMPLITUDE** 

CAL. OUT

connector

CALIBRATOR\*

To initially set up the Type 516 Oscilloscope for operation, proceed as follows:

- 1. Connect jumper wires from the CAL, OUT connector to the A and B VERTICAL Channel input connectors and set the oscilloscope controls as shown in Fig. 3-1.
- 2. Adjust the FOCUS, INTENSITY, and ASTIGMATISM controls to obtain sharply defined traces of useful brightness.

3. Center one waveform in the upper half of the graticule with the A VERTICAL Channel POSITION control and the other waveform in the lower half of the graticule with the B VERTICAL Channel POSITION control.

#### **Functions of Controls**

With the oscilloscope set up as described, turn the MODE switch through each of its positions. Note that in the A ONLY and B ONLY positions only one waveform is displayed, and in the ALTERNATE and CHOPPED positions, both waveforms are displayed. (When the MODE switch is in the CHOPPED position, the display may not be stable.) The difference between the ALTERNATE and CHOPPED positions is described later in this section of the manual. Set the MODE switch to the A ONLY position.

Set the red TRIGGER SELECTOR control to AC. Move the TRIGGERING LEVEL control throughout its range, noting that the waveform appears when the control is set toward the middle of its range and disappears as it is set toward either end of its range. Set the AMPLITUDE CALIBRATOR control to 5. Move the TRIGGERING LEVEL control throughout its range again, noting that the waveform is present during a narrower portion of its range than before. Set the AMPLITUDE CALIBRATOR control to 2. Note that the waveform appears during still less of the range of the TRIGGERING LEVEL control.

Set the AMPLITUDE CALIBRATOR control to 10 and the A VERTICAL VOLTS/DIV. control to 10, and then to 20, adjusting the TRIGGERING LEVEL control as described in the previous paragraph at each step. Note that the adjustment of the TRIGGERING LEVEL control becomes more critical as the triggering signal is reduced, whether by actual reduction of the signal or reduction of vertical sensitivity. (Changing the vertical sensitivity affects the amplitude of the triggering signal only when an internal triggering signal is used.)

Set the AMPLITUDE CALIBRATOR control to 2 and the A VERTICAL VOLTS/DIV. control to 10. Set the STABILITY and TRIGGERING LEVEL controls fully clockwise. Turn the STABILITY control counterclockwise until the trace disappears. Then set the TRIGGERING LEVEL control for a stable display. This is the method to use if a stable display is not present when the red TRIGGER SELECTOR control is in the AUTO. position, or cannot be obtained by means of the TRIGGERING LEVEL control alone with the STABILITY control set in the PRESET position.

Set the AMPLITUDE CALIBRATOR control to 1 and the A VERTICAL VOLTS/DIV. control to 2, then to 5, then to 10, and then to 20. At each setting of the A VERTICAL VOLTS/DIV. control, adjust the STABILITY and TRIGGERING LEVEL controls, as described in the previous paragraph, for a stable display. Note that as the vertical deflection on the screen becomes less, stable triggering becomes more difficult to obtain, until at the higher settings of the VOLTS/DIV. control, it may be impossible to obtain. This is because the setting of the VOLTS/DIV. control also affects the amplitude of the internal triggering signal. Normally, vertical deflection of at least one-fifth of a major graticule division is required for proper triggering from an internal signal.

Remove the jumper wire from the B VERTICAL input connector and connect it from the CAL. OUT connector to the

#### Operating Instructions — Type 516

TRIGGER INPUT connector (leave the jumper wire connected from the CAL. OUT connector to the A VERTICAL input connector). Set the black TRIGGER SELECTOR control to +EXT. You have now set up the oscilloscope for external triggering, with the Calibrator supplying the external triggering signal as well as the displayed signal.

Set the A VERTICAL VOLTS/DIV. control to 1 and the AMPLITUDE CALIBRATOR control to 1, then to .5, then to .2, then to .1, and then to .05. At each setting of the AMPLITUDE CALIBRATOR control, set the STABILITY and TRIGGERING LEVEL controls for a stable display, if possible. Note that at the lowest settings of the AMPLITUDE CALIBRATOR control, stable triggering becomes very difficult or impossible to obtain. This is because there is not sufficient triggering signal being applied to the Time-Base Trigger. Normally, an external triggering signal of about 0.5 volt peak-to-peak amplitude is required for proper triggering.

Set the AMPLITUDE CALIBRATOR control to 1 again. Turn the A VERTICAL VOLTS/DIV. control counterclockwise. Note that no matter how small the deflection on the crt becomes, the display continues to be triggered. This is because the setting of the VOLTS/DIV. control has no effect on an external triggering signal.

Set the AMPLITUDE CALIBRATOR control to 20 and the A VERTICAL VOLTS/ DIV. control to 5. Set the black TRIGGER SELECTOR control to +INT., and set the STABILITY and TRIGGERING LEVEL control for a stable display. Rotate the TRIGGERING LEVEL control back and forth and note the vertical variation in the starting point of the square wave at the left-hand edge of the screen. (It may be necessary to move the display slightly to the right with the HORIZONTAL POSITION control to observe this.) Note also that the display starts on a positive-going portion of the waveform. Set the black TRIGGER SELECTOR control to —INT. and rotate the TRIGGERING LEVEL control as before. Note that the display now starts on a negative-going portion of the square-wave.

With the A VERTICAL POSITION control, move the display up and down on the screen. Note that the point at which the waveform starts at the left-hand edge of the screen remains fixed relative to the top and bottom of the waveform as the display moves. This is because when the red TRIGGER SELECTOR control is set to AC, triggering occurs at a given level relative to the average dc level of the entire signal. Set the red TRIGGER SELECTOR control to DC, and move the display up and down on the screen with the A VERTICAL POSI-TION control. (If the display disappears, turn the A VERTICAL POSITION control in the opposite direction to make it reappear.) Note that the point at which the waveform starts at the left-hand edge of the screen remains approximately fixed relative to the graticule. This is because when the red TRIGGER SELECTOR control is set to DC, triggering occurs at a specific dc level with respect to ground.

Set the AMPLITUDE CALIBRATOR control to 10 and the red TRIGGER SELECTOR control to AUTO. Switch the TIME/DIV. and VARIABLE TIME/DIV. controls through their ranges and note the effect on the display. Note that the associated UNCALIBRATED lamp lights when the VARIABLE TIME/DIV. control is moved away from the CALIBRATED position. Re-

turn the TIME/DIV. control to the  $.5\mu$ SEC position and the VARIABLE TIME/DIV. control to the CALIBRATED position. With the HORIZONTAL POSITION control, position the display horizontally so that one of the vertical portions of the waveform coincides with the center graticule line. Set the HORIZ. DISPLAY switch to the MAG. position. Note that the center portion of the waveform has been expanded. Specifically, the portion of the waveform which occupied the center two divisons horizontally when the HORIZ. DISPLAY switch was in the NORM. position now occupies the entire graticule. Return the HORIZ. DISPLAY switch to the NORM. position.

Switch the A VERTICAL POLARITY switch to the NORM. DC position. Note the change in the position of the waveform on the screen. This is because the dc component of the square wave is now included in the display whereas, in the AC position, the display showed only the variation around the average dc level of the square wave.

To see the effect of the INV. positions of the POLARITY switch, set the black TRIGGER SELECTOR control to +EXT. (The jumper wire must still be connected from the CAL. OUT connector to the TRIGGER INPUT and A VERTICAL input connectors.) Now move the A VERTICAL POLARITY switch back and forth between NORM. DC and INV. DC. Note that as the switch moves from the NORM. positions to the INV. positions, the square wave becomes inverted on the screen. Note further that the position of the bottom of the square wave on the screen when the switch is at NORM. DC corresponds to the position of the top of the square wave on the screen when the switch is at INV. DC. This is because the negative portion of the Calibrator square wave is at dc ground and the signal is being inverted about the dc ground potential point.

The effect of all of the preceding operations is the same when the MODE switch is in the B ONLY position except that the B VERTICAL Channel controls must be used instead of the A VERTICAL Channel controls. The effect of the operations with the MODE switch in the ALTERNATE or CHOPPED positions is approximately the same, except that external triggering should be used for best results.

# Graticule Illumination

Graticule illumination can be adjusted to suit the lighting conditions of the room by means of the POWER AND SCALE ILLUM, control. Turning the control clockwise increases the graticule illumination. It is possible to extinguish the graticule illumination completely by turning the control counterclockwise.

The graticule of the Type 516 Oscilloscope can be illuminated so that it appears to have either red or white graticule markings. The markings can be changed from white to red or red to white simply by removing the graticule cover and inverting the graticule. As a general rule, white graticule lines are superior to red for photographic purposes.

# **Operating Information**

# Input Coupling

The Type 516 Oscilloscope is provided with two input connectors to the Vertical Amplifier. Signals may be connected to both of these input connectors at the same time and, by means of the MODE switch, they may be displayed one at a time or both together on the crt screen. When only one signal is to be displayed, it may be connected to either input connector.

## **Polarity Switches**

Input signals to both channels may be either ac or dc coupled and may be displayed either normally or inverted on the crt screen by placing the corresponding POLARITY switches to the appropriate positions. Dc coupling applies both the ac and dc components of the input signal to the amplifier circuits. This permits you to measure the dc voltage level as well as the amplitude of the ac component. However, it is sometimes neither necessary nor desirable to display the dc component, and in such cases ac coupling should be used. Placing the POLARITY switch to either of the AC positions inserts a capacitor in series with the input connector. This capacitor blocks the dc component while allowing the ac component to be displayed.

#### **MODE Switch**

When the MODE switch is in the A ONLY position or the B ONLY position, only the signal which is applied to the corresponding vertical channel will be displayed on the crt. When the MODE switch is in the ALTERNATE position, the oscilloscope will display the signals in each of the channels on alternate sweeps of the trace. When the MODE switch is in the CHOPPED position, the oscilloscope will display the signals alternately at about a 150-kHz rate; in other words, the signal in the A Channel is displayed for about 31/3 microseconds and then the signal in the B Channel is displayed for about  $3\frac{1}{3}$  microseconds. At the faster sweep rates, the CHOPPED mode of operation causes the traces to take on a dotted appearance. At the slower sweep rates, the ALTERNATE mode of operation causes the alternate appearance of the traces to become quite noticeable which makes it difficult to compare the two. Therefore, in general, the ALTERNATE mode of operation is most useful at the faster sweep rates and the CHOPPED mode of operation is most useful at the slower sweep rates.

#### DC Balance Adjustment

Occasionally, there is need for adjustment of the dc balance of one or both input channels. This need is indicated by a vertical shift in the position of a no-signal trace as the VARIABLE VOLTS/DIV, control is rotated.

To make this adjustment, set the MODE switch to ALTER-NATE. With no signal connected to the input connectors, set the STABILITY control fully clockwise and position the two

free-running traces on the screen. Rotate each VARIABLE VOLTS/DIV. control back and forth, and simultaneously adjust the corresponding DC BAL. adjustment until the trace position is no longer affected by rotation of the VARIABLE VOLTS/DIV. control.

#### **Input Connections**

Certain precautions must be observed in connecting the oscilloscope to the signal source to prevent errors due to stray electric or magnetic coupling in the leads. Shielded cables should be used whenever possible, with the shield connected to the chassis of both the oscilloscope and the signal source. Regardless of the type of input lead used, it should be kept as short as possible.

In broadband applications, it might be necessary to terminate a coaxial input cable with a resistor or an attenuating pad presenting a resistance equal to the characteristic impedance of the cable. This is to prevent resonance effects and "ringing" (high-frequency damped oscillation). It becomes more necessary to terminate the cable properly as the length of the cable is increased. The termination is generally placed at the oscilloscope end of the cable, although many sources require an additional termination at the source end of the cable as well.

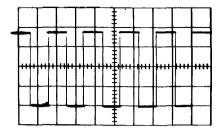
As nearly as possible, the actual operating conditions of the equipment being tested must be maintained. For example, the equipment should work into a load impedance equal to that which it will see in actual use. The input connectors of the Type 516 Oscilloscope present an input impendance of 1 megohm in parallel with 20 picofarads. With a few feet of shielded cable, the input capacitance may well be as much as 100 picofarads. In cases where the effects of these resistive and capacitive loads are significant in terms of the equipment being tested, you should use an attenuator probe as described in the next paragraph.

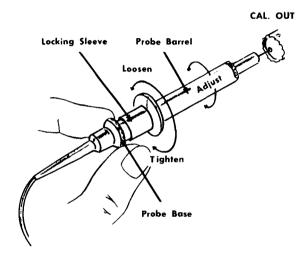
## **Use of Probes**

Use of the attenuator probes furnished with the Type 516 Oscilloscope reduces the capacitive and resistive loading effect on the equipment under test and, at the same time, reduces sensitivity. Connected to the input connectors of the Type 516 and properly compensated, these probes present a characteristic input impedance of 10 megohms in parallel with 8 picofarads and have an attenuation ratio of 10:1. The maximum-voltage rating of the probes is 600 volts. Exceeding this rating, either in peak ac volts or dc volts, may result in damage to the probes.

When making amplitude measurements with an attenuator probe, be sure to multiply the observed amplitude by the attenuation factor of the probe. If the waveform being displayed contains fast-changing portions, it is generally necessary to clip the probe ground lead to the chassis of the equipment being tested.

<u>3-5</u>





Display several cycles of Calibrator waveform on crt screen. Hold probe barrel and loosen locking sleeve. Hold probe base and adjust probe barrel for flat-topped square waves. Hold probe barrel and carefully tighten locking sleeve.

Fig. 3-2. Probe compensation adjustment.

An adjustable capacitor in the probe body compensates for variations in the input capacitance from one instrument to another or between two input connectors of one instrument. To insure the accuracy of pulse and transient measurements, this adjustment should be checked frequently. Fig. 3-2 illustrates the method of adjusting the compensation of the probe.

#### TRIGGERED OPERATION

For most uses of your oscilloscope, you will need to get a stable display of some recurrent waveform. In order to obtain a stable display, it is necessary to start the horizontal sweep consistently at the same time relative to recurring cycles of the input waveform. The sweep therefore must be triggered by the input waveform itself or by some waveform which bears a fixed time relationship to the input waveform. The following instructions tell you how to select and use the proper triggering signal for various applications of the oscilloscope.

# **Selecting the Trigger Source**

For most applications the sweep can be triggered by the input waveform. The only requirement is that the input signal be large enough to provide at least one-fifth of a major graticule division of deflection on the screen at the sensitivity level for which the VOLTS/DIV. control is set. To obtain triggering of the sweep from the input waveform, set the black TRIGGER SELECTOR control to either the +INT. or the —INT. position (the significance of the + and — will be explained in a later paragraph).

For dual-trace operation, it is usually advantageous to trigger the sweep with some external trigger. In this way, the true phase relationship of the two waveforms can be shown and triggering is obtained more easily. In order to obtain a stable display, it is necessary that the external triggering waveform bear a fixed time relationship to the input waveforms. The external triggering signal must be at least 0.2 volt in amplitude to assure proper triggering. To use an external waveform for triggering the horizontal sweep, connect the signal to the TRIGGER INPUT connector and set the black TRIGGER SELECTOR control to the +EXT. or -EXT. position. External triggering may also be used with single trace operation where desired.

When you are observing a waveform which bears a fixed time relationship to the ac line frequency, you may wish to trigger the sweep from the line-frequency waveform. To do this, place the black TRIGGER SELECTOR control in the +LINE or -LINE position.

When the black TRIGGER SELECTOR control is set to one of the + positions (+INT., +EXT., +LINE), the sweep is triggered on a positive slope of the triggering signal. When the control is set to one of the — positions (—INT., —EXT., —LINE), the sweep is triggered on a negative slope of the triggering signal. In most cases, selection of the triggering slope is not critical.

# Selecting the Triggering Mode (HF SYNC, AUTO., AC, DC)

The most useful triggering mode is the automatic mode, obtained by setting the red TRIGGER SELECTOR control to the AUTO, position. The automatic mode of triggering provides a stable display with virtually any triggering signal from about 50 Hz to 2 MHz in frequency. No adjustment of the STABILITY or TRIGGERING LEVEL controls is necessary since these two functions are automatically set to their optimum values internally. (In the other triggering modes, it is necessary to set one or both of these controls to obtain a stable presentation.) In addition, the automatic mode of triggering provides a free-running trace on the screen in the absence of any triggering signal. This trace may be used, under certain conditions, as a base line from which to make dc measurements. It also gives a constant indication of the readiness of the oscilloscope to accept a triggering signal for the display of a stable waveform.

In the ac triggering mode, obtained by setting the red TRIGGER SELECTOR control in the AC position, suitable triggering can be obtained with signals from about 16 Hz to 5 MHz. When the ac triggering mode is used, it is usually

necessary to adjust the STABILITY and TRIGGERING LEVEL controls to obtain stable triggering. In this way, the point on the signal at which triggering will occur may be selected; in the automatic triggering mode, triggering occurs at the average dc level of the waveform.

The dc triggering mode, obtained by setting the red TRIG-GER SELECTOR control in the DC position, is particularly useful in triggering from very low-frequency waveforms. The trigger pulses are generated when the signal reaches a given dc level relative to ground. Suitable triggering can be obtained from triggering signals from dc to about 5 MHz. In the dc triggering mode, as in the ac triggering mode, it is usually necessary to adjust the STABILITY and TRIGGERING LEVEL controls to obtain stable triggering.

The high-frequency synchronized mode, obtained by setting the red TRIGGER SELECTOR control in the HF SYNC position, should be used to display signals from about 5 MHz to 15 MHz. In this mode of operation, the Time-Base Trigger circuit is bypassed and the triggering signal itself, either internal or external, synchronzies the free-running operation of the Time-Base Generator. For this reason, neither the TRIGGERING LEVEL control nor the + and - feature of the black TRIGGER SELECTOR control lave any effect on the display. The sweep will always be started on a negative-going slope of the signal. It is usually necessary to adjust the STABILITY control to obtain a stable display.

# Adjusting the STABILITY and TRIGGERING LEVEL Controls

As mentioned previously, it is not necessary to adjust the STABILITY and TRIGGERING LEVEL controls as long as stable triggering can be obtained in the automatic mode of operation. However, if stable triggering cannot be obtained in the automatic mode, it will be necessary to use one of the other modes. In this case, it will usually be necessary to adjust the STABILITY control or the TRIGGERING LEVEL control, or both, to obtain a stable presentation.

Normally, in the ac and dc modes, the STABILITY control can be set in the PRESET position, and stable triggering obtained through use of the TRIGGERING LEVEL control only. When it is not possible to obtain stable triggering with the STABILITY control in the PRESET position, then you should use the following procedure in adjusting the STABILITY and TRIGGERING LEVEL controls:

- 1. Set both controls fully clockwise.
- Turn the STABILITY control counterclockwise until the trace just disappears.
- 3. Set the TRIGGERING LEVEL control for a stable display.

When the TRIGGERING LEVEL control is set in the + region of its range, triggering of the sweep will take place at a point on the triggering signal above its average dc level. When the TRIGGERING LEVEL control is set in the — region of its range, triggering of the sweep will take place at a point on the triggering signal below its average dc level. With some signals it will be possible to obtain stable triggering at several positions of the TRIGGERING LEVEL control

or throughout a considerable portion of its range. You should set the control where it produces the most favorable display for the purpose desired.

When the red TRIGGER SELECTOR control is in the HF SYNC position, the TRIGGERING LEVEL control has no effect and a stable display is obtained through the use of the STA-BILITY control only. In this case, rotate the STABILITY control through its range until the best display is obtained on the screen.

#### Setting the Sweep Rate

Throughout the operation of the oscilloscope, the TIME/DIV. control may be adjusted at any time to present the desired waveform on the crt. For most applications, the TIME/DIV. control is set to display several cycles of a recurring waveform. Sometimes, when it is desired to inspect some portion of a single cycle closely, it will be best to adjust the TRIGGERING LEVEL control so that the portion of the waveform to be inspected is at the extreme left-hand edge of the graticule and then set the TIME/DIV. control to the fastest sweep rate (smallest numercial setting) possible, still keeping the desired portion of the waveform on the screen. The black numbers in the inner ring around the TIME/DIV. control indicate the sweep rate when the HORIZ. DISPLAY switch is in the NORM, position.

# Magnification of the Sweep

Any part of the trace may be expanded horizontally by a factor of 5 through the use of the HORIZ. DISPLAY switch. To expand a given portion of the sweep, first set that portion to the center of the graticule by means of the HORIZONTAL POSITION control. Then set the HORIZ. DISPLAY switch to the MAG. position. The magnified sweep feature is particularly useful for inspecting portions of the waveform which cannot be expanded sufficiently and still kept on the screen by the method described in the previous paragraph. The small blue numbers in the outer ring around the TIME/DIV. control give the true sweep rates when the HORIZ. DISPLAY switch is in the MAG. position.

#### FREE-RUNNING OPERATION

When the STABILITY control is set fully clockwise (and the red TRIGGER SELECTOR control is not in the AUTO. position), the Time-Base Generator free runs to produce a sweep independent of any synchronizing signal. The frequency of the sweep can be varied from about 0.012 Hz to 100 kHz by means of the TIME/DIV. control and the associated VARIABLE control.

Two outputs coincident with the sweep are available at front-panel connectors. A positive gate about 25 volts in amplitude is available at the +GATE OUT connector, and a positive-going sawtooth waveform, rising to a maximum amplitude of about 150 volts, is available at the SAWTOOTH OUT connector. Applications for free-running operation are discussed later in this section.

# **Applications**

The following paragraphs describe procedures for making measurements of voltage, elapsed-time, and phase-shift with the Type 516 Oscilloscope. No attempt has been made to describe any but the most basic techniques. Familiarity with the instrument will enable the operator to apply the essence of these techniques to a wide variety of applications, depending upon the problem at hand.

# **Voltage Measurements**

To measure the ac component of a signal, proceed as follows:

- 1. Display the waveform over as large a portion (vertically) of the screen as possible by adjusting the appropriate VOLTS/DIV. control.
- 2. With the aid of the graticule, measure the vertical distance (in graticule divisions) between the two points on the waveform at which the voltage measurement is desired. Make sure the appropriate VARIABLE VOLTS/DIV. control is in the CALIBRATED position. On small voltage measurements, the width of the trace can make up an appreciable part of the entire measurement. For this reason, it is important to take all readings in a given measurement from the same side of the trace.
- 3. Multiply the distance between the two points by the setting of the appropriate VOLTS/DIV. control and by the attenuation factor, if any, of the probe. This is the voltage between the two points of the waveform.

To measure the dc level at some point on a signal, proceed as follows:

- 1. Set the POLARITY switch of the channel to which the signal is applied to the NORM. DC position.
- 2. Set the corresponding VOLTS/DIV. control such that the expected voltage (at the oscilloscope input connector) is not more than six times the setting of the control. Make sure the VARIABLE VOLTS/DIV. control is in the CALIBRATED position.
- 3. Set the oscilloscope controls to produce a free-running trace.
- 4. Touch the oscilloscope probe tip to a ground point, and with the appropriate VERTICAL POSITION control position the trace so that it lies along one of the horizontal lines of the graticule. This line will be used as a ground reference line; its position in any given case will depend upon the polarity and amplitude of the input signal to be measured. Do not adjust the VERTICAL POSITION control after the reference line has been established.
- 5. Remove the probe tip from ground and connect it to the signal source. Adjust the triggering controls for a stable display.
- 6. Measure the distance, in graticule divisions, from the ground reference line established in step 4 to the point at which the dc voltage level is desired.

7. Multiply this distance by the setting of the appropriate VOLTS/DIV. control and the attenuation factor, if any, of the probe. This is the dc level of the point measured.

# **Time and Frequency Measurements**

To measure the time interval between two points on a waveform, proceed as follows:

- 1. Apply the signal to either of the Vertical Channels and set the triggering controls for a stable display. Make sure the VARIABLE TIME/DIV. control is in the CALIBRATED position.
- 2. Measure the horizontal distance, in graticule divisions, between the two points whose interval you wish to find.
- 3. Multiply the distance measured by the setting of the TIME/DIV. control. (If the HORIZ. DISPLAY switch is in the MAG. position, use the blue numbers). This is the time interval between the two points measured.

To determine the frequency of a recurrent waveform, simply take the reciprocal of the time interval between corresponding points on two consecutive cycles of the waveform.

# **Phase-Shift Measurements**

A phase comparison of two signals of the same frequency can be made by making use of the dual-trace feature of the Type 516 Oscilloscope. To make this comparison, proceed as follows:

- 1. Apply the two signals to the two input connectors and set the MODE switch to the CHOPPED or ALTERNATE position. (In general, the CHOPPED position is more suitable for low-frequency signals and the ALTERNATE position is more suitable for high-frequency signals). Apply one of the signals as an external trigger and set the triggering controls for a stable presentation. Center both signals vertically on the graticule.
- 2. Set the TRIGGERING LEVEL and HORIZONTAL PO-SITION controls so that the leading waveform starts exactly at the point where the horizontal centerline meets the vertical line at the left-hand edge of the graticule.
- 3. Measure the distance, in graticule divisions, from the left-hand edge of the graticule to the point where the other waveform crosses the horizontal centerline. For the most accurate measurement, use the fastest sweep rate possible. Make sure that the VARIABLE TIME/DIV. control is in the CALIBRATED position.
- 4. Multiply the distance measured in step 3 by the setting of the TIME/DIV. control and divide by the period of one cycle of the waveform. Multiply the result by 360°. This is the phase difference between the two signals.

# **Application of Free-Running Operation**

Sometimes it is desired to display the output of a device which requires a triggering signal for each cycle of output, such as a monostable or bistable multivibrator. In this case, the Type 516 Oscilloscope can be used for triggering the device as well as displaying its output waveform. To set the oscilloscope up for this type of operation, proceed as follows:

- 1. Set the STABILITY control fully clockwise.
- 2. Connect a lead between either the +GATE OUT connector or the SAWTOOTH OUT connector and the input to the device under observation. (The signal at the +GATE OUT connector has an amplitude of about 25 volts and the signal at the SAWTOOTH OUT connector has an amplitude of about 150 volts; therefore, it may be necessary to scale

these down by means of a voltage divider, depending upon the equipment with which they are to be used.)

#### CAUTION

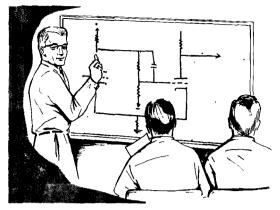
To avoid shorting out the +gate or sawtooth voltage from the oscilloscope, make the connection to the device under observation before connecting the lead to the oscilloscope connector. Do not connect a resistance of less than about 2k between the +GATE OUT connector and ground, or less than about 10k between the SAWTOOTH OUT connector and ground; to do so might damage components in the oscilloscope.

3. Connect the output of device under observation to the vertical input connector of the Type 516.

The output of the device will now be synchronized to the free-running sweep of the oscilloscope and will appear as a stable display on the screen.

# **NOTES**

# **SECTION 4**



# **GENERAL DESCRIPTION**

A block diagram of the Type 516 Oscilloscope is shown in Fig. 4-1. In general, operation of the oscilloscope is as follows:

The signal or signals to be displayed are applied to the Vertical Amplifier through the A and/or B input connectors. The Vertical Amplifier amplifies the signals and applies them through the Delay Line to the vertical deflection plates of the cathode-ray tube. A sample of the signal is taken out at the Vertical Amplifier and applied to the Time-Base Trigger. The Time-Base Trigger generates a trigger pulse which bears

# CIRCUIT

a selected fixed time relationship to the incoming signal. This trigger pulse triggers the Time-Base Generator which generates a linear sawtooth waveform. The sawtooth waveform is applied through the Horizontal Amplifier to deflect the crt beam horizontally at a selected fixed rate. Provisions are also made for generating trigger pulses which bear a selected fixed time relationship to some external signal or to the ac line waveform. If desired, the Time-Base Generator may be disabled and an external signal applied directly to the Horizontal Amplifier. The Amplitude Calibrator provides an amplitude-calibrated square wave for use in calibrating the gain of the Vertical Amplifier and adjusting the compensa-

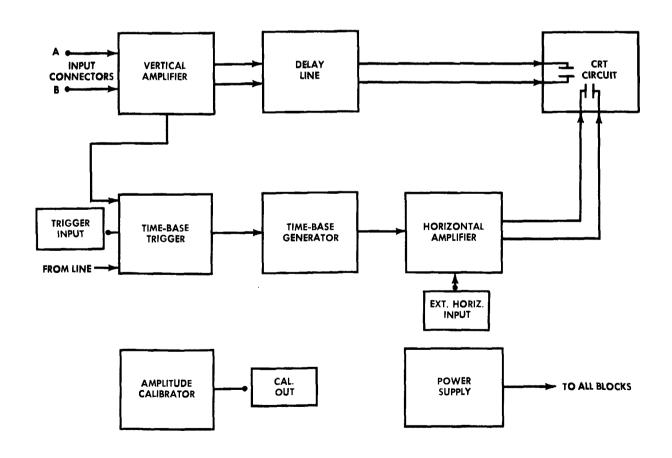


Fig. 4-1. Type 516 Oscilloscope block diagram.

#### Circuit Description - Type 516

tion of probes. The Power Supply supplies regulated voltages and currents as required throughout the oscilloscope.

The remainder of this section of the manual presents a more detailed description of the operation of each of the oscilloscope circuits. Throughout the discussion, you should refer to the circuit diagrams contained in the Diagrams section of this manual.

## **VERTICAL AMPLIFIER**

The Vertical Amplifier of the Type 516 Oscilloscope contains two separate Input Amplifier Channels (A and B), a common Output Amplifier, and a Switching Circuit. Signals may be applied to both Input Amplifier Channels, and the Switching Circuit makes it possible to display one signal as a single trace on the crt or to display both signals as two separate traces. It does this by turning the outputs of the two Input Amplifier Channels alternately off and on, either on alternate sweeps of the trace or at a 150-kHz switching rate.

The Type 516 can also be used as a conventional single-channel oscilloscope with an input signal on one channel only.

A block diagram of the Vertical Amplifier is shown in Fig. 4-2.

# **Input Amplifier Channels**

Since the two Input Amplifier Channels are identical, except for the CHANNEL A GAIN SET adjustment, the following discussion refers to the A Channel only. Operation of the B Channel is identical.

The signal to be displayed is applied through the POLARITY switch and the VOLTS/DIV. control to one of the grids of the Input Cathode Follower, V423. When the POLARITY switch is in either of the AC positions, the signal is coupled through C400 which prevents the dc component of the signal from being applied to the Input Cathode Follower grid. When the POLARITY switch is in one of the NORM. positions, the signal is applied to the grid of V423A and is displayed on the crt in the normal position, with the positive portions toward the top of the screen. When the POLARITY switch is in one of the INV. positions, the signal is applied to the grid of V423B and is displayed on the crt in the inverted position, with the positive portions toward the bottom of the screen.

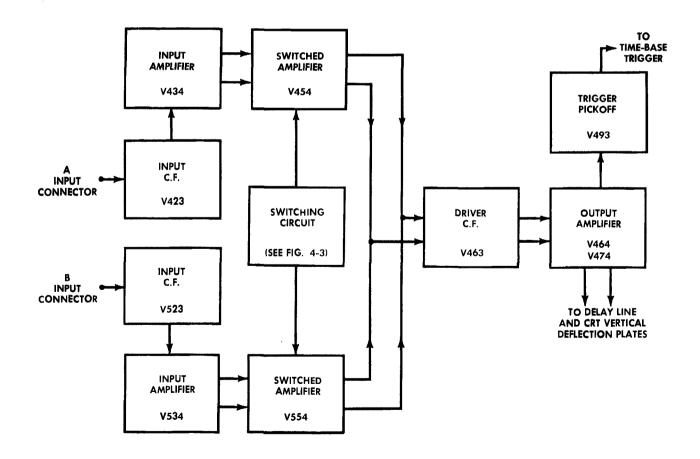


Fig. 4-2. Vertical Amplifier block diagram.

The basic sensitivity of the Vertical Amplifier is 0.05 volt per divison. In other words, 0.05 volt of signal at the grids of the Input Cathode Followers will produce one division of deflection on the crt. Changes in sensitivity of the oscilloscope are accomplished by attenuation of the input signal before it reaches the Input Cathode Followers. A "straight-through" position and eight different values of attenuation may be selected by the VOLTS/DIV. control to provide nine fixed calibrated sensitivities of 0.05, 0.1, 0.2, 0.5, 1, 2, 5, 10, and 20 volts per division. The attenuators are resistance and capacitance dividers which provide constant attenuation throughout the frequency range of the instrument while maintaining a constant input impedance of 1 megohm and 20 picofarads at the input connectors.

The Input Amplifier, V434, is a cathode-coupled paraphase amplifier which converts the single-ended input from the Input Cathode Follower to a push-pull output. The gain of this stage can be adjusted by means of the VARIABLE VOLTS/DIV. control which varies the coupling resistance between the two cathodes. This control permits a continuous range of uncalibrated sensitivity values between the steps selected by the VOLTS/DIV. control and up to about 50 volts per division of deflection. Switch SW442, which switches the UNCALIBRATED lamp, B442, in and out of the circuit, is linked mechanically to the VARIABLE VOLTS/DIV. control. Thus, when the VARIABLE VOLTS/DIV. control is moved away from the CALIBRATED position (fully clockwise), the UNCALIBRATED lamp is energized to indicate that the vertical sensitivity is not calibrated.

The CHANNEL A GAIN SET is used to set the gain of the A Channel equal to that of the B Channel. This provides equal deflection with equal signals through both channels, provided both VOLTS/DIV. controls are set to the same position and both VARIABLE controls are in the CALIBRATED position.

The DC BAL, control in the cathode circuit of V423B is adjusted to place both cathodes of the Input Amplifier at the same potential so that under no-signal conditions there is no current flowing through R440 and R441. This provides vertical stability of the trace as the value of R441 is changed with the VARIABLE VOLTS/DIV, control.

The POSITION control varies the average dc level about which the signal moves at the grids of the Switched Amplifier, V454. Since the entire Vertical Amplifier is dc coupled, this varies the position of the trace on the screen. The POSITION control also affects the average dc plate voltage of the Input Amplifier and, therefore, can disturb the dc balance of that circuit. For this reason, a small portion of the positioning voltage at each plate is fed back to the opposite cathode (through R443 and R444) to counteract this effect.

The turning off and on of the two channels is accomplished in the Switched Amplifiers by the operation of the Switching Circuit, which will be discussed later in this section.

## **Output Amplifier**

The push-pull output of the Switched Amplifier is applied to the Driver Cathode Followers, V463, which, in turn, drive the Output Amplifier, V464 and V474. The output of the Output Amplifier is connected through the Delay Line to the vertical deflection plates of the crt. The Delay Line delays the application of the signal to the deflection plates until

the crt has been unblanked and the horizontal sweep started. This delay allows the leading edge of fast-rising pulses to be displayed.

The overall gain of the Vertical Amplifier is adjusted by varying the degeneration between the two cathodes of the Output Amplifier. This is done with the GAIN ADJ. potentiometer, R478.

The parallel network of D477, D478, and R477 lowers the effective resistance between the two cathodes as the dc potential between them exceeds about 0.2 volt. The purpose of this network is to compensate the compression of the crt display which might otherwise occur when the POSITION control is near the end of its range.

# Trigger Pickoff

The output from V464 is also applied through two cathode followers, V493A and V493B, to the TRIGGER SELECTOR switch, SW10A. Thus, when SW10A is in either of the INT. positions, a portion of the output of the Output Amplifier is applied to the Time-Base Trigger to start the horizontal sweep.

B493 and B494 protect V493 against cathode-to-grid breakdown when the instrument is first turned on by holding the cathodes within a safe operating level of the grids until the heater has warmed up enough to allow the tube to conduct. After V493 starts conducting, B493 and B494 stop conducting and are therefore effectively removed from the circuit.

### **Switching Circuit**

Selection of the input channel whose output is to be applied to the Driver Cathode Followers is accomplished by means of the Switching Circuit. A block diagram of the Switching Circuit is shown in Fig. 4-3.

The Switching Circuit may be operated in any one of four modes, depending upon the setting of the MODE switch, SW585. When the MODE switch is in the A ONLY position, the grid of V594A (one side of the Switching Amplifier) is at about ground potential and the grid of V594B is at about -96 volts. Therefore, V594A is conducting heavily, and V594B is cut off. The plate of V594A and the cathodes of the A Channel Switched Amplifier are at about +57 volts. The plate of V594B and the cathodes of the B Channel Switched Amplifier are at about +63 volts (they are prevented from going further positive by conduction through D554). The average dc potential of the grids of both Switched Amplifiers is about +56 volts. Therefore, with only about one volt of bias, the A Channel Switched Amplifier will conduct and amplify any signal applied to its grids. Meanwhile, with about 7 volts of bias, the B Channel Switched Amplifier will be cut off. So the signal in the A Channel is applied through the Driver Cathode Followers and the Output Amplifier to the crt deflection plates, and the signal in the B Channel is blocked at the B Channel Switched Amplifier.

When the MODE switch is in the B ONLY position, just the opposite set of conditions exists; the B Channel Switched Amplifier conducts and the A Channel Switched Amplifier is cut off. So the B Channel signal is displayed on the crt.

When the MODE switch is in the ALTERNATE position, the crt displays the signal in one channel for one sweep of

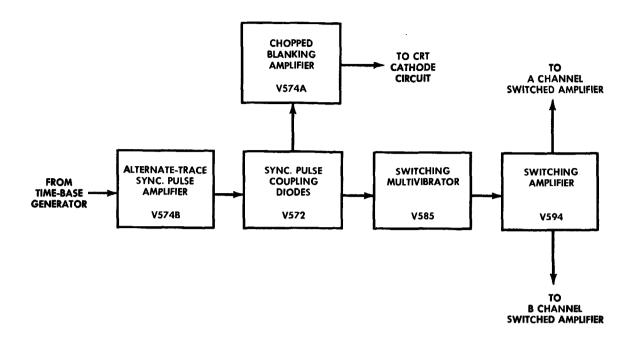


Fig. 4-3. Switching Circuit block diagram.

the beam, and then the signal in the other channel for the next sweep of the beam. In this mode of operation, V585 becomes a bistable Switching Multivibrator which is switched from one state to the other at the end of each sweep. When the Switching Multivibrator is in one state—say, with V585A conducting and V585B cut off—the low plate voltage on V585A is coupled to the grid of V594B. This cuts off V594B whose plate voltage then rises to about +63 volts, which cuts off the B Channel Switched Amplifiers. The high plate voltage of V585B is coupled to the grid of V594. This allows V594A to conduct which lowers its plate voltage and the voltage at the cathodes of the A Channel Switched Amplifiers to about +57 volts. Therefore, the A Channel Switched Amplifiers conduct. When the Switching Multivibrator is in the other state—V585B conducting and V585A cut off—the opposite set of conditions exists; the B Channel Switched Amplifiers conduct and the A Channel Switched Amplifiers are cut off.

The switching of the Switching Multivibrator in the ALTERNATE mode takes place as follows. At the end of each sweep, a positive spike is formed at the screen of V145A in the Time-Base Generator. This positive spike is applied to the grid of the Alternate Trace Sync Pulse Amplifier, V574B, and, since V574B is normally cut off, allows the tube to conduct. This allows current to flow also through the Sync Pulse Coupling Diodes, V572, and applies a negative spike to both grids of the Switching Multivibrator. Since one side of the multivibrator is already cut off, the negative spike will have no effect on that side. However, the negative spike will cut off the side which is conducting and cause the multivibrator to switch states.

When the MODE switch is in the CHOPPED position, V585 becomes an astable multivibrator with a frequency of about 150 kc. The outputs from the two sides of the Switching Multivibrator are applied to the two sides of the Switching

Amplifier which turns the Switched Amplifiers off and on alternately at a 150-kc rate. This causes the signals in the two channels to be displayed on the crt alternately for a little over three microseconds at a time.

Since the POSITION controls of the two channels are normally set at different levels, it is necessary, in the CHOP-PED mode, to turn off the crt beam as the switching between channels is taking place. This action takes place in the Chopped Blanking Amplifier, V574A. The negative-going pulses formed as the two sides of the Switching Multivibrator alternately come into conduction are differentiated and applied as negative spikes to the grid of V574A. V574A amplifies and inverts the spikes and applies them to the crt cathode via SW848. These positive spikes on the crt cathode, then, cut off the beam as switching is taking place.

#### TIME-BASE TRIGGER

The Time-Base Trigger consists basically of the Trigger Selector switches, SW10A and SW10B, the Trigger Input Amplifier, V24, and the Trigger Multivibrator, V45. The TRIGGER SELECTOR switches select the triggering source, the triggering slope, and the triggering mode. The Trigger Input Amplifier amplifies (and, when desired, inverts) the incoming triggering signal and applies it to the input of the Trigger Multivibrator. The Trigger Multivibrator is a Schmitt circuit which is switched from one state to the other by the signal at its input. Its square-wave output is differentiated to form negative and positive spikes which are applied to the Time-Base Generator where the negative spikes are used to start the horizontal sweep. The positive spikes are not used

A block diagram of the Time-Base Trigger is shown in Fig. 4-4.

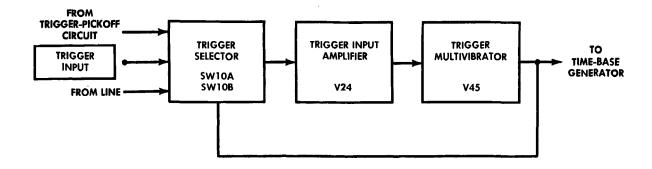


Fig. 4-4. Time-Base Trigger block diagram.

#### Trigger Input Amplifier

The input to the Trigger Input Amplifier, V24, may be selected from any one of three sources by means of the black TRIGGER SELECTOR control, SW10A. When SW10A is in one of the INT. positions, the signal is obtained from the Trigger Pickoff circuit in the Vertical Amplifier. When SW10A is in one of the EXT. positions, the signal may be obtained from an external source through the TRIGGER INPUT connector on the front panel. When SW10A is in one of the LINE positions, the signal is obtained from one of the 6.3-volt secondary windings of the line transformer.

The + and — positions of SW10A provide a means of inverting or not inverting, as desired, the triggering signal in the Trigger Input Amplifier. This is done so that the negative spike at the output of the Time-Base Trigger can be made to occur during either a positive-going or a negative-going portion of the triggering signal. (The negative spike occurs only when there is a negative-going signal at the input to the Trigger Multivibrator.)

When SW10A is in any of the —positions, the incoming triggering signal is applied to the grid of V24A, and V24 is a cathode-coupled amplifier. With this configuration, the signal at the plate of V24B (output of the Trigger Input Amplifier) is in phase with the incoming triggering signal. When SW10A is in any of the +positions, the incoming triggering signal is applied to the grid of V24B, and V24B is a plateloaded amplifier. In this case, the signal at the plate of V24B is 180° out of phase with the incoming triggering signal.

The TRIGGERING LEVEL control, R17, varies the average dc level at the plate of V24B from about +60 volts to +100 volts. This is true whether SW10A is in a + position or a — position. The minimum level of +60 volts represents the point where V24B is taken into saturation by a sufficiently positive setting of R17 when SW10A is in a — position or by a sufficiently negative setting of R17 when SW10A is in a + position. The maximum level of +100 volts represents

the point where V24B is taken into cutoff by a sufficiently negative setting of R17 when SW10A is in a — position or by a sufficiently positive setting of R17 when SW10A is in a + position. As will be seen later, the voltage at the plate of V24B must pass through the approximate center of this range (about +80 volts) in order to cause the Trigger Multivibrator to change states.

For small triggering signals, R17 is set such that the average dc level at the plate of V24B is close to the center of its range so that the small signal, as amplified by V24, is sufficient to carry the plate voltage through the +80-volt point. When a large triggering signal is applied and it is desired to trigger on an extreme positive or negative point of it, R17 is set such that V24B is well into saturation, or cutoff (depending upon whether triggering is desired at a more negative or more positive point on the signal and on a negative-going or positive-going slope). In this case, the triggering signal must be sufficient to overcome the saturation or cutoff of V24B and produce an additional 20 volts of swing at the plate of V24B in order to cause the Trigger Multivibrator to switch states.

It should be noted that the voltages given in the foregoing discussion are typical nominals only and will vary somewhat from instrument to instrument and with time.

#### Trigger Multivibrator

The Trigger Multivibrator is a two-state Schmitt circuit. When the voltage at its input grid (grid of V45A) is above a certain critical level (neglecting hysteresis), the Trigger Multivibrator is in one state, with V45A conducting and V45B cut off. When the Trigger Multivibrator is in this state, the voltage as its output (plate of V45B) is at 300 volts. When the voltage at the input grid is below the critical level (still neglecting hysteresis), the Trigger Multivibrator is in the other state, with V45A cut off and V45B conducting. When the Trigger Multivibrator is in this second state, the voltage at its output is about +290 volts. The transition from one

# Circuit Description — Type 516

state to the other occurs very rapidly regardless of how slowly the voltage at the input grid passes the critical level. Thus the output of the Trigger Multivibrator is a 10-volt square wave. The negative-going transition occurs when the voltage at the input passes the critical level while moving in the negative direction, and the positive-going transition occurs when the voltage at the input passes the critical level while moving in the positive direction. As mentioned before, only the negative-going transistion is of significance timewise, and, by means of the black TRIGGER SELECTOR control and the TRIGGERING LEVEL control, this point can be made to coincide with virtually any point on the incoming triggering signal.

The TRIG. LEVEL CENTERING adjustment, R39, varies the level of the grid of V45B with respect to the plate of V45A. Thus, it controls the input voltage at which the Trigger Multivibrator changes states. This level is normally set at the voltage which the plate of V24B assumes when both grids of V24 are at ground potential. In most instruments, this is within a volt or two of +80 volts.

Actually, the input voltage level at which the Trigger Multivibrator changes states on a negative-going signal is normally slightly lower than the input voltage level at which it changes states on a positive-going signal. The difference between the two input voltage levels at which the two changes in state occur is the hysteresis of the circuit. In order to obtain maximum triggering sensitivity, the hysteresis must be kept as small as possible. To reduce the hysteresis, resistance is introduced between the two cathodes by means of the TRIGGER SENSITIVITY adjustment, R47. Due to the drop across this resistance, the cathode voltage of the nonconducting tube is moved closer to its grid voltage, so that the grid will not have to move as far to bring the tube into conduction and cause the multivibrator to change states. Too much resistance between the cathodes causes the multivibrator to oscillate during transitions from one state to the other.

As will be seen in the discussion of the Time-Base Generator, not every negative trigger pulse from the Time-Base Trigger initiates a sweep. Negative trigger pulses which arrive at the Time-Base Generator during the time that a sweep is in progress will have no effect on the circuit. It is only after a sweep has been completed and all circuits have returned to their quiescent state that the Time-Base Generator will be retriggered by a trigger pulse from the Time-Base Trigger.

## **Triggering Mode**

The red TRIGGER SELECTOR control, SW10B, selects the type, or mode, of triggering. In the DC position, the triggering signal is dc-coupled to the Trigger Input Amplifier, which in turn is dc-coupled to the input of the Trigger Multivibrator. R30 isolates the plate of V24B from the capacitance of the switch; R32 isolates the grid of V45A from the switch. It should be noted that with SW10A in the INT. position, and SW10B and the appropriate POLARITY switch in the DC position, complete dc-coupling exists from the input of the Vertical Amplifier to the Trigger Multivibrator.

In the AC position of SW10B, capacitor C10 is connected into the input circuit. This prevents the dc component of the

triggering signal from reaching the Trigger Input Amplifier. The Trigger Input Amplifier, however, is still dc-coupled to the Trigger Multivibrator.

In the AUTO. position of SW10B, the Trigger Multivibrator is converted from a bistable configuration to an astable (free-running) configuration by the addition of feedback from the grid of V45B to the grid of V45A through R40. In addition, the dc coupling between the Trigger Input Amplifier and the Trigger Multivibrator is removed when the switch is in the AUTO. position.

To understand the operation of the Trigger Multivibrator in the free-running mode of operation, first assume that V45B is cut off and V45A is just being driven into cutoff by the charge on C31. The voltage at the plate of V45A starts to rise, carrying with it the voltage at the grid of V45B. V45B then starts to conduct, causing a negative step at its plate. Since the two grids are coupled through R40, the grid of V45A will start moving positively at the same time as the grid of V45B. However, the time constant of C31 and the resistances in the grid circuit of V45A is such that it takes about 0.01 second for the voltage at the grid of V45A to rise exponentially from its starting point, below cutoff, to a point where the tube will start conducting.

When V45A does start conducting, its plate voltage will drop, carrying with it the grid of V45B. V45B will cut off causing a positive step at its plate. At the same time that the grid of V45B goes negative, the grid of V45A will also start negative. Once again, it will take about 0.01 second for C31 to charge up sufficiently to cut V45A off. When V45A does cut off, the cycle starts over. Thus, the Trigger Multivibrator free runs at about 50 Hz.

During the automatic mode of operation, the total voltage change at the grid of V45A is about 3 volts. Since the grid of V45A is never more than 3 volts from cutoff, a triggering signal from the Trigger Input Amplifier with a peak-to-peak amplitude of 3 volts or more can drive the grid to cutoff at any time and produce a trigger output. Smaller signals than 3 volts can drive the V45A grid into cutoff if they occur at a time when the exponentially changing grid voltage is approaching the cutoff level of the tube (relatively later in the time interval during which V45A is conducting).

Hence, in the absence of any triggering signal, the Trigger Multivibrator free runs at about 50 Hz. However, since the triggering signals are still coupled to the Trigger Multivibrator through C31, virtually any triggering signal over 50 Hz in frequency and of sufficient amplitude will produce synchronized operation of the Trigger Multivibrator. The 50-Hz free-running sweep produced in the absence of a triggering signal provides a base line from which to make voltage measurements and also indicates that the instrument is adjusted to display any signal that might be applied to the input.

The Trigger Multivibrator has a maximum switching frequency of about 5 MHz. Therefore, when it is desired to display signals above that frequency, the Time-Base Trigger is bypassed by placing SW10B in the HF SYNC position and the incoming signal is applied directly to the Time-Base Generator. The STABILITY control is advanced to produce a free-running sweep and the signal becomes a synchronizing signal, rather than a triggering signal, which synchronizes the free-running sweep to a submultiple of its frequency.

#### TIME-BASE GENERATOR

The Time-Base Generator, upon receipt of a negative trigger pulse from the Time-Base Trigger, produces a linearly rising (sawtooth) voltage which is applied through the Horizontal Amplifier to the crt horizontal deflection plates. This causes the spot to move from left to right on the crt screen and form the trace. The amplitude of the sawtooth voltage is about 150 volts. Its rate of rise is controlled by the values of the Timing Capacitor and Timing Resistor switched into the circuit by the TIME/DIV. control on the front panel.

The Time-Base Generator consists of three main circuits; the Sweep-Gating Multivibrator, the Miller Runup Circuit, and the Hold-Off Circuit. The Sweep-Gating Multivibrator consists of V135A, V135B, and V145A. The essential components of the Miller Runup Circuit are: The Miller Runup Tube, V161A; the Runup Cathode Follower, V161B; the Disconnect Diodes, V152A and V152B; the Timing Capacitor, C160; and

the Timing Resistor, R160. The Hold-Off Circuit consists of the Hold-Off Cathode Followers, V183A and V183B; the Hold-Off Capacitor, C180; and the Hold-Off Resistors, R180A and R180B. Also considered a part of the Time-Base Generator are the Unblanking Cathode Follower, V145B, the + Gate-Out Cathode Follower, V193A, and the Sawtooth Out Cathode Follower, V193B.

A block diagram of the Time-Base Generator is shown in Fig. 4-5.

#### **Sweep Generation**

In the quiescent state—that is, when no sweep is being generated—V135A is conducting and V145A is cut off. (The STABILITY control, R110, or the PRESET STABILITY adjustment, R111, whichever is in the circuit, is set so that the grid of V135A is just above cutoff voltage.) The plate of V145A is

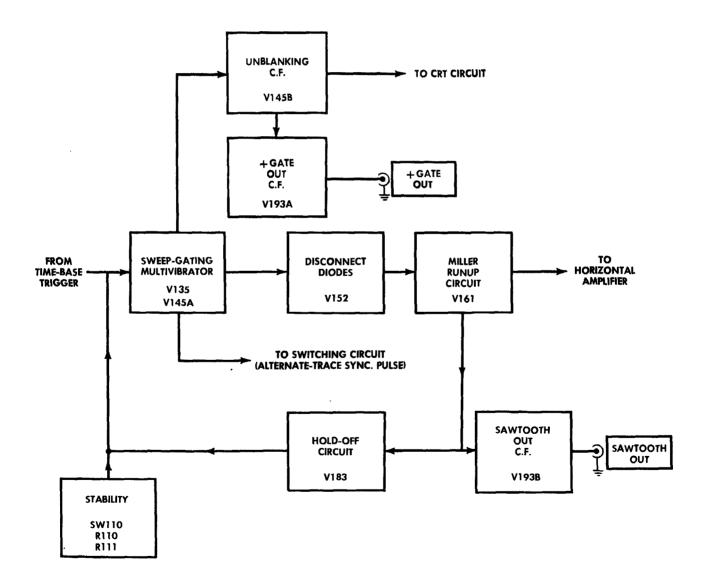


Fig. 4-5. Time-Base Generator block diagram.

at about -2.5 volts with respect to ground. The Disconnect Diodes are conducting and hold both sides of the Timing Capacitor at about -2.5 volts. With its cathode grounded and its grid at -2.5 volts, the Miller Runup tube, V161A, is conducting heavily and its plate is at about +30 volts.

A negative trigger pulse, arriving at the grid of V135A from the Time-Base Trigger, causes the Sweep-Gating Multivibrator to switch rapidly to its other state. That is, V135A cuts off and V145A conducts. V135A is held in cut off after the trigger pulse passes by a rise in voltage at the common cathode connection. As V145A conducts, its plate voltage goes down, cutting off the Disconnect Diodes. When the Disconnect Diodes cut off, the plates of the Timing Capacitor are no longer held at -2.5 volts, and the Timing Capacitor starts to charge toward the instantaneous potential difference between the -150-volt supply and the potential at the cathode of V161B. However, as the lower side of the Timing Capacitor starts to move in a negative direction, it takes the grid of V161A with it. This produces a positive swing at the plate of V161A which is coupled, through B167 and V161B, to the top of the Timing Capacitor. This increases the voltage to which the Timing Capacitor is trying to charge. The effect is to "straighten out" the charging curve by increasing the charging voltage with each increment of charge on the capacitor. The positive swing at the top of the Timing Capacitor also tends to prevent the lower side from swinging negatively. Since the gain of V161A is about 150, the potential at the top of the Timing Capacitor moves about 150 volts with respect to ground while the potential at the lower side moves about one volt. The result is an extremely linear sawtooth at the cathode of V161B, which is applied through the Horizontal Amplifier to the horizontal deflection plates of the crt. This sawtooth is also coupled through the cathode follower V193B to the SAWTOOTH OUT connector on the front panel.

# Sweep Length

The length of the sweep—that is, the distance the spot moves across the crt—is determined by the setting of the SWP. LENGTH adjustment, R176. As the sweep voltage rises linearly at the cathode of V161B, there will be a proportionate rise in the voltage at the wiper arm of the SWP. LENGTH adjustment. This rise is coupled through the two Hold-Off Cathode Followers, V183A and V183B, to the grid of V135A. When the voltage rises to the point where V135A comes out of cutoff, the Sweep-Gating Multivibrator will rapidly revert to its original state with V135A conducting and V145A cut off. The voltage at the plate of V145A will then rise, carrying with it the voltage at the plates of the Disconnect Diodes.

V152B starts conducting and discharges the Timing Capacitor. This brings the grid of V161A quickly back up to its quiescent level. The rise in voltage at the grid causes the tube to conduct more, so the plate voltage drops, carrying with it the grid and cathode of V161B. When the voltage at the cathode of V161B returns to about —2.5 volts, V152A conducts, clamping the voltage at this point

The time during which the voltage at the plate of V161A and the cathode of V161B drops is the sweep trace interval. The trace is blanked during this time and during the hold-off time following it.

# **Hold-Off Circuit**

The Hold-Off Circuit prevents the Time-Base Generator from being triggered following the completion of a sweep until after the Miller Runup Circuit has stabilized in the quiescent condition. It does this by holding the grid of V135A positive enough to prevent it from being taken into cutoff by the negative trigger pulses from the Time-Base Trigger.

When no sweep is being generated, the arm of the SWP. LENGTH potentiometer and, therefore, the grid of V183B are at about -90 volts. Normal operating bias on V183B is about 5 volts, so the Hold-Off Capacitor, C180 and/or C181, is charged to about 85 volts. During the sweep, the Hold-Off Capacitor discharges to about 45 volts through V183B as a result of the rise in voltage on the grid. At the end of the sweep, the voltage at the grid of V183B drops back to -90 volts. The cathode tries to follow but is held up by the charge on the Hold-Off Capacitor. The Hold-Off Capacitor starts to charge again exponentially toward 85 volts, carrying the grid of V183A negative. The cathode of V183A and the grid of V135A follow the grid of V183A. When the Hold-Off Capacitor charges to the point where V183A cuts off (V183A is normally cut off between sweeps), it loses control over the grid of V135A, which then returns to the level established by the setting of the STABILITY or PRESET STA-BILITY control.

The amount of hold-off time, then, is determined by the value of capacitance and/or resistance switched into the Hold-Off Circuit by the TIME/DIV. control. (See the Timing Switch circuit diagram.) The amount of hold-off time required is determined by the sweep rate. For this reason the TIME/DIV. control changes the time constant of the Hold-Off Circuit as well as that of the Timing Circuit.

#### Sweep Stability

The STABILITY control, R110, (or the PRESET STABILITY adjustment, R111, depending upon the mode of operation) regulates the dc level at the grid of V135A. This control is adjusted so that the voltage at the grid of V135A is just high enough to hold V135A out of cutoff. When the circuit is adjusted in this manner, a sweep will be produced only when a negative trigger pulse from the Time-Base Trigger drives V135A into cutoff. Turning the STABILITY control fully clockwise (when SW10B is in the DC, AC, or HF SYNC position) places a sufficiently negative voltage on the grid of V135A that V135A cuts off immediately upon decay of the hold-off voltage and initiates the next sweep without a trigger pulse. The result is a free-running sweep whose period is the total of the sweep time plus the hold-off time at any given setting of the TIME/DIV. control.

#### Unblanking

The positive rectangular pulse appearing at the cathode of V135B in the Sweep-Gating Multivibrator is coupled through the Unblanking Cathode Follower, V145B, to the grid circuit of the crt to unblank the crt during sweep time. This unblanking pulse is also coupled through the cathode follower V193A to the +GATE OUT connector on the front panel.

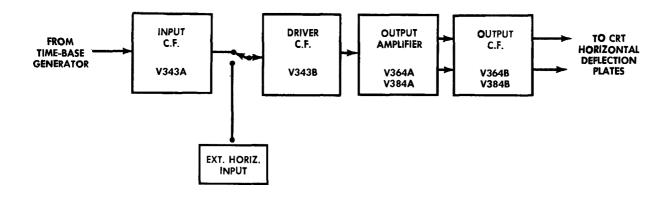


Fig. 4-6. Horizontal Amplifier block diagram.

#### HORIZONTAL AMPLIFIER

The Horizontal Amplifier consists of the Input Cathode Follower, the Driver Cathode Follower, the Output Amplifier, and the Output Cathode Follower stage. A block diagram of the Horizontal Amplifier is shown in Fig. 4-6.

The sweep waveform is coupled to the grid of the Input Cathode Follower, V343A, via the frequency-compensated voltage divider, R330-R332. The HORIZONTAL POSITION control, R333A, supplies a manually adjustable dc voltage to the grid of V343A for horizontal positioning of the crt beam. The C340-R340 network produces a small step at the start of the waveform at the faster sweep rates. This step is necessary to compensate for the bandpass-limiting effect of the stray capacitance in the amplifier. By its application the sweep will start linearly at the faster sweep rates. The Input Cathode Follower provides the necessary low impedance to drive the switch capacitances and the Driver Cathode Follower, V343B; the Driver Cathode Follower isolates the Output Amplifier from the HORIZONTAL DISPLAY switch.

In the MAG. position of the HORIZ. DISPLAY switch, the sweep waveform is coupled by the Driver Cathode Follower directly to the Output Amplifier, V364A-V384A. This stage is a cathode-coupled paraphase amplifier which converts the single-ended input to a push-pull output. The waveform is then coupled by the Output Cathode Follower stage, V364B-V384B, to the horizontal deflection plates. The MAG. GAIN ADJ., R259, varies the degeneration in the cathode circuit of the Output Amplifier and thus sets the gain of the stage. C260 reduces the degeneration at higher frequencies and thus compensates the amplifier for faster sweep rates. Bootstrap capacitors C364 and C384 also improve the response at the faster sweep rates by supplying current from the output cathode followers to charge the stray capacitance at the plates of the Output Amplifier. Neon lamp B346 is connected in the circuit when the HORIZ. DISPLAY switch is in the MAG. position to indicate that the magnifier circuits are in operation.

In the NORM, position of the HORIZ, DISPLAY switch, the gain of the amplifier is reduced by a factor of five by nega-

tive feedback. The feedback, from the cathode of V364B, is developed across R225 and R347, which have been switched into the grid circuit of V343B. When the HORIZ. DISPLAY switch is in the NORM. position, both the MAG. GAIN ADJ. R259, and the SWP. GAIN ADJ., R225, will vary the gain of the Horizontal Amplifier; for this reason, the MAG. GAIN ADJ. is adjusted only when the HORIZ. DISPLAY switch is in the MAG. position.

The SWP./MAG. REGIS. adjustment, R358, sets the dc-level at the grid of V343B so that the portion of the waveform at the horizontal center of the graticule will not shift as the HORIZ. DISPLAY switch is switched between the NORM. and MAG. positions.

In the EXT. position of the HORIZ. DISPLAY switch, the grid of the Driver Cathode Follower is connected to the EXT. HORIZ. INPUT connector on the front panel. With this arrangement, the horizontal signal is obtained from an external source rather than from the Time-Base Generator. The HORIZ.-INPUT ATTEN. control, R349 (ganged with the STA-BILITY control), provides adjustable attenuation of the externally applied signal. Horizontal positioning is provided by R333B rather than R333A when the HORIZ. DISPLAY switch is in the EXT. position.

Placing the HORIZ. DISPLAY switch in the EXT. position also disables the Time-Base Generator by opening the cathode circuit of the Sweep-Gating Multivibrator.

#### POWER SUPPLY

Plate and filament power for the tubes in the Type 516 Oscilloscope is furnished by a single transformer, T601. The primary has two equal tapped windings which are connected in parallel for 117-volt operation or in series for 234-volt operation. Two auxiliary windings are also provided to furnish aiding or bucking currents, as desired, for certain other primary source voltages (see Section 2).

The three main full-wave power supplies furnish regulated voltages of -150, +100, and +300 volts. The +300-volt supply also has an unregulated output of about 400 volts

for the high-voltage supply for the crt. It is unnecessary to regulate this supply as the high-voltage supply has its own regulating circuit.

Reference voltage for the —150-volt supply is established by a gas diode, VR tube V609. This tube, which has a constant voltage drop, establishes a fixed potential of about —84 volts at one grid of the Difference Amplifier, V634. The grid potential for the other half of the Difference Amplifier is obtained from a voltage divider consisting of R615, R616 and R617. The setting of the —150 ADJ., R616, determines the percentage of the total output voltage that appears at the grid of V634B and thus determines the total voltage across the divider. When this adjustment is properly set, the output voltage of the —150-volt supply will lie somewhere between —147 volts and —153 volts.

Should the loading on the supply tend to change the output voltage, the potential at the grid of V634B will change in proportion, and an error voltage will exist between the two grids of the Difference Amplifier. The error signal is amplified by V634A, whose plate is dc-coupled to the grid of the Series Tube, V627. The error voltage appearing at the grid of V627 will change the bias and, hence, the effective resistance of V627. This will allow more or less current, as required, to flow through the load to bring the output voltage back toward its original level. C617 improves the ac gain of the feedback loop, and thus increases the response of the circuit to sudden changes in output voltage.

The -150-volt supply serves as a reference for both the +100-volt and +300-volt supplies. In the +100-volt supply, the voltage divider R650-R651 establishes a voltage of essentially zero at the grid of the Amplifier, V654. (The actual voltage at this grid will be equal to the bias voltage required by the tube.) If the loading should tend to change the output voltage, an error voltage will appear at the grid of the amplifier. The error voltage will be amplified and will appear at the grid of the Series Tube, V667A. The cathode of V667A will follow the grid, and thus the output voltage will be returned to its established value of +100 volts. C650 improves the response of the regulator circuit to sudden changes in output voltage.

A small sample of the unregulated-bus ripple will appear at the screen of V654 through R657. This ripple signal appearing at the screen (which acts as an injector grid) will produce a ripple component at the grid of V667A which is opposite in polarity to the ripple appearing at the plate of V667A. This tends to cancel the ripple at the cathode of V667A, and hence reduces the ripple on the  $\pm 100$ -volt bus. This same circuit also improves the regulation of the circuit in the presence of line voltage variation.

The +300-volt supply functions in the same manner as the +100-volt supply. Rectified voltage from terminals 8 and 9 of the power transformer is added to the voltage supplying the +100 volt regulator to supply power for the +300-volt regulator. As mentioned previously, the +300-volt supply also provides an unregulated output of about 400 volts for the crt high-voltage supply.

# **CRT CIRCUIT**

A 60-kHz Hartley oscillator circuit furnishes energy for the three power supplies that provide accelerating voltages for the crt. The main components of the oscillator are V800 and the primary of T801 tuned by C808. The rectifier circuits are the half-wave type, with capacitor-input filters. Separate supplies are required for the grid and cathode circuits in order to provide dc-coupled unblanking to the grid supply.

V822 supplies —1675 volts for the cathode of the crt. V842 supplies +2325 volts for the post-anode acceleration. This provides an accelerating voltage of 4000 volts for the crt beam. V832 supplies about —1750 volts for the grid of the crt (the actual voltage depends upon the setting of the INTENSITY control, R826).

In order to maintain a constant deflection sensitivity in the crt, and thereby maintain the calibration of the oscilloscope, it is necessary that the accelerating potentials in the crt remain constant. This is accomplished by regulating the three supplies by comparing a "sample" of the high voltage to the regulated —150-volt supply. The "sample" voltage, obtained from the arm of the HV ADJ., R841, is applied to the grid of V814B; the cathode of this tube is connected to the regulated —150-volt supply. The error signal is amplified by V814B and V814A; the output of V814A varies the screen voltage of the oscillator tube, thereby controlling the amplitude of its output.

# Unblanking

As mentioned previously, dc-coupled unblanking is accomplished by employing separate high-voltage supplies for the grid and cathode. The cathode supply is tied to the +100-volt supply via the decoupling network, R801-C801A. The grid supply, on the other hand, is not tied to any other supply, and is therefore "floating". The unblanking pulses from the Time-Base Generator are transmitted to the grid of the crt via the floating grid supply.

The stray capacitance in the circuit makes it difficult to move the floating supply fast enough to unblank the crt in the required time. To overcome this, an isolation network composed of C827, R827, and R828 is employed. By this arrangement, the fast leading edge of the unblanking pulse is coupled directly to the grid of the crt via C827. For short-duration unblanking pulses (at the faster sweep rates), the power supply itself is not appreciably moved. For longer unblanking pulses (at the slower sweep rates), however, the stray capacitance of the circuit is charged through R827. This holds the grid at the unblanked potential for the duration of the unblanking pulse.

#### AMPLITUDE CALIBRATOR

The Amplitude Calibrator is a square-wave generator whose output is available at the CAL. OUT connector on the front panel. It consists of an astable multivibrator, V885A and V875, connected so as to switch the Cal. Out Cathode Follower, V885B, between two operating states, cutoff and conduction, at approximately 1 kHz.

During the negative portion of the multivibrator waveform, the grid of V885B is driven well below cutoff and the cathode rests at ground potential. The CAL ADJ. potentiometer, R879, is adjusted so that the voltage at the cathode of V885B (CAL TEST PT.) rises to exactly +100 volts during the positive portion of the multivibrator waveform. The tapped cathode resistor provides eleven calibrated outputs, from 0.05 volt to 100 volts, which are used for calibrating the gain of the Vertical Amplifier and adjusting the compensation of attenuator probes.

# SECTION 5



# MAINTENANCE

# **Preliminary Instructions**

#### PREVENTIVE MAINTENANCE

# **Visual Inspection**

Every few months, the oscilloscope should be visually inspected for possible circuit defects. These defects may include such things as loose or broken connections, damaged binding posts, improperly seated tubes, scorched wires or resistors, missing tube shields, and broken terminal strips. For most visual troubles the remedy is apparent; however, particular care must be taken when heat-damaged components are detected. Overheating of parts is often the result of other, less apparent, defects in the circuit. It is essential to determine the cause of overheating, before replacing heat-damaged parts, in order to prevent further damage.

#### Calibration

The Type 516 Oscilloscope is a stable instrument which should provide many hours of trouble-free operation. However, to insure the reliability of measurements we suggest that you calibrate the instrument after each 500 hours of operation (or every six months if used intermittently). A complete step-by-step procedure for calibrating the instrument is presented in Section 6 of this manual.

## Air Filter

The Type 516 Oscilloscope is cooled by forced, filtered air. This instrument is equipped with a washable air filter, constructed of aluminum wool coated with an adhesive. If the filter becomes dirty it may restrict the flow of air and cause the instrument to overheat. The filter should be inspected, and cleaned or replaced if necessary, every three to four months.

To remove the loose dirt, the filter may be rapped gently on a hard surface. It should then be rinsed briskly, from the dirty side, with hot water. Or, if preferred, it may be washed with hot, soapy water. After rinsing and drying, the filter should then be coated with "Handi-Coater" or "Filtercoat", products of the Research Products Corporation. These products are generally available from air-conditioner suppliers.

#### Fan Motor

The bearings in the fan motor are sintered bronze oilite bearings and do not require oiling.

#### REMOVAL AND REPLACEMENT OF PARTS

#### General Information

Procedures required for replacement of most parts in the Type 516 Oscilloscope are obvious. Detailed instructions for their removal are therefore not required. Other parts, however, can best be removed if a definite procedure is followed. Instructions for the removal of some of these parts are contained in the following paragraphs. Parts ordering information is included in the Parts List Section.

Because of the nature of the instrument, replacement of certain parts will require recalibration of sections of the oscilloscope to insure proper operation.

#### Removal of Side Panels

The side and bottom panels of the Type 516 Oscilloscope are held in place by small screwhead fasteners. To remove the panels, use a screwdriver to rotate the fasteners approximately two turns counterclockwise and pull the panels away from the instrument (see Fig. 5-1). In replacing the side panels, be sure that they are fitted properly at the bottom before fitting the top portion into place. The two side panels are perforated differently to promote proper air flow within the instrument. They should not be interchanged. Fig. 2-2 shows the two side panels correctly mounted.

## Replacement of the Cathode-Ray Tube

To remove the cathode-ray tube, first disconnect the tube socket and the five leads connected to the neck of the tube. Remove the graticule cover, spacer washers, graticule, and graticule light shield. Loosen the tube clamp at the base of



Fig. 5-1. Removing the escillascape side panels

the crt. Remove the crt by pushing forward on the tube base and then pulling the tube straight out through the front panel (see Fig. 5-2).

When the new crt is in place, the leads can be properly connected to the neck of the tube by following the color code information provided on the tube shield. After replacement of the crt, it may be necessary to calibrate certain portions of the oscilloscope. Special attention should be given to calibration of the sweep timing and the Vertical Amplifier gain.

#### Replacement of Switches

Methods for removal of defective switches are, for the most part, obvious and only a normal amount of care is required. Single wafers are normally not replaced on the switches used in the instrument. If one wafer is defective, the entire switch should be replaced. Switches can be ordered from Tektronix either wired or unwired, as desired

## **Tube Replacement**

Care should be taken both in preventive and corrective maintenance that tubes are not replaced unless they are actually causing trouble. Often during routine maintenance it will be necessary to remove tubes from their sockets. It is important that these tubes be returned to their original sockets unless they are actually defective. Unnecessary replacement or switching of tubes will often necessitate calibration of the instrument. If tubes do require replacement, it is recommended that they be replaced by previously checked, high-quality tubes.

#### **Soldering Precautions**

In the production of Tektronix instruments, a special silverbearing solder is used to establish a bond to the ceramic

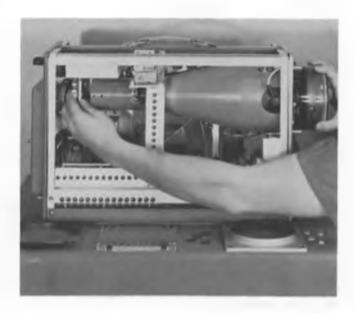


Fig. 5-2. Removing the cathode-ray tube

terminal strips. This bond can be broken by repeated use of ordinary tin-lead solder, or by the application of too much heat. However, occasional use of ordinary solder will not break the bond if too much heat is not applied.

It is advisable to have a stock of solder containing about 3% silver if you frequently perform work on Tektronix instruments. This type of solder is used frequently in printed circuitry and should be readily available. It may also be purchased directly from Tektronix in one-pound rolls (order by Tektronix Part Number 251-0514-00).

Because of the shape of the terminals on the ceramic terminal strips, it is advisable to use a wedge-shaped tip on your soldering iron when installing or removing parts from the strips. A wedge-shaped tip allows you to apply heat directly to the solder in the terminals and reduces the amount of heat required. It is important to use as little heat as possible.

#### **Ceramic Terminal Strips**

Damaged ceramic terminal strips are most easily removed by unsoldering all connections, then using a plastic or hard

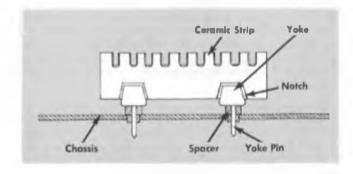


Fig. 5-3. Installation of ceramic terminal strips.

rubber mallet to knock the yokes out of the chassis. This can be done by pounding on the ends of the yoke protruding through the chassis. This removes both the strip and the yoke assembly.

When the damaged strip and yoke assembly have been removed, place the spacers for the new strip into the holes in the chassis. Snap the ceramic strip into the yokes and

place the tip of the yoke pins into the spacers. Using a plastic or hard rubber mallet, tap the ceramic strip lightly above the yokes to drive the yoke pins down through the spacers. Be certain that the yoke pins are driven completely through. Using a pair of diagonal cutters, cut off the excess length of the yoke pins protruding beyond the spacers. Fig. 5-3 illustrates the way that the parts fit together.

# **Troubleshooting**

The Troubleshooting section is divided into two parts, Circuit Isolation and Circuit Troubleshooting. When a trouble occurs in the instrument, refer first to the Circuit Isolation section for instructions on how to isolate the trouble to a given circuit. After determining which circuit is at fault, refer to the Circuit Troubleshooting section, where procedures for troubleshooting each circuit are given. Before attempting to troubleshoot the Type 516 Oscilloscope, however, make sure that any apparent trouble is actually due to a malfunction within the instrument and not due to improper control settings. Operating instructions are contained in Section 3 of this manual.

Apparent trouble may be due to improper calibration of one or more circuits. One of the first steps in any trouble-shooting procedure should be to check the calibration of the suspected circuit.

Power-supply output voltages should be checked whenever any type of trouble occurs in the instrument. Due to the circuit configurations employed in the Type 516, it is possible for an incorrect power-supply voltage to affect one circuit more than others. When all but one circuit in the oscilloscope is functioning properly, there is a tendency to overlook the Power Supply as a source of the trouble and to concentrate on the circuit where the trouble apparently exists. In cases of this type, valuable time may be saved by checking the power supplies first. If the output and ripple voltages of the regulated power supplies are correct, the power supplies can be assumed to be operating correctly.

When the trouble has been isolated to a definite circuit, perform a complete visual check on that circuit. Many troubles can be found easily by visual means. If the visual check fails to reveal the cause of the trouble, check the tubes by substitution. Do not depend on tube testers to adequately indicate the suitability of a tube for use in the instrument. The criterion for usability of a tube is whether or not it works satisfactorily in the instrument. Be sure to return good tubes to the same sockets from which they were removed.

Separate circuit diagrams for each circuit are contained in the Diagrams section of this manual. In addition, a block diagram provides an overall picture of instrument operation. The reference designation of each electronic component of the instrument is shown on the circuit diagrams. The following chart lists the reference designation numbers associated with each circuit.

All numbers less than 100	Time-Base Trigger
All 100 numbers	Time-Base Generator
All 200 and 300 numbers	Horizontal Amplifier
All 400 and 500 numbers	Vertical Amplifier
All 600 numbers	Power Supply
All 800 numbers	Crt Circuit and Calibrator

Switch wafers shown in the circuit diagrams are coded to indicate the position of the wafer on the switch. The number portion of the code refers to the wafer number on the switch assembly. Wafers are numbered from the front of the switch to the rear. The letters F and R indicate whether the front or the rear of the wafer is used to perform the particular switching function.

#### **CIRCUIT ISOLATION**

This portion of the Troubleshooting procedure lists some of the troubles that can be caused by a circuit failure in the Type 516 Oscilloscope. It also describes checks that can be made to isolate the faulty circuit or circuits. In some cases simple front panel checks can determine which circuit is defective, but in other cases internal checks and/or measurements are required.

The crt display can often be used to isolate the trouble to one particular circuit. If there is no vertical deflection, for example, when the intensity and horizontal deflection appear to be normal, an open condition probably exists in the Vertical Amplifier and this circuit should be investigated first

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Although the Type 516 Oscilloscope is a complex instrument, it can be thought of as consisting of six main circuits in addition to the Calibrator Circuit. These are:

- 1. Power Supply
- 2. Crt Circuit
- 3. Time-Base Generator
- 4. Time-Base Trigger
- 5. Vertical Amplifier
- 6. Horizontal Amplifier

Whenever any trouble occurs in the instrument, first check the voltages at the points shown in Fig. 6-4, at the -1675-and +2325-volt checkpoints at the top rear of the instrument, and at the center arm of the INTENSITY potentiometer, R826. If all of these voltages are approximately as indicated (the voltage at the center arm of the INTENSITY potentiometer should be about -1700 volts when the INTENSITY control is set fully clockwise), proceed to the troubleshooting procedures described under the appropriate symptom. If the low voltages (-150, +100, and +300) are not as indicated, refer to the paragraphs on Troubleshooting the Power Supply. If the low voltages are as indicated but the high voltages (-1675, +2325, and INTENSITY potentiometer voltage) are not, refer to the paragraphs on Troubleshooting the Crt Circuit.

#### WARNING

Be careful of the power-supply voltages. The low-voltage Power Supply can cause more harm than the high-voltage supply (in the CRT Circuit) due to the higher current capabilities of the circuit. When you reach into the instrument to measure the voltage, do not touch the metal frame. If possible, stand on an insulated surface and use insulated tools and measuring probes.

The following troubleshooting information is divided according to the symtoms presented to the operator. Upon detecting an apparent trouble, use the symptoms to determine which circuit is at fault. Then refer to the Circuit Troubleshooting information, where the procedure for troubleshooting within the circuit is given.

## No Trace on the Crt Screen

Normally, you should obtain two traces on the crt screen when the STABILITY control is set for a free-running sweep (fully clockwise), the MODE switch is set to the ALTERNATE position, the HORIZ. DISPLAY switch is set to NORM., and the POSITION controls are set to about midrange. If one of the two traces is missing under these conditions, there is an unbalance in the vertical input channel for which there is no trace. (The channel which controls the displayed trace can be determined by noting which VERTICAL POSITION control moves the trace up and down.) In this case, adjust the DC BAL. control of the faulty channel. If the trace is obtained, try to set the DC BAL. control according to the

procedure given in the Calibration section of this manual. If a trace is not obtained, or if it is not possible to set the control as described, turn to the instructions for trouble-shooting the Vertical Amplifier later in this section of the manual

If only two spots are present when the controls are set as described in the previous paragraph, and the spots can be moved with the HORIZONTAL POSITION control, then the trouble is in the Time-Base Generator. If the spots cannot be moved with the HORIZONTAL POSITION control, then the trouble is in the Horizontal Amplifier.

If no spot or trace at all is present on the face of the crt, the trouble may be in the Power Supply, the Vertical Amplifier, the Horizontal Amplifier, the Time-Base Generator, or the Crt Circuit. To determine which circuit is at fault in this case, set all POSITION controls to midrange, the HORIZ. DISPLAY switch to the EXT. position, and turn the INTENSITY control clockwise. If a spot now appears on the screen, either the Time-Base Generator or the first stage of the Horizontal Amplifier is at fault.

If a spot does not appear, short the vertical deflection plates together at the neck pins of the crt.

#### CAUTION

Be careful not to short the deflection plates to the metal shield around the crt, or to the GEOM. neck pin. Also be ready to turn the INTENSITY control down quickly when a spot does appear so as not to burn the crt screen.

If a spot now appears on the screen, either the Vertical Amplifier or the Delay Line is at fault. If no spot appears, remove the shorting strap from the vertical deflection plate pins and short the horizontal deflection plates together. If a spot now appears, the trouble is in the Horizontal Amplifier.

If no spot or trace appears during any of the previous checks, a defect in the Crt Circuit, possibly in the crt itself, is indicated.

# Insufficient or No Vertical Deflection, or Waveform Distortion

These troubles are all caused by a defective Vertical Amplifier and/or Delay Line. Refer to the instructions for troubleshooting these circuits. If there is reduced deflection sensitivity both vertically and horizontally, the trouble is probably in the Crt Circuit.

# **Insufficient Horizontal Deflection**

This condition can be produced by the Time-Base Generator or the Horizontal Amplifier.

The operation of the Time-Base Generator can be checked from the front panel. Set the HORIZ. DISPLAY switch to the NORM. position, the TIME/DIV. switch to .5 SEC, and adjust the STABILITY control for a free-running sweep (fully clock-

wise). Connect a voltmeter between the SAWTOOTH OUT connector and ground. If the voltage varies between zero and  $\pm 150$  volts  $\pm 15\%$ , as the Miller circuit runs up and back, the Time-Base Generator is operating correctly. This means that the trouble is in the Horizontal Amplifier. No voltage variation at this point, or a variation of significantly less than 150 volts, indicates trouble in the Time-Base Generator circuit.

# **Nonlinear Horizontal Sweep**

The linearity of the horizontal deflection circuit can be checked by connecting a marker generator or the Calibrator output to the Vertical input of the oscilloscope. Set the MODE switch to correspond to the Vertical Channel being used, and adjust the Time-Base controls for a stable display. If the displayed markers, or the square waves produced by the Calibrator, are not symmetrically spaced across the crt, a nonlinear sweep is indicated. This condition can be caused by nonlinear amplification in the Horizontal Amplifier, or by the generation of a nonlinear sawtooth in the Time-Base Generator.

To determine which circuit is at fault, connect a jumper wire from one of the Vertical input connectors to the SAW-TOOTH OUT connector.

#### CAUTION

To avoid shorting the 150-volt sawtooth to the chassis, connect the jumper wire to the input connector before connecting it to the SAWTOOTH OUT connector.

Set the appropriate VOLTS/DIV. switch to 20, and adjust the STABILITY control for a free running sweep (fully clockwise). If the TIME/DIV. control is set to a rate of 1 mSEC or faster, a steady diagonal trace will be seen on the crt. This trace is the sweep portion of the sawtooth voltage produced by the Time-Base Generator. If the slope of the trace is constant, the nonlinearity is being produced in the Time-Base Generator. If the slope of the trace is not constant, the nonlinearity is being produced in the Horizontal Amplifier. Refer to the instructions for troubleshooting the Horizontal Amplifier.

#### **Improper Sweep Timing**

If the sweep timing is off in some, but not all, positions of the TIME/DIV. control, one of the timing resistors or timing capacitors has changed in value. By comparing the switch positions in which the timing is incorrect with the Timing Switch diagram, you will be able to tell which components are common to these positions.

If the timing is off in all positions of the TIME/DIV. control, the Horizontal Amplifier is probably the circuit at fault. However, it is important that the Power Supply voltages be checked first. Check to see if the timing circuits can be calibrated in accordance with the instructions presented in the Calibration section of this manual. If the circuits cannot be adjusted for correct timing, then refer to the instructions for troubleshooting the Horizontal Amplifier.

#### Improper Triggering

The most probable cause of poor triggering (as compared with no triggering at all) is lack of calibration of the Time-Base Trigger. The first thing to check in the event of poor or erratic triggering, then, is the calibration of the Time-Base Trigger. Procedures for calibration are given in Section 6

If the oscilloscope will not trigger at all, and yet a freerunning trace can be produced by turning the STABILITY control fully clockwise, trouble exists in the Trigger Pickoff circuit of the Vertical Amplifier or in the Time-Base Trigger itself. If only the internal triggering is absent, the trouble is in the Trigger-Pickoff circuit or in SW10A in the Time-Base Trigger. If all triggering (internal, external, and line) is absent, then the trouble is in the Time-Base Trigger.

If stable triggering cannot be obtained and the free-running trace cannot be turned off with the STABILITY control, the trouble lies in the Sweep-Gating Multivibrator of the Time-Base Generator.

#### CIRCUIT TROUBLESHOOTING

This portion of the Troubleshooting procedure contains information for locating a defective stage within a given circuit. Once the stage at fault is known, the component or components causing the trouble can be located by tube and component substitution or by voltage and resistance measurements.

Tube failure is the most prevalent cause of circuit failure. For this reason, the first step in troubleshooting any circuit is to check for defective tubes, preferably by direct substitution.

If replacement of a defective tube does not correct the trouble, then check to see whether components, through which the tube draws current, have been damaged. Shorted tubes will often overload plate-load and cathode resistors. These components can often be located by a visual inspection of the circuit. If damaged components are not apparent, it will be necessary to make measurements or other checks within the circuit to locate the trouble.

## **Troubleshooting the Power Supply**

Proper operation of every circuit in the Type 516 Oscilloscope depends upon proper voltages from the Power Supply. The voltages must remain within their specified tolerances for the instrument to maintain its calibration.

The regulated supply voltage busses are identified by color-coded wires, following the standard color code. The -150-volt bus is coded brown-green-brown; the +100-volt bus is coded brown-black-brown; and the +300-volt bus orange-black-brown. The widest stripe always identifies the first color in the code.

No Output Voltage. If there are no output voltages from the Power Supply, note whether the graticule lamps are lighted. If they are, then the trouble is in the Power Supply. If the graticule lamps are not lighted and the fan is not going, check the power cord connections, the fuse, and the source of power. If your instrument is wired for 220-, 234-, or 248-volt operation, also check the Thermal Cut-Out Switch by checking the continuity through it with an ohmmeter. If your instrument is wired for 110-, 117-, or 125-volt operation and the fan is running but the graticule lamps are not lighted, check the Thermal Cut-Out Switch. If none of the above checks reveals the trouble, then the primary of T601 is probably open.

Failure to Regulate at the Correct Voltage. If any of the supplies fail to regulate, the first thing to check is the line voltage. The supplies are designed to regulate at line voltages between 105 and 125 volts at a nominal line voltage of 117 volts and within proportionate limits at the other nominal line voltage for which the instrument can be wired. Improper line voltages may cause the supply voltages to be off.

If the line voltage is the correct value, the next step is to turn the instrument off and measure the resistance between the Power Supply test points, shown in Fig. 6-4, and ground. The —150-volt test point should measure approximately 4500 ohms to ground, the +100-volt test point should measure approximately 4000 ohms to ground, and the +300-volt test point should measure approximately 12,000 ohms to ground.

If these values check out, the next step is to check the tubes. Then check the rms voltage across each secondary winding of the transformer. The voltage between terminals 8 and 9 should be about 195 volts ac; between terminals 15 and 16, about 125 volts ac; and between terminals 5 and 6, about 180 volts, ac. If these voltages are correct, check the rectified voltage at the input to each regulator. The voltage at the plate of V667 should be about +410 volts; at the plate of V667A, about +175 volts; and at the plate of V627, about +72 volts. If these voltages are all correct, check for open or leaky capacitors and improper resistance values, especially in the dividers. If it becomes necessary to adjust the -150 ADJ, potentiometer, it will be necessary to recalibrate the instrument.

The material that follows may be used as a quick index to troubleshooting the low voltage Power Supply.

If the output voltage is high with excessive ripple, check:

- 1. For high line voltage.
- 2. The amplifier tubes (V634, V654, V674).
- 3. For insufficient loading.

(Proper loading of the supplies can be checked by measuring the voltage drop across the 10-ohm resistors, R640, R642, and R644, in each supply. With the AMPLITUDE CALIBRATOR control set to OFF and the STABILITY control set so that no sweep is being produced (counterclockwise but not to PRESET), the drop across the resistors should be as follows: R640, 1.65 volts  $\pm 0.1$  volt; R642, 1.4 volts  $\pm 0.1$  volt: R644, 1.75 volts  $\pm 0.1$  volt.)

If the output voltage is high with normal ripple, check for proper resistance values in the dividers (R615, R616, and R617; R650 and R651; R670 and R671).

If the output voltage is low with excessive ripple, check:

- 1. For low line voltage.
- 2. The series tubes (V667, V627).
- 3. For excessive loading.
- 4. Open or leaky filter capacitors.
- 5. Bad diode rectifiers.

If the output voltage is low with normal ripple, check:

- 1. The resistance values in the dividers.
- 2. The capacitors across the dividers.

# **Troubleshooting the CRT Circuit**

If the trouble has been isolated to the Crt Circuit, first check the voltages at the —1675-volt test point and the +2325-volt test point at the top rear of the oscilloscope. If there is no voltage at either of these points, remove the high-voltage shield containing the two test point openings.

#### WARNING

Always turn the instrument off before removing or replacing either high-voltage shield.

Check for heater glow in the three high-voltage rectifier tubes, V822, V832, and V842. If there is no heater glow in any of these tubes, the high-voltage oscillator circuit is not oscillating. If there is heater glow in any or all of the three rectifier tubes, measure the voltage at the grid of the Oscillator tube, V800. If this voltage is about —55 volts, the oscillator is operating properly. If it is significantly less than —55 volts, the oscillator is not operating properly.

If the oscillator is not operating properly, and replacement of V814 and V800 does not correct the trouble, check the other components associated with the circuit, including the primary and secondary of T801.

If there is no voltage at either of the high-voltage test points, but the high-voltage Oscillator is found to be operating properly, then the trouble lies in the secondary of T801 or in one of the high voltage rectifiers. If this is the case, measure the voltage at the plate of V832. (This is the high-voltage rectifier tube nearest the centerline of the instrument; the plate is the connection at the forward end of the tube.) The voltage at this point should be in the vicinity of -1775 volts. If it is not, then trouble exists in some part of the circuit common to all three high-voltage supplies. If the voltage at the plate of V832 is in the vicinity of -1775 volts, then trouble exists in some part of the circuit peculiar to the -1675- and +2325-volt supplies.

If there are voltages present at the -1675-volt test point and the +2325-volt test point, but they are not of the proper value and cannot be brought to the proper value by adjustment of the HV ADJ. (R841), the regulator circuit (V814) is probably faulty, or the supply is being loaded down by a short somewhere in the circuit.

If the high-voltage supplies are normal but no spot is visible on the crt, check the continuity of the cathode circuit. Also, check the ASTIGMATISM control, R864, for proper re-

sistance and voltage. If a badly distorted spot or trace is visible on the crt, check the GEOMETRY ADJ. adjustment, R861, and its connection to the neck pin on the crt.

If the Crt Circuit checks out satisfactorily, but trouble still exists, replace the crt.

# **Troubleshooting the Vertical Amplifier**

No Spot or Trace. If a trace or spot is visible when the vertical deflection plates are externally shorted together, but disappears when the short is removed, the vertical deflection system is in a state of dc unbalance. This could be caused by an open inductor in one side of the Delay Line, by an open peaking coil in the Vertical Amplifier, or by a faulty tube. To check for this condition, short the plates of the Output Amplifier together (V464 pin 6 and V474 pin 6). If the trace does not appear, one side of the circuit, between the Output Amplifier and the crt, is open. A continuity check with an ohmmeter is perhaps the best way to determine which side is open.

If the trace does appear, however, when the plates of the Output Amplifier stage are shorted together, the circuit is all right between this point and the crt. This means that the trouble lies somewhere ahead of the plate circuit of the Output Amplifier. Short the grids of the Driver Cathode Follower stage together (pins 2 and 7 of V463). If a trace does not appear, the trouble is in the Driver Cathode Follower stage or in the grid circuit of the Output Amplifier. Check tubes, peaking coils, resistors, and capacitors in the circuit.

If a trace does appear when the grids of the Cathode Follower are shorted together, check L455, L555, R455, and R555 in the grid circuit of the Driver Cathode Follower.

If a spot or trace appears on one vertical channel and not on the other, a dc unbalance condition exists in the defective channel. In this case, set the MODE switch to correspond to the defective vertical channel and use a shorting strap as follows to locate the defective stage. Short between corresponding points on opposite sides of the circuit, starting at the plates of the Switched Amplifier, V454 or V554, and working back toward the input connector. When a point is reached where the trace no longer appears as opposite sides of the circuit are shorted together, the stage immediately following this point is the stage in which the unbalance is being produced. (When the grids of the Input Amplifier or the cathodes of the Input Cathode Follower are shorted to-

gether, the VERTICAL POSITION control may have to be adjusted to bring the trace into view.) The trouble may be a defective tube, a shorted capacitor, a defective resistor, or a broken lead.

Insufficient or No Vertical Deflection. Insufficient vertical deflection indicates a change in the gain of the Vertical Amplifier. If the change is slight, the Vertical Amplifier can usually be recalibrated to allow for it. Refer to Section 6 of this manual for this procedure.

If the change in gain is more pronounced, or if there is no vertical deflection at all, check the tubes first. Then check components which can affect the gain but not the dc balance of the circuit; for example, the VARIABLE controls, the GAIN ADJ. potentiometer, plate-dropping resistors R422, R522, R432, R457, and R489, and the screen resistor, R468.

If gain trouble exists in only one vertical channel, then one of the first two stages (Input Cathode Follower or Input Amplifier) is probably at fault. If the gain trouble exists in both vertical channels then a Driver Cathode Follower and/or the Output Amplifier is probably at fault.

**Waveform Distortion.** Waveform distortion can be divided into two categories—low frequency and high frequency. High-frequency distortion normally appears as "overshoot" or "undershoot" at the leading edge of a high-frequency square wave, while low-frequency distortion is normally characterized by a "tilt" to the top and bottom of a low-frequency square wave.

The first thing to check, if waveform distortion is detected, is the adjustment of the probe. The procedure for adjusting a probe is described in Section 3 (see Fig. 3-2).

Faulty tubes are the most common cause of low-frequency distortion. Even though the vertical deflection system is decoupled from the attenuators to the crt, if certain tubes (depending on their grid circuits) start drawing grid current, a time-constant network can be established which will affect the low-frequency response of the circuit. If low-frequency distortion is observed, change the tubes in the suspected circuits

Three types of high-frequency distortion are illustrated in Fig. 5-4. Insufficient high frequency peaking, which limits the risetime and consequently the bandwidth, will produce the "undershoot" type of distortion shown in Fig. 5-4 (a). This condition can be caused by tubes, particularly when

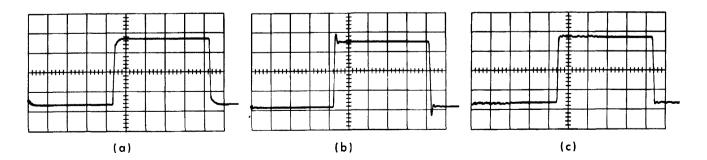


Fig. 5-4. Three types of high-frequency distortion. A 450-kHz square wave with a risetime of less than 0.020 microsecond was used for these pictures.

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used as cathode followers working into peaking circuits. Shorted, or partially shorted, peaking coils are another common source; hot solder spilled on a peaking coil can cause this condition.

An "overshoot" such as shown acccentuated in Fig. 5-4 (b), is the result of excessive high-frequency peaking. This can be caused by a type of tube distortion known as cathode interface. The first thing to check, therefore, for this type of distortion is the tubes. If tube replacement does not correct the trouble, the variable peaking coils in the Vertical Amplifier may need adjusting. Refer to the Calibration section of this manual for the adjustment procedure.

The "wrinkle" condition, illustrated in Fig. 5-4 (c), is produced by an improperly adjusted Delay Line. Refer to the Delay Line Adjustment procedure in the Calibration section.

**Improper Triggering.** If the Time-Base Generator triggers normally in the EXT. and LINE position (+ or —), but does not trigger properly in the INT. position, a defect in the Trigger Pick-Off Circuit is indicated. In this case, the trouble would be either a defective Trigger Cathode Follower, V493, or a defective resistor in the circuit.

# **Troubleshooting the Time-Base Trigger**

If the sweep cannot be triggered, and if the Trigger-Pickoff circuit in the Vertical Amplifier has been eliminated as the source of trouble, the Time-Base Trigger is at fault. (This assumes that the trace can be turned on and off with the STABILITY control.) The first thing to do is to replace both tubes in the Time-Base Trigger and try to recalibrate the circuit according to the procedure given in the Calibration section of this manual. If this does not correct the trouble, then trouble in the circuitry is indicated.

To check the quiescent state of the circuit, set the red TRIGGER SELECTOR control to AC, the black TRIGGER SELECTOR control to —INT., and the TRIGGERING LEVEL control to 0. Next, connect a jumper wire between the junction of R19, R20, and R21 and ground (see Fig. 6-6). This fixes the voltage at the grid of V24B at ground potential. Then measure the voltage at the plate of V24B (pin 1); it should be about +80 volts. If this voltage does not measure close to +80 volts, replace tube V24 again. Then, if necessary, check for such things as off-value resistors, broken leads, and poor switch contacts.

The next step is to measure the voltage difference between the two grids of V45 (pins 2 and 7). Since connecting a voltmeter at the grid of V45A (pin 2) may produce an adverse loading effect, it is recommended that this measurement be made between the plate (pin 1) of V24B and the grid (pin 7) of V45B. With the TRIG. LEVEL CENTERING adjustment, you should be able to bring the voltage at the grid of V45B to within 4 or 5 volts of the voltage at the plate of V24B. This indicates that the hysteresis of the Trigger Multivibrator can be set at the proper level with respect to the grid of V45A.

If the voltage at the grid of V45B cannot be brought to within 4 or 5 volts of the voltage at the plate of V24B with the TRIG. LEVEL CENTERING adjustment, trouble in the Trigger Multivibrator is indicated. In this case, replace V45 again; then, if necessary, check the rest of the circuit for off-value resistors or other troubles.

Remove the jumper wire between the junction of R19, R20, and R21 and ground.

# **Troubleshooting the Time-Base Generator**

**No Horizontal Sweep.** If the Time-Base Generator is not producing a sawtooth waveform when the STABILITY control is set fully clockwise, some defect in the circuit is causing the output to remain at some fixed voltage. A clue to the cause of this trouble can be obtained by measuring the plate voltage of the Miller tube, V161A pin 6.

#### NOTE

All voltages in this section should be measured with a 20,000-ohms-per-volt voltmeter or a vacuum-tube voltmeter.

The voltage reading obtained at the plate of V161A will probably be approximately +260 volts, or approximately +30 volts. A reading of +260 volts indicates that the Miller Runup Circuit has run up and has not been reset, while a reading of +30 volts indicates that the Miller Runup Circuit is not being allowed to run up. The condition that exists will depend on the type of trouble. The two conditions of plate voltage will be handled separately in the following paragraphs.

High voltage at the plate of the Miller tube, V161A, indicates that the tube is cut off. If this is the case, momentarily ground the grid of the tube while monitoring the plate voltage. If the tube is good, the plate voltage will drop to about +10 volts. Remove the ground as soon as the reading is taken. If the Miller tube is found to be good, measure the voltage at its grid. If this voltage is more than 20 volts negative, V152B is probably not conducting. In this case, check V152, R147, and R148.

If the voltage at the grid of V161A is not more negative than -20 volts (it should be about -5 volts), measure the voltage at the cathode of V161B. If this voltage is approximately +200 volts, the Runup Cathode Follower stage may be assumed to be operating correctly. If this voltage is significantly lower than +200 volts, the stage is defective, and its grid and cathode circuits should be checked.

If the Runup Cathode Follower is found to be operating properly, measure the voltage at the cathode of V183A. If this voltage is more positive than —45 volts, the trouble is in the Sweep-Gating Multivibrator. Check the tubes and resistors in this circuit. The voltage divider network in the cathode circuit of V135B is particularly critical.

If the voltage at the cathode of V183A is more negative than —55 volts, check the tube in the Hold-Off Circuit, the Hold-Off capacitors, and the resistors in the cathode circuits of the tube.

Low voltage at the plate of the Miller tube indicates that the tube is conducting quite heavily and is not being allowed to perform its normal run-up operation. If this trouble exists on only a few ranges of the TIME/DIV. control, the trouble is probably an open timing resistor. If the trouble exists on all ranges of the TIME/DIV. control, the trouble is probably due to a defective Sweep-Gating Multivibrator.

Check the voltage at the grid (pin 2) of V135A. If the voltage at this point is in the vicinity of —65 volts or lower (more negative), the Sweep-Gating Multivibrator is faulty. In this case, replace V135 and V145. If this does not remedy the trouble, check the rest of the circuit by voltage and resistance measurements.

If the voltage at the grid of V135A is more positive than —60 volts, measure the voltage at the grid (pin 2) of V183B. If the voltage at this point is —70 volts or lower (more negative), the Hold-Off Circuit is faulty. In this case, replace V183, and if this does not remedy the trouble, check the rest of the circuit by voltage and resistance measurements. (The cathode of V183B and the grid of V183A should be about 5 to 7 volts more positive than the grid of V183B. However, there should be a considerably greater voltage differential between the grid and cathode of V183A.) If the voltage at the grid of V183B is more positive than about —70 volts, the Runup Cathode Follower circuit is faulty. If this case, replace V161, and if this does not remedy the trouble, check the rest of the circuit by voltage and resistance measurements.

**Nonlinear Sweep.** A nonlinear sweep will be generated if the current charging the Timing Capacitor does not remain constant. If the nonlinearity occurs at all sweep rates, a defective Miller Tube is probably the cause. If the nonlinearity occurs only at certain sweep rates, a leaky timing Capacitor is probably the cause. A defective C165 can also cause the sweep to be nonlinear at the faster sweep rates.

Constant Free-Running Trace. If the free-running trace cannot be turned off with the STABILITY control, the Sweep-Gating Multivibrator is at fault. The most probable cause is a change in resistance in either of the grid circuits or in the cathode circuit.

Insufficient Horizontal Deflection. If the horizontal trace starts at the left-hand side of the screen, but does not extend to the right-hand side, the Hold-Off Circuit is resetting the Sweep-Gating Multivibrator before the sweep is complete. If the sweep cannot be adjusted to normal length with the SWP. LENGTH adjustment, R176, the resistances in the cathode circuit of V161B should be checked.

# **Troubleshooting the Horizontal Amplifier**

**No Spot or Trace.** If a spot is visible when the horizontal deflection plates are externally shorted together, but disappears when the short is removed, the horizontal deflection system is in a state of unbalance. The procedure for locating the defective stage in the Horizontal Amplifier is somewhat the same as that described for the Vertical Amplifier. That is, the shorting strap should be removed from the deflection plates, and connected between the grids of the Output Cathode Follower, V364B and V384B. If no spot then appears, the Output Cathode Follower stage is at fault.

If the spot does appear, however, the shorting strap should then be moved back to the grids of the Output Amplifier, V364A and V384A. No spot on the crt when these points are shorted together indicates trouble in the Output Amplifier stage. Check for open resistors in the cathode and plate circuits.

If the spot appears when the grids of the Output Amplifier are shorted together, the trouble lies ahead of the Output Amplifier.

**Insufficient or no Horizontal Deflection.** If the gain of the Horizontal Amplifier decreases, the timing will no longer correspond to the calibrated values indicated by the TIME/DIV. control.

If the change in gain is only slight, as indicated by improper timing and a slightly decreased horizontal sweep, the amplifier can usually be recalibrated. However, since the gain of the Horizontal Amplifier regulates the timing of the sweep, care must be taken to insure that the gain adjustments are accurately made. Refer to the Calibration section if it is necessary to adjust the gain of the Horizontal Amplifier.

If the decrease in gain of the Horizontal Amplifier is more pronounced, or if there is no horizontal deflection at all, check for defective components that can affect the gain but not the dc balance of the circuits. Such components, in addition to the tubes, would be the R225, (SWP. GAIN ADJ.), R347, R377, and R259 (MAG. GAIN ADJ.).

# **NOTES**

Type 516

# INTRODUCTION

We recommend that the Type 516 Oscilloscope be calibrated after each 500 hours of operation or every six months, whichever comes sooner. It should not require more frequent calibration. However, whenever tubes or other circuit components are replaced, the calibration of the circuit involved should be checked and, if necessary, readjusted.

Also, apparent troubles in the instrument can be caused by improper calibration of one or more circuits. Consequently, this section of the manual should be used in conjunction with the Maintenance section during troubleshooting work. If a trouble occurs in the instrument, you should be sure that it is not due to improper calibration before proceeding with more detailed troubleshooting.

In the instructions that follow, the steps are arranged in the proper sequence for a complete calibration of the instrument. However, any single step may be performed individually or out of order as long as the entire step is performed, including references to other steps or adjustments, with certain exceptions. Due to interaction between adjustments in the Horizontal and Vertical Amplifiers and the sweep timing circuits of the Time-Base Generator, single adjustments in these circuits cannot be made. When amplifier adjustments are rerequired, the entire amplifier should be recalibrated. In addition, if the -150-volt Power Supply is adjusted, the entire instrument should be recalibrated. Front-panel controls not mentioned in a given step are assumed to be set at the positions they were in at the end of the previous step.

Figs. 6-1, 6-2, and 6-3 show the location of the internal adjustments referred to in the Calibration procedures.

# **EQUIPMENT REQUIRED**

The following equipment is necessary for a complete calibration of the Type 516 Oscilloscope:

- 1. DC voltmeter (sensitivity of at least 5000  $\Omega$ /volt calibrated for an accuracy of 1% or better at 100 volts, 150 volts, and 300 volts, and for an accuracy of 3% or better at 1675 volts.
- 2. Accurate rms-reading ac voltmeter, having a range of at least 0-125 volts (0-250 volts for 234-volt instruments).
- 3. Variable autotransformer having a rating of at least 3 amperes at 125 volts (or 1.5 amperes at 250 volts for 234volt instruments).
- 4. Time-Mark Generator, Tektronix Type 184 or equivalent. Time-Mark Generator used must have markers at 1 usec,

# SECTION

# CALIBRATION

 $10 \, \mu \text{sec}$ ,  $100 \, \mu \text{sec}$ , 1 msec, 5 msec, 10 msec, 100 msec, 1 sec, and a sine-wave output of 20 megahertz; accuracy of at least 1%.

- 5. Square-Wave Generator, Tektronix Type 106 or equivalent. Required specifications are: (1) Output frequencies of 1 kHz and 450 kHz, (2) risetime of 15 nanoseconds or less, and (3) output amplitude variable from approximately 40 millivolts to 100 volts.
- 6. Constant-Amplitude Sigal Generator, Tektronix Type 191 or equivalent. Required specifications are: (1) output amplitude 200 millivolts and equal at all frequency settings, and (2) output frequency variable from 1 MHz to over 15 MHz.
- 7. 52-ohm termination resistor, Tektronix Type B52-R or equivalent.
- 8. 52-ohm 10-1 "L" pad, Tektronix Type B52-L10 or equivalent.
- 9. Coaxial cable suitable for applying the outputs of the square-wave generator and the time-mark generator to the inputs of the Type 516, Tektronix Type P52 Coaxial Cable or equivalent.
- 10. 20 pF Input Capacitance Normalizer, Tektronix Calibration Fixture 067-0538-00.
- 11. Low-capacitance calibration tools: Tektror x Part Nu nbers 003-0000-00, 003-0301-00 and 003-0007-60 or equivalent.

# CALIBRATION PROCEDURE

# **Preliminary**

Remove the side covers from the Type 516 Oscilloscope. Set the front panel controls as follows (controls not listed may be left in any position):

TIME BASE

TRIGGER SELECTOR

+INT.

(black)

TRIGGER SELECTOR

AC

0

(red)

STABILITY fully counterclockwise but not PRESET

TRIGGERING LEVEL

HORIZ DISPLAY

NORM.

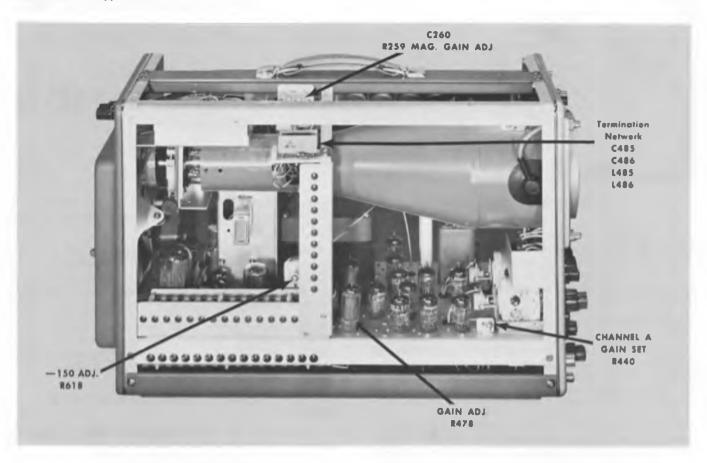


Fig. 6-1, Left zide of Type 516 Oscilloscope, showing location of internal adjustments

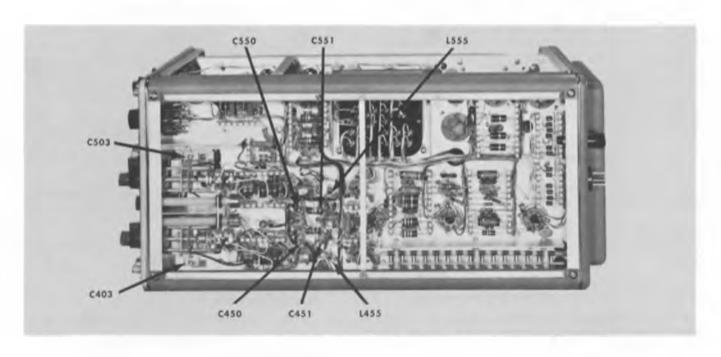


Fig. 6-2 Bottom of Type 516 Oscilloscope, showing location of internal adjustments.

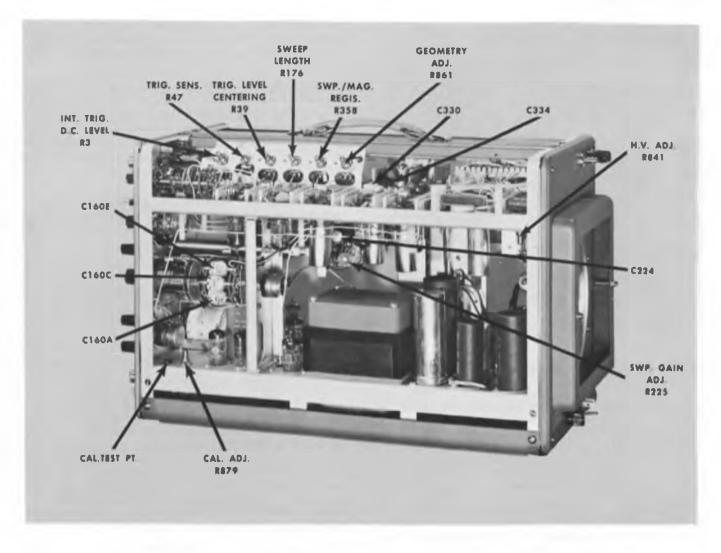


Fig. 6-3. Right side of Type 516 Oscilloscope, showing location of internal adjustments.

1 mSEC	TIME/DIV.
CALIBRATED	VARIABLE TIME/DIV.
counterclockwise	INTENSITY
PWR. OFF	POWER
	A VERTICAL
1	VOLTS/DIV.
CALIBRATED	VARIABLE VOLTS/DIV.
NORM. AC	POLARITY
centered	POSITION
	B VERTICAL
1	VOLTS/DIV.
CALIBRATED	VARIABLE VOLTS/DIV.
NORM. AC	POLARITY
centered	POSITION

MODE	A ONLY
HORIZONTAL POSITION	centered
AMPLITUDE CALIBRATOR	OFF

Connect the Type 516 Oscilloscope to the autotransformer, and turn on all equipment. Set the output of the autotransformer to the nominal voltage for which your instrument is wired. Allow the instrument to warm up for at least 5 minutes before proceeding.

# **Power Supply**

Proper operation of every circuit in the Type 516 Oscilloscope is dependent on correct power supply voltages. Since the regulated —150-volt supply serves as the reference for the other regulated supplies it is very important that this supply be properly adjusted.

1. -150 ADJ. (R616). Measure the output voltage of the -150-volt, the +100-volt and the +300-volt supplies at the

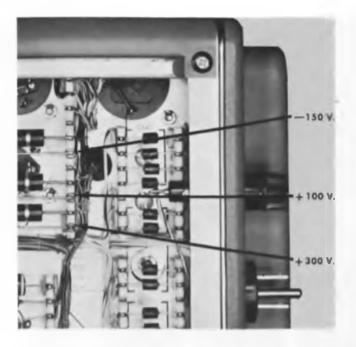


Fig. 6-4. Location of Power Supply test points

points indicated in Fig. 6-4. Set the -150 ADJ, control so that the -150 volt supply is within 2% and the other supplies within 3% of their rated values. Do not adjust the -150 ADJ, control unless one or more of the supplies is actually out of tolerance or unless you are planning to perform a complete recalibration of the instrument. The calibration of most of the circuits of the instrument will change if this control is adjusted.

2. Regulation. To check the operation of the regulator circuits, vary the output voltage from the autotransformer between the limits specified in Section 2 of this manual, at the same time observing the required voltages. All of the voltages should remain essentially constant and within specified tolerances over this range of line voltage.

## **Amplitude Calibrator**

3. CAL. ADJ. R879. When the CAL ADJ. control is properly set, the calibrator output will be within 3% of the voltages indicated on the front panel. To make this adjustment, connect a voltmeter between the CAL TEST PT. jack and ground, turn the AMPLITUDE CALIBRATOR control to OFF, and adjust the CAL ADJ. control for a meter reading of exactly 100 volts. To assure suitable symmetry of the calibrator waveform, the voltage at this point should fall to between 45 and 55 volts when the calibrator is turned on (to any of the output voltage settings). Readings outside of this range are generally caused by unbalanced multivibrator tubes V875 or V885.

## **CRT Circuit**

4. HV ADJ. R841. The adjustment that sets the high voltage determines the total accelerating voltage on the crt and

thus affects the deflection sensitivity. Do not adjust the high voltage unless it is actually of tolerance or unless you are planning to perform a complete recalibration of the instrument.

Measure the voltage at the -1675 ADJ. TEST PT. at the top rear of the instrument. Set the HV ADJ. adjustment for a reading of -1675 volts. This voltage should not vary more than 10 volts between the following limits: line voltage set at its lower limit (see Section 2) and the INTENSITY control set fully clockwise; line voltage set at its upper limit and the INTENSITY control set fully counterclockwise.

5. Crt Alignment. Check to see that the face of the crt rests snugly against the graticule. If it does not, loosen the crt clamp and move the tube forward by pushing on the crt tube socket. Then tighten the crt clamp.

Set the STABILITY control clockwise to free run the sweep. Turn up the INTENSITY control until a trace is visible (it may be necessary to adjust the VERTICAL POSITION control) and adjust the FOCUS and ASTIGMATISM controls for the narrowest trace width. With the VERTICAL POSITION control, position the trace directly behind the center horizontal graticule line. If the trace is tipped relative to the graticule line, rotate the crt alignment knob until the trace coincides with the graticule line.

6. Graticule Alignment. To check the alignment of the graticule, obtain a free-running trace on the oscilloscope, as explained in the previous step. Next move the trace, with the VERTICAL POSITION control, to the top of the graticule until the trace disappears. Then move the trace to the bottom of the graticule until the trace disappears. If the graticule lines are not centered in the usable viewing area, the graticule is improperly aligned in the vertical plane. The graticule may be repositioned by means of a nylon adjusting cam, located in the lower left corner of the graticule.

To make this adjustment, remove the graticule cover and loosen the set screw that holds the positioning cam. By inserting a pointed tool (such as a scriber or center punch) into the small hole, the cam may be rotated until the graticule lines are centered in the usable viewing area. Then tighten the set screw that holds the nylon cam, and replace the graticule cover.

7. Crt GEOMETRY ADJ. R861. The geometry of the crt display is adjustable over a limited range by means of the GEOMETRY ADJ. potentiometer. To achieve optimum linearity, vertical lines are displayed on the crt and the GEOMETRY potentiometer is adjusted for minimum curvature in the lines. Nonlinearity is most noticeable at the edges of the graticule.

Connect a time-mark generator to the A VERTICAL input connector and adjust the TIME/DIV. control and the A VERTICAL VOLTS/DIV. control to obtain vertical lines as illustrated in Fig. 6-5. Adjust the oscilloscope controls to obtain a stable display. Adjust the GEOMETRY ADJ. potentiometer for straight vertical lines running parallel to the left and right edges of the graticule.

The calibrator output waveform can be used in place of the time-mark generator to make this adjustment, but due to the dimness of the trace, the adjustment is more difficult.

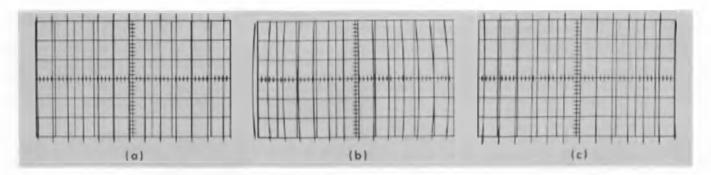


Fig. 6-5. Chacking the geometry of the crt. When the GEOMETRY ADJ, is properly set, vertical lines will be parallel with the graticule lines, as shown in (a). If GEOMETRY ADJ, is not properly set, lines will appear as shown in (b) or (c).

# **Triggering Circuits**

Set the front panel controls as follows:

TIME BASE

MODE

AMPLITUDE CALIBRATOR

	INTE BASE
+INT,	TRIGGER SELECTOR (black)
AC	TRIGGER SELECTOR (red)
Fully clockwise	STABILITY
0	TRIGGERING LEVEL
NORM.	HORIZ. DISPLAY
.5 mSEC	TIME/DIV.
	VERTICAL
1	VOLTS/DIV.
CALIBRATED	VARIABLE VOLTS/DIV.
NORM. DC	POLARITY

Connect a test lead from the CAL. OUT connector to the A input connector; this should result in a free-running trace having an amplitude of one minor graticule division (one-fifth of major division). Center the trace vertically on the screen and adjust the INTENSITY, FOCUS and ASTIGMATIM controls for best definition. Then ground the junction of R19, R20, and R21 with a short clip lead. This junction is located on the Trigger switch (see Fig. 6-6).

A ONLY

.2

8. TRIG. LEVEL CENTERING R39. Set the TRIG. SENS. control fully counterclockwise and the TRIG. LEVEL CENTERING control fully clockwise. Turn the STABILITY control counterclockwise until the trace just disappears from the crt screen, then two or three degrees further counterclockwise.

Rotate the TRIG. LEVEL CENTERING control slowly counterclockwise until the trace appears (it may just flicker), then back off the control approximately 2 degrees clockwise from that point.

Whenever you adjust the TRIG. LEVEL CENTERING adjustment you should also adjust the TRIG. SENS. as described in the following step.

9. TRIG. SENS. R47. (Always perform step 8 before this step). Leave the jumper wire and calibrator signal connected as in the previous step. Rotate the TRIG. SENS. control slowly clockwise until the sweep triggers. Rotate the TRIG. LEVEL CENTERING control until stable triggering is obtained with the black TRIGGER SELECTOR control in both the +INT, and -INT. position. Rotate the TRIG. SENS, control clockwise until unstable triggering occurs, then back off a few degrees into a stable region. Check that the sweep is triggered on the + slope of the calibrator waveform when the black TRIGGER SELECTOR control is set at +INT., and on the - slope when the switch is set at -INT. If the slopes are reversed, reset the TRIG. SENS and TRIG. LEVEL CEN-TERING controls slightly until the slopes are correct and stable triggering is obtained. Perform the next step immediately.

10. INT. TRIG. DC LEVEL R3. Set the A VERTICAL VOLTS/DIV. control to the .5 position, center the display vertically,

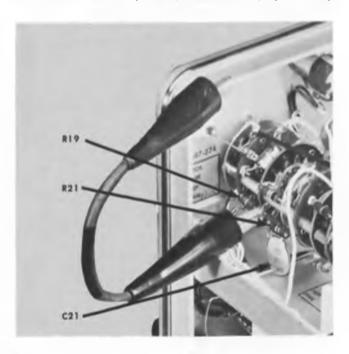


Fig. 6-6. Grounding the Junction of R19, R20, and R21 to simplify the adjustment of the TRIG. LEVEL CENTERING and TRIG. SENS. adjustments.

and turn the red TRIGGER SELECTOR control to the DC position. Then, while switching the black TRIGGER SELECTOR control back and forth between +INT and -INT., adjust the INT. TRIG. DC LEVEL adjustment for stable triggering in both positions. It may be necessary to adjust the A VERTICAL POSITION control slightly to obtain stable triggering. Disconnect the jumper wire and the calibrator signal.

11. PRESET STABILITY R111.\* Set the red TRIGGER SELECTOR control to the AUTO, position. Set the PRESET STABILITY control (screwdriver adjustment, front panel) to its fully counterclockwise position and connect a voltmeter between the center arm of the PRESET STABILITY potentiometer and ground (see Fig. 6-7). Next, advance the PRESET STABILITY control clockwise until a trace first appears on the crt. Note the voltmeter reading for this setting of the control. Then, advance the PRESET STABILITY control further clockwise until the trace brightens and note the voltmeter reading for this setting. Finally, back off the control until the voltmeter indicates a reading midway between the two previous readings. Disconnect the voltmeter.

\* Although this control is shown on the Time-Base Generator circuit diagram, it is more closely associated with the triggering circuitry and its adjustment is therefore included in this section.

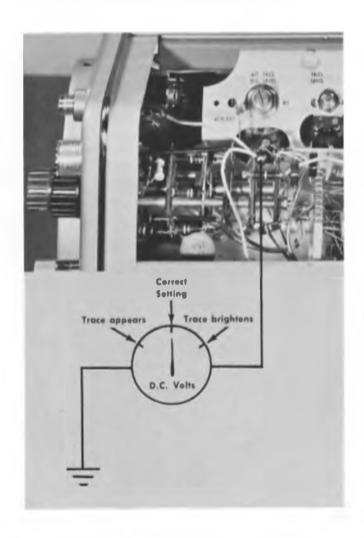


Fig. 6-7. Adjustment of the PRESET STABILITY control.

# Time-Base Generator and Horizontal Amplifier

The time-base circuits of the Type 516 should not require frequent readjustment. As a general rule, if the need for adjustment is indicated, you should first check all of the time-base ranges before making any adjustments. Often, only one control is misadjusted and can be recalibrated according to the following procedures.

Any nonlinearity present in the time base will generally be confined to the first major division of horizontal deflection. In these instructions, therefore, we recommend calibrating the time-base circuits on the basis of time markers appearing between the second and tenth vertical graticule lines (one division in from the left- and right-hand edges of the graticule).

In the instructions that follow, some of the adjustments interact to a degree. For this reason, it is important that you make the adjustments in the proper sequence.

Some of the Horizontal Amplifier adjustments affect the horizontal position of the crt display. As a result, it will be necessary to reposition the display with the HORIZONTAL POSITION control to keep the time markers properly positioned with respect to the graticule lines.

Set up the Type 516 front panel controls as follows:

A A			

TRIGGER SELECTOR (red)	AC
TRIGGER SELECTOR (black)	+INT.
STABILITY	PRESET
TRIGGERING LEVEL	0
HORIZ. DISPLAY	NORM.
TIME/DIV.	1 mSEC.
VARIABLE TIME/DIV.	CALIBRATED
MODE	A ONLY

\* Make sure the VARIABLE TIME/DIV, control stays in the CALI-BRATED position during all timing adjustments.

12. MAG. GAIN ADJ. R259. Connect the output of the time-mark generator to the A VERTICAL input connector and set the time-mark generator controls for a 100-microsecond marker output. Adjust the Type 516 A VERTICAL VOLTS/DIV. control for a vertical deflection of about 3 major divisions. Center the display on the graticule with the PO-SITION controls.

To calibrate the magnifier circuit, turn the HORIZ. DIS-PLAY switch to the MAG. position and adjust the MAG. GAIN ADJ. so that there are two time markers per major division.

13. SWP. GAIN ADJ. R225. Turn the HORIZ, DISPLAY switch to the NORM, position, remove the 100-microsecond markers and apply 1000-microsecond (1-millisecond) markers to the A VERTICAL input connector of the oscilloscope. Adjust the SWP, GAIN ADJ, for one marker per major division.

- **14. SWP. LENGTH R176.** With 1-millisecond markers applied to the Vertical Amplifier, adjust the SWP. LENGTH adjustment for a sweep length of 10.5 divisions.
- 15. SWP./MAG. REGIS. R358. Set the HORIZ. DISPLAY switch to the MAG. position and adjust the HORIZONTAL POSITION control to align the first marker with the vertical centerline of the graticule. Then set the HORIZ. DISPLAY switch to the NORM. position and adjust the SWP./MAG. REGIS. adjustment to again align the first marker with the center of the graticule. Repeat this step until the marker remains horizontally centered as the HORIZ. DISPLAY switch is changed from MAG. to NORM. Set the HORIZ. DISPLAY switch to the NORM. position.
- 16. Check Sweep Timing—2 seconds/div. to .1 millisecond/div. This step checks the accuracy of the sweep timing components for sweep rates between 2 seconds per division and .1 millisecond per division. There are no adjustments to be made. Table 6-1 lists the time markers to be applied for the indicated settings of the TIME/DIV. switch and the number of markers per division to be observed for each setting. When checking sweep rates between .1 SEC and 2 SEC, it will be necessary to adjust the STABILITY control for a stable display.

TIME/DIV.	TIME MARKERS	OBSERVE
.1 mSEC	100 μsec	1 marker/div.
1 mSEC	1 msec	1 marker/div.
2 mSEC	1 msec	2 marker/div.
5 mSEC	5 msec	1 marker/div.
10 mSEC	10 msec	1 marker/div.
.1 SEC	100 msec	1 marker/div.
1 SEC	1 sec	1 marker/div.
2 SEC	1 sec	2 marker/div.

TABLE 6-1

17. 1-, 2-, and 5- $\mu$ sec/div. Sweep Rates. Capacitor C160C determines the sweep rate for the 1-, 2-, and 5- $\mu$ sec positions of the TIME/DIV. control.

Set the oscilloscope controls as outlined at the beginning of the Time-Base Generator and Horizontal Amplifier section with the exception of the TIME/DIV. control and the HORIZ. DISPLAY switch. Set these controls to  $5\mu$ SEC and MAG., respectively. The STABILITY control may have to be adjusted to obtain a stable display in the following timing adjustments.

Connect the time-mark generator to the A VERTICAL input connector and display 1-microsecond markers. With the HORIZONTAL POSITION control, position the display to the left so that the last ten timing markers are visible. Adjust C160C so that each time marker is directly coincident with a vertical graticule line. It will be necessary to adjust the HORIZONTAL POSITION control, as C160C is adjusted, to align the time markers with the graticule lines.

18. 1-, 2-, and 5- $\mu$ sec/div. Linearity. Capacitor C330 compensates a voltage divider network at the input circuit of the Horizontal Amplifier, and affects the sweep rate of the early part of the display at the faster sweep rates.

With the controls unchanged from Step 17, position the display to the right until the first ten timing markers are visible. Adjust C330 to align each time marker with a vertical graticule line. Again, it will be necessary to adjust the HORIZONTAL POSITION control, as C330 is adjusted, to align the time markers with the graticule lines.

There is some interaction between C160C and C330. It may be necessary to go back and forth two or three times between steps 17 and 18 to obtain satisfactory calibration.

19. 10-, 20-, and 50- $\mu$ sec/div. Sweep Rates. Capacitor C160E determines the sweep rate for the 10-, 20-, and 50- $\mu$ sec positions of the TIME/DIV. switch. To make this adjustment, set the HORIZ. DISPLAY switch to NORM. and the TIME/DIV. switch to 10  $\mu$ SEC. Display 10-microsecond markers from the time-mark generator.

Adjust C160E to obtain one time marker for each graticule line.

**20.** 2- $\mu$ sec/div. Linearity. Capacitor C224 is one of three adjustments that determine the high-frequency response of the horizontal amplifier, and thereby the linearity of the crt display at the faster sweep rates. (The other two adjustments are C260 and C384). The time constants of these circuits are such as to permit adjustments at different sweep rates. C224 affects the response when the HORIZ. DISPLAY switch is in the NORM. position only, and is initially adjusted to provide the best linearity in the 2- $\mu$ sec/div. range.

To make this adjustment, turn the TIME/DIV. control to 2  $\mu$ SEC and display 1-microsecond markers from the time-mark generator. Adjust C224 so that two markers per division are displayed on the left half of the screen. You may ignore the linearity on the right half at this point.

21. .2- and .5- $\mu$ sec/div. Sweep Rate. Capacitor C160A determines the sweep rate for the .2 and .5  $\mu$ SEC positions of the TIME/DIV. control.

Continue displaying 1-microsecond markers, and turn the TIME/DIV. control to .5  $\mu$ SEC. Make sure the VARIABLE TIME/DIV. control remains in the CALIBRATED position. Adjust C160A so that a time marker is aligned with every other graticule line.

22. .2- $\mu$ sec/div. Linearity. At the .2- $\mu$ sec/div. sweep rate the linearity of the crt display can be altered slightly by adjusting C260. In addition, the linearity adjustments for lower sweep rates become timing adjustments at the higher sweep rates. This is particularly true of C224 (step 20), which has a pronounced effect on the .2- $\mu$ sec/div. sweep rate. To adjust C260 (which may require a readjustment of C224), set the oscilloscope controls as follows:

TRIGGER SELECTOR (red)	HF SYNC
TRIGGER SELECTOR (black)	+INT.
HORIZ. DISPLAY	NORM.
TIME/DIV.	.2 μSEC
VARIABLE TIME/DIV	CALIBRATED

Set the time-mark generator for a 50 nanosecond sinewave output, and adjust the STABILITY and A VERTICAL VOLTS/

### Calibration — Type 516

DIV. controls for a stable display of about five divisions of vertical deflection. Now, adjust C260 for the most linear display. If the initial adjustment of C224 was correct, there will be four cycles per division in the display. If this is not the case, readjust C224 to obtain two cycles per division.

If you find it necessary to adjust C224 in this step, repeat steps 21 and 22.

23. .04-µsec/div. Sweep Rate and Linearity. Capacitor C384 affects the sweep rate and linearity of the display when the TIME/DIV. control is set to .2 µSEC and the HORIZ. DISPLAY control is in the MAG. position (this increases the sweep rate five times from .2 µsec/div. to .04 µsec/div.).

Set the oscilloscope controls as outlined in step 22 with the exception of the HORIZ. DISPLAY control; set this control to the MAG. position. Display a 50 nanosecond sine wave from the time-mark generator and adjust C384 to obtain eight cycles within the entire 10 divisions of horizontal deflection.

Disconnect the time-mark generator.

# **Vertical Amplifier**

This section contains instructions for adjusting the Vertical Amplifier in the Type 516 Oscilloscope. There is interaction between some of the vertical amplifier adjustments; for this reason, the adjustments should be made in the sequence that follows:

**24. DC Balance.** This adjustment is performed by the operator of the oscilloscope in the course of normal operation (it should be done every day). However, the maintenance technician must check the adjustment at this point before proceeding with the calibration of the vertical amplifier.

Misadjustment of the control is indicated if a free-running trace (no signal) shifts vertically as the VARIABLE VOLTS/DIV. controls are rotated. To perform this adjustment, it is necessary to obtain a reference trace on the crt. Set the STABILITY control to the clockwise (free run) position and the TIME/DIV. control to 1mSEC. Set the MODE switch to the ALTERNATE position. Two traces should be displayed on the crt, one controlled by the A VERTICAL POSITION control and one controlled by the B VERTICAL POSITION control. Adjust the A VERTICAL DC BAL. control (front panel) until there is no longer any vertical shift in the A Channel trace as the A VERTICAL VARIABLE control is rotated. Adjust the B VERTICAL DC BAL. control until there is no longer any vertical shift in the B Channel trace as the B VERTICAL VARIABLE control is rotated.

**25. Amplifier Gain.** Two adjustments, GAIN ADJ. and CHANNEL A GAIN SET, determine the gain of the vertical amplifiers and therefore the calibration of the VOLTS/DIV. controls.

To adjust the GAIN ADJ. adjustment (R478) set the oscilloscope controls as follows:

TIME BASE

TRIGGER SELECTOR

(red)

AUTO.

TRIGGER SELECTOR (black)		+INT.
STABILITY	)	not used in
TRIGGERING LEVEL	•	AUTO. mode.
HORIZ. DISPLAY		NORM.
TIME/DIV.		1 mSEC
VARIABLE TIME/DIV.		CALIBRATED
MODE		B ONLY
B VERTICAL		
VOLTS/DIV.		.05
VARIABLE TIME/DIV.		CALIBRATED
POLARITY		NORM. AC
AMPLITUDE CALIBRATOR		.2

Connect a jumper wire from the B VERTICAL input connector to the CAL. OUT connector, and adjust the INTENSITY, FOCUS, ASTIGMATISM and POSITION controls for a suitable display. Make sure the B VERTICAL VARIABLE control is in the CALIBRATED position, and adjust the GAIN ADJ. adjustment for a deflection of exactly four major divisions.

To adjust the CHANNEL A GAIN SET adjustment (R440), set the oscilloscope controls as follows. Controls not mentioned are left in the position they were in at the beginning of step 25.

MODE	A ONLY
A VERTICAL	
VOLTS/DIV.	.05
VARIABLE	CALIBRATED
POLARITY	NORM. AC
POSITION	midrange

Connect the jumper wire from the A VERTICAL input connector to the CAL. OUT connector. Make sure the A VERTICAL VARIABLE control is in the CALIBRATED position, and adjust the CHANNEL A GAIN SET adjustment for a deflection of exactly four major divisions. Remove the jumper wire.

**25.** Attenuator High Frequency Compensation. To adjust the high-frequency compensation of the attenuators, set the front-panel controls as follows:

+INT.
AUTO.
NORM.
.5 mSEC
CALIBRATED
A ONLY
.05
CALIBRATED
NORM. AC
midrange

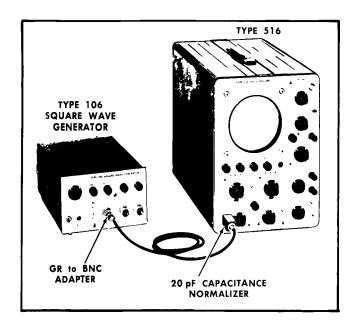


Fig. 6-8. Tektronix Type 106 Square-Wave Generator connected for attenuator high frequency compensation.

Connect the output of the square-wave generator through the 20-pF capacitance normalizer to the A VERTICAL input connector of the Type 516. Adjust the square-wave generator for an output frequency of 1 kHz. Fig. 6-8 shows the Tektronix Type 106 Square-Wave Generator connected for high-frequency compensation of the A Channel attenuator.

Set the A VERTICAL VOLTS/DIV. control to the .05 po-

sition and adjust the output amplitude of the square-wave generator to produce about 4 divisions of deflection on the crt screen. With the A VOLTS/DIV. switch set to the .05 position adjust the Output Amplitude control of the Type 106 to obtain 4 divisions of vertical deflection on the crt of the Type 516. If the signal can't be reduced to 4 divisions you may insert the 10XT/B52L10 attenuator. The attenuator should be placed between the Type 106 and the 20 pF capacitance normalizer.

Observing the top of the square wave adjust C403 for SN range 101-569, and C420 for SN 570 and up, for the best square wave response. For instruments with SN 570 or higher change the POLARITY switch to the INV. AC position. Observing the bottom of the square wave this time adjust C421 for the best square wave response. Then set the A VERTICAL VOLTS/DIV. control to each of its other settings, and at each setting adjust the two capacitors exposed in the SHUNT-SERIES opening below the A VERTICAL VOLTS/ DIV. control for best square-wave response. In each case the SERIES capacitor (the adjustment on the right) affects the attenuator high-frequency compensation (see Fig. 6-9), and the SHUNT capacitor (the adjustment on the left) affects the attenuator input capacitance (see Fig. 6-10). Maintain about four divisions of vertical deflection on the crt screen by adjusting the output amplitude of the square-wave generator as you switch the VOLTS/DIV, control from one setting to the next. (With the Tektronix Type 106 Square-Wave Generator, the maximum vertical deflection will be less than four divisions when the VOLTS/DIV. control is in the 20 VOLTS/ DIV. position. The square wave presentation should be checked in the NORM and INV. positions of the POLARITY switch. They should look the same in either position of the switch.

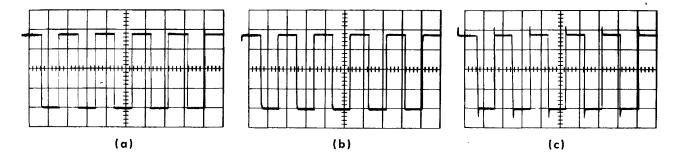


Fig. 6-9. Compensating the attenuator high-frequency response; (a) proper compensation, (b) undercompensation, and (c) overcompensation.

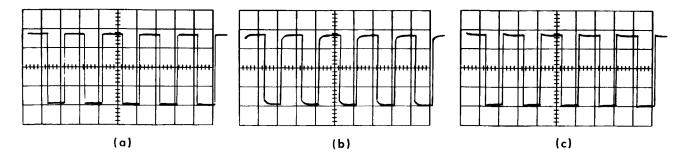


Fig. 6-10. Adjusting the input capacitance; (a) proper adjustment, (b) and (c) improper adjustment.

Adjustment of B channel is done in the same manner as outlined above. C503 is adjusted for instruments below SN 570 while C520 and C521 are adjusted for instruments above this serial number.

27. Switched Amplifier Interelectrode Capacitance Compensation. When a signal with fast-rising portions is applied to one of the input channels, and that input channel is "turned off" by the action of the Switching Circuit, the grid-to-plate capacitance of the Switched Amplifier in that channel tends to couple a part of the signal through to the Output Amplifier. When this happens, the coupled-through portions of the signal in the turned-off input channel appears as spikes on the signal in the other channel. To counteract this tendency, C450, C451, C550, and C551 must be adjusted to compensate for the grid-to-plate capacitance of their respective Switched Amplifier tubes.

To adjust C450 and C451, apply a 1-kHz square wave from the Type 106 Square-Wave Generator to the A VERTICAL input connector. Set the black TRIGGER SELECTOR control to EXT. and apply an external triggering signal from the square-wave generator to the TRIGGER INPUT connector. Set both VOLTS/DIV. controls to .05. Set the MODE switch to A ONLY and the TRIGGERING LEVEL and STABILITY controls for a stable display and several cycles of the squarewave signal. Set the amplitude control of the square-wave generator so that the peak-to-peak amplitude of the displayed square waves is slightly greater than the diameter of the face of the crt. Set the MODE switch to B ONLY. Inspect the trace very carefully for small negative and positive spikes along its length. Adjust C450 and C451 as necessary to reduce the amplitude of these spikes, if any, as much as possible. Try to adjust both capacitors approximately the same amount in order to maintain the balance of the channel. When the adjustment is completed, both adjusting screws should extend about the same distance above the capacitor bodies.

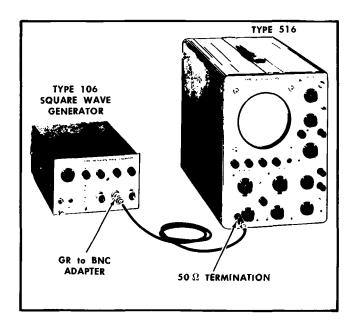


Fig. 6-11. Connecting the Tektronix Type 106 Square-Wave Generator to the Type 516 Oscilloscope for adjustment of the Delay Line.

To adjust C550 and C551, apply the output of the square-wave generator to the B VERTICAL input connector, and set the MODE switch to A ONLY. Then adjust C550 and C551 in the manner described for C450 and C451 in the previous paragraph.

These adjustments interact with the adjustment of the Delay Line described in the following paragraphs. Therefore, whenever you have changed the setting of any of these capacitors, you should check the adjustment of the Delay Line.

# Delay Line and Vertical Amplifier High Frequency Compensation

Of all the adjustments you may be called upon to perform on the Type 516 Oscilloscope, the adjustment of the Delay Line and the high-frequency compensation of the Vertical Amplifier will be the most critical. This is due largely to the interaction between the adjustments. There are 42 variable capacitors and 2 variable inductors associated with the Delay Line, and 2 variable inductors and 4 variable capacitors in the Vertical Amplifier, and all of the adjustments interact to some degree.

Before making any of the adjustments decribed in this section, read the instructions carefully so that you will be sure of what is to be done. Study the photographs and illustrations carefully in order to obtain a clear understanding of the result of each adjustment. Attempts to adjust the Delay Line without adequate preparation frequently lead to a misadjustment more severe than the initial condition.

Displaying the Test Signal. To determine whether the Delay Line is in need of adjustment, and to make the necessary adjustments, you will need to apply a square wave to the oscilloscope and observe the waveform displayed on the crt. The risetime of the square wave applied will affect the accuracy of the tuning of the Delay Line; the shorter the risetime of the applied square wave, the greater will be the ability to tune accurately. For this reason, the risetime of the square wave applied should be kept as short as possible; 15 nanoseconds or less may be considered satisfactory for most uses of the oscilloscope. Also, the waveform should be free of overshoot and irregularities.

The Type 106 Square-Wave Generator which has a risetime of 1 nanosecond or less when connected as shown in Fig. 6-11 is suitable for use in tuning the delay line of the Type 516 oscilloscope. All tuning should be done on the positive half-cycle of the waveform.

A square-wave frequency of about 450 kHz is optimum for tuning the Type 516 Oscilloscope. This frequency permits the broadest oscilloscope display while preventing the appearance of reflected aberrations from the preceding rise or fall of the waveform (these reflected aberrations can appear at frequencies over 500 kHz).

To check the response of the Vertical Amplifier and Delay Line, apply the 450-kHz square wave to the A VERTI-CAL input connector and set the front-panel controls of the Type 516 as follows:

AC

TRIGGER SELECTOR	+INT.
(black)	DDFCFT
STABILITY	PRESET
HORIZ. DISPLAY	NORM.
TIME/DIV.	2 μSEC
VARIABLE TIME/DIV.	CALIBRATED
MODE	A ONLY
A VERTICAL	
VOLTS/DIV.*	.05
VARIABLE TIME/DIV.	CALIBRATED
POLARITY	NORM. DC

\*It is very important to bypass the effect of the attenuators in the procedure that follows. Make sure that the VOLTS/DIV. control is set to the .05 position and that the VARIABLE control remains in the CALIBRATED position.

Adjust the TRIGGERING LEVEL control for a stable display and adjust the output amplitude control of the square-wave generator for approximately four divisions of vertical deflection. Position the display so that it is similar to Fig. 6-12. It may be necessary to adjust the frequency control of the square-wave generator slightly to obtain the desired number of cycles on the crt screen.

After obtaining the desired display you can begin your examination for waveform distortion. There are three general characteristics to appraise in determining the response of the Vertical Amplifier and the Delay Line. The first is the "level" of the display, the second is the amount of "wrinkle" in the flat portion of the display, and the third is the "squareness" of the leading edge and corner of the display.

Determining the Level of the Display. To determine the level of the display, position the waveform so that the flat top of the positive portion coincides with a horizontal graticule line. If the top of the waveform is coincident with the graticule line for the entire positive half-cycle, the display is level. If there is either an upward or a downward slope to the top of the waveform, the display is not level. When the Delay Line is properly adjusted the display will be level. Any departure from a level display is the result of a collective misadjustment of several Delay-Line capacitors. While it is possible to observe the level at a sweep rate of  $2 \mu \text{sec/div.}$ , the level is most easily obeserved at the 2 or  $5 \mu \text{SEC}$  position of the TIME/DIV. control. A waveform that is level, and two waveforms that are not level, are shown in Fig. 6-13.

Determining the Amount of Wrinkle in the Display. There are two general types of "wrinkle conditions" that may appear in a display as a result of an improperly adjusted Delay Line. One type is that shown in Fig. 6-14 (a). This type of irregularity, which is most easily observed at a sweep rate of .2  $\mu$ sec/div., is generally caused by the misadjustment of a group of capacitors. However, if there appears to be a certain rhythmic waviness or symmetry to the distortion, the condition may be caused by faulty adjustment in either the termination network or in the high-frequency compensation of the Vertical Amplifier.

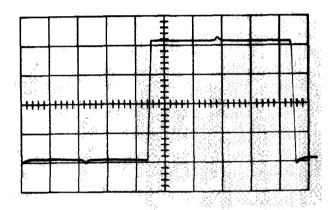
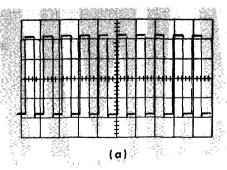
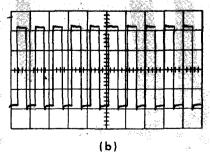


Fig. 6-12. 450-kHz square wave displayed on the Type 516 Oscilloscope at a sweep rate of 0.2  $\mu$ sec/div. The termination bump has been accentuated for this photograph.

The second type of wrinkle to look for is shown in Fig. 6-14 (b). This condition, a single aberration in the waveform, is also most easily observed at a sweep rate of  $.2~\mu sec/div.$ , and is caused by the misadjustment of a single Delay-Line capacitor.

Determining the Squareness of the Corner and Leading Edge. The risetime of the leading edge of the displayed square wave, and the squareness or sharpness of the corner at the top of the leading edge, are determined by the high-frequency response of the Delay-Line adjustments adjacent to the Vertical Amplifier. This condition is most easily observed at a sweep rate of  $.2 \,\mu sec/div$ . The corner at the leading edge of the waveform should be as sharp or square as possible, but must be free of overshoot and wrinkles. The





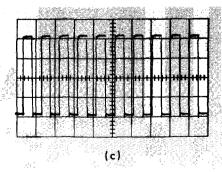


Fig. 6-13. Determining the level of the display. The frequency of the square wave is 450-kHz; the sweep rate is 2  $\mu$ sec/div. (a) shows a level waveform, (b) shows an upward slope to the waveform.

6-11

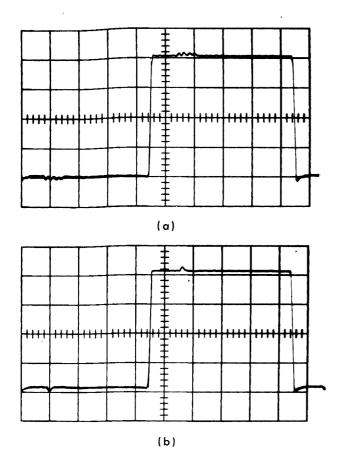


Fig. 6-14. Determining the amount of wrinkle in the display. The aberrations in (a) are caused by a misadjustment of a group of capacitors; the single bump in (b) is caused by the misadjustment of a single capacitor.

correct type of waveform is shown in Fig. 6-15 (a). The rounded-corner type of waveform, shown in Fig. 6-15 (b) is the result of insufficient high-frequency compensation, and the overshoot condition shown in Fig. 6-15 (c) is the result of overcompensation.

# **Delay Line Adjustment Procedure**

There are four major steps in adjusting the Delay Line and the Vertical Amplifier in the Type 516 Oscilloscope. These are:

- 1. Presetting the adjustments
- 2. Establishing a level display
- 3. Removing the wrinkles
- 4. Compensating the amplifier

It may be necessary, in some instances, to perform all of the steps listed in these instructions. In other cases, and particularly where the degree of misadjustment is minor, the first step can usually be omitted.

**Presetting the Adjustments.** If the displayed square-wave on the crt indicates that the Vertical Amplifier and the Delay Line are considerably out of adjustment, presetting the controls before starting the adjustment procedure will generally render the best results.

The variable inductors, L455 and L555, in the Vertical Amplifier, should be preset so that the slugs are positioned deeper into the coil form than the windings of the coil extend. When the slugs are positioned below the coil windings, light can be observed between the top of the slugs and the bottom of the windings. Presetting the inductors in this manner reduces their effect in the circuit, and the adjustment of the Delay Line is simplified.

The variable capacitors in the Delay Line should be preset so that the top of the adjusting screw extends about one-quarter of an inch above the top of the contact springs. The important characteristic is that the tops of all the adjusting screws should be at about the same height, both after the preset procedure and after the final adjustment.

The inductors and the capacitors in the termination network (L485, L486, C485, C486) are the first to be adjusted in the adjustment procedure; for this reason it is not necessary to preset these controls. However, it is important to check the inductors for balance. The slugs in each inductor should be set to the same depth within the coil form.

**Establishing a Level Display.** The reference level for the displayed waveform is established by that portion of the

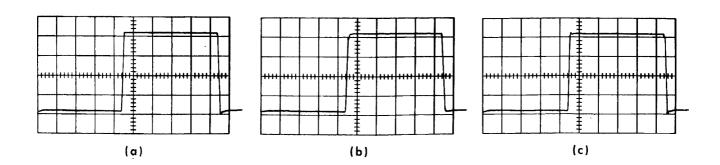


Fig. 6-15. Determining the squareness of the corner and leading edge. (a) shows the ideal waveform; the rounded corner in (b) is the result of insufficient high-frequency compensation in the amplifier; and the overshoot in (c) is the result of excessive high-frequency compensation.

square wave following the termination network. Make sure that the VOLTS/DIV. control is set to .05 and the VARIABLE control is set to the CALIBRATED position. Set the TIME/DIV. control to .2  $\mu$ SEC. Observe the waveform closely in the region near the center of the positive portion, and adjust the two inductors and the two capacitors in the termination network for the minimum termination bump. Be sure to check the inductors for balance as described in the previous paragraph.

The level of that portion of the square wave preceding the termination bump is determined by the collective effect of all the Delay-Line capacitors. Set the TIME/DIV. control to 2  $\mu$ SEC; this will produce narrow pulses on the crt screen and make it easy to ascertain any departure from a level display. Each capacitor must now be adjusted a small amount in a direction that will make the top of the square wave level. Start at the terminated end of the Delay Line and adjust each capacitor a small amount, carefully observing the top of the waveform for the result. Make sure that you retain a level top to the waveform as you progress toward the amplifier end of the line.

After you have been over the line once, and have established an average level for the waveform, advance the sweep rate to 1  $\mu$ sec/div. and repeat the procedure. This time, try to adjust the capacitors for a smooth transition from bump to bump, at the same time maintaining the level. Do not try to obtain a wrinkle-free line at this time. Just try to reduce the amplitude of all the bumps the same amount. The important consideration is to maintain the level of the waveform.

Removing the Bumps and Wrinkles. After you have established a level display, and reduced the amplitude of the bumps a bit, you can start removing the wrinkles and bumps to a greater degree. Again, start at the terminated end of the line and work toward the amplifier end.

Set the TIME/DIV, control to .5  $\mu$ SEC and adjust the termination network again to reduce the wrinkles in the vicinity of the termination bump. Do not try to achieve a perfectly straight line across the top of the waveform at this time. Just reduce the amplitude of the bumps by about one-half. Then advance to the first group (about 4 or 5) of the capacitors in the line and adjust them for a reasonably smooth line over that portion of the display they affect. Keep in mind that each capacitor will only require a slight adjustment, and that it is combined effect of a group of capacitors with which you must be concerned. As you advance along the line, from each group of capacitors to the next, turn the VARI-ABLE TIME/DIV. control full left from time to time; this will decrease the sweep rate and narrow the pulse width so that you may more easily check the level of the waveform. It is just as important to maintain the level of the waveform as it is to achieve a wrinkle-free display.

After you have traversed the entire length of the Delay Line, advance the sweep rate to .2  $\mu$ sec/div. and repeat the process. (Adjust the VARIABLE TIME/DIV. control to obtain the entire positive half-cycle on the crt.) Be extremely careful in your adjustments at this time. Any capacitors that require adjusting will only need a slight "touch". Any overadjustment might nullify all of your efforts up to this point. From time to time switch back to a sweep rate of 2  $\mu$ sec/div. to check the level.

Upon completing this portion of the adjustment procedure, the display on the oscilloscope should appear similar to Fig. 6-15 (b). That is, the display should be level and free from bumps and wrinkles, with a pronounced rolloff at the leading corner. The final step in the adjustment procedure, therefore, is to square up the corner.

**Squaring up the Corner.** With the TIME/DIV. control set to .2  $\mu$ SEC, position the display so that you have a good view of the leading edge and corner. Then, adjust L455 and L555 in the Vertical Amplifier, being careful to adjust each inductor the same amount. Continue adjusting each inductor, maintaining balance, until the leading edge comes up to a sharp, square corner. This process will introduce some new wrinkles in the display, but these can be easily removed by adjusting the first few capacitors in the Delay Line. When the Vertical Amplifier and the Delay Line are in proper adjustment, the display should appear similar to Fig. 6-15 (a).

A slight readjustment of C450 and C451 may also improve the shape of the corner without seriously degrading the response for which they were adjusted in step 27 of the preceding procedure. If you find it necessary to adjust them more than a slight amount, you should go back and check the adjustment as described in step 27. The final setting will be a compromise between the squareness of the corner and the amount of feedthrough allowable; the most emphasis should be put upon the squareness of the corner, since feedthrough from one channel to the other will not be a problem in most applications of the oscilloscope.

Apply the 450-kHz output of the square-wave generator to the B VERTICAL input connector and set the MODE switch to B ONLY. Note the shape of the leading corner. You may wish to adjust C550 and C551 and/or the first few capacitors of the Delay Line to improve the response of the B Channel. The final setting will be a compromise between the response obtained through the A Channel and the response obtained through the B Channel. Usually, it is best to set up one of the channels for optimum square-wave response, even if it means degrading the response on the other channel slightly. Then all critical high-frequency observations can be made on the optimum channel.

Check Bandwidth of Vertical Amplifier. Connect the output cable from the constant-amplitude signal generator to the A VERTICAL input connector of the Type 516. Set up the front panel controls as follows:

TIME BASE

TRIGGER SELECTOR AC (red)

TRIGGER SELECTOR +INT.

(black)

STABILITY Fully clockwise (free run)

HORIZ. DISPLAY NORM.
TIME/DIV. 1 mSEC

VARIABLE TIME/DIV. CALIBRATED

# Calibration — Type 516

WODE	A ONLY
A VERTICAL	
VOLTS/DIV.	.05
VARIABLE VOLTS/DIV.	CALIBRATED
POLARITY	NORM, DC

Set the frequency control on the signal generator for an output frequency of 1 megahertz, and adjust the amplitude control on the signal generator for a vertical deflection of exactly 4 major divisions. Center the display vertically on the crt with the A VERTICAL POSITION control.

Then increase the output frequency of the signal generator to 15 megahertz. The deflection should be at least 2.8 major divisions. This corresponds to an attenuation of 3 dB or less at 15 MHz as specified (see Section 1).

# SECTION 7 MECHANICAL PARTS LIST

**7-1** 

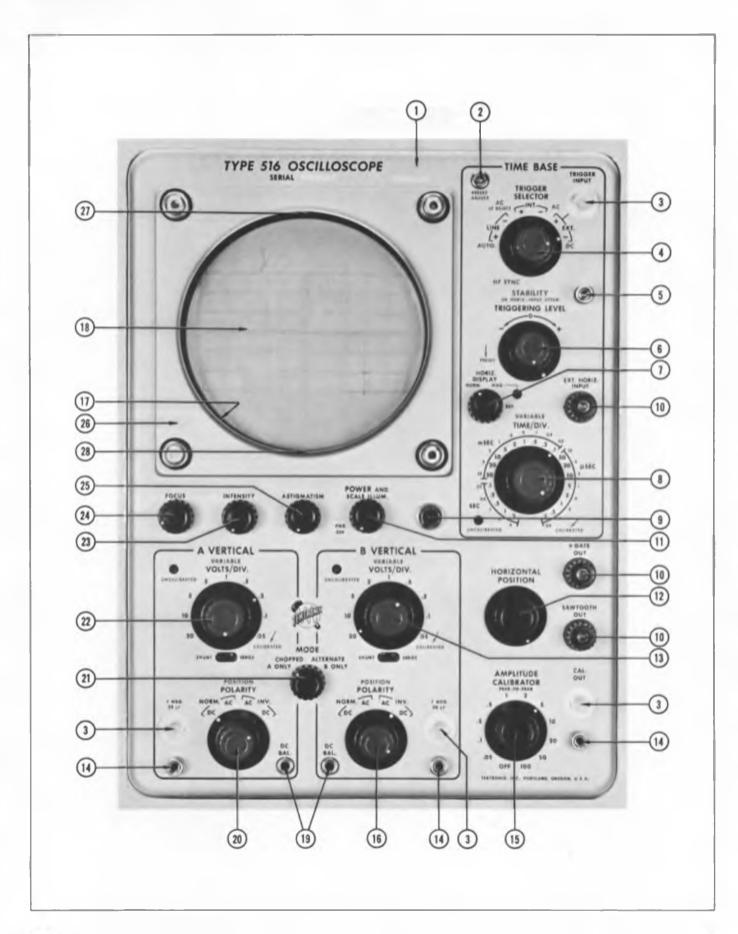


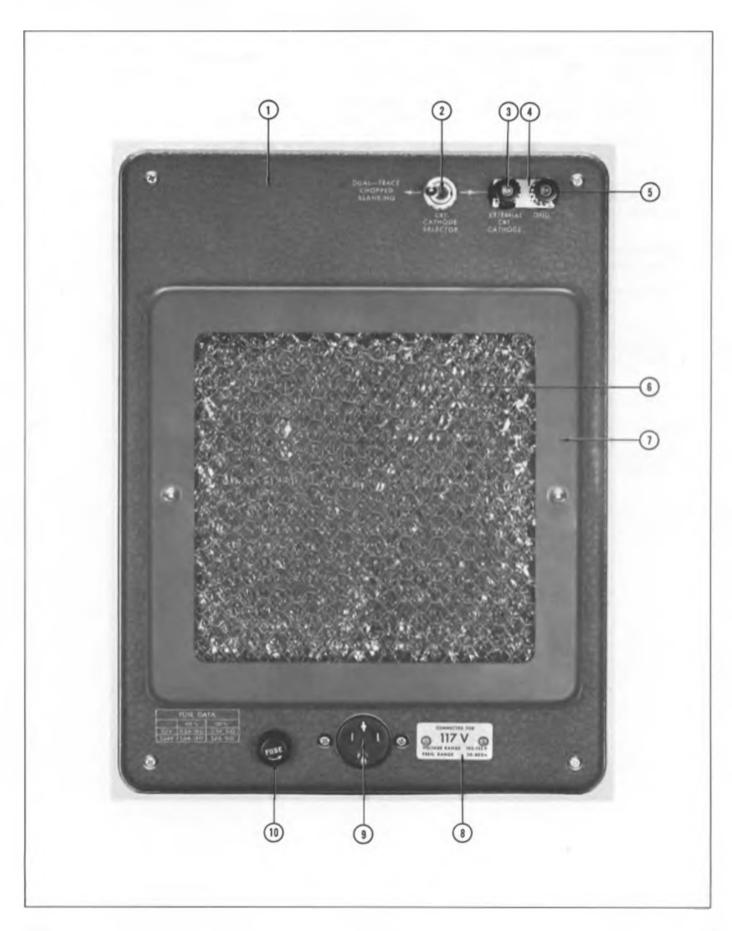
Fig. &				Q	(Com)
Index	Tektronix	Serial/Ma	del No	t	
No.	Part No.		Disc		Description
	1411 110.		D13C	У	1 2 3 4 5
1	333-0603-00	101	269	1	DANET front
1	333-0644-00	270	209	1 1	PANEL, front PANEL, front
	387-0224-00	270		1	
	367-0224-00			- T	PLATE, front subpanel plate includes:
	354-0057-00			1	RING, ornamental
	355-0043-00			4	STUD, graticule (replacement)
				-	each stud includes:
	212-0507-00			1	SCREW, $10-32 \times 3/8$ inch, PHS
	210-0010-00			1	LOCKWASHER, internal, #10
2 '				1	RESISTOR, variable
				-	mounting hardware for each: (not included w/resistor)
	210-0471-00			1	NUT, hex., $1/4-32 \times 19/32$ inch
	210-0223-00			1	LUG, solder, $1/4$ ID $\times$ $7/16$ inch OD, SE
	210-0046-00			1	LOCKWASHER, internal, 0.261 ID $_{ m X}$ 0.400 inch OD
•					
3	131-0081-00	101	2029		CONNECTOR, coaxial, 1 contact, UHF
	131-0126-00	2030		4	CONNECTOR, coaxial, 1 contact, BNC
4	366-0031-00			1	KNOB, small redTRIGGER SELECTOR
				-	knob includes:
	213-0004-00			1	SCREW, set, 6-32 x 3/16 inch, HSS
	366-0029-00	101	3209	1	KNOB, large blackTRIGGER SELECTOR
	366-0142-00	3210		1	KNOB, large charcoalTRIGGER SELECTOR
				-	knob includes:
	213-0004-00	101		1	SCREW, set, 6-32 x 3/16 inch, HSS
	262-0333-00	101	269	1	SWITCH, wiredTRIGGER SELECTOR
	262-0402-00	270	1958	1	SWITCH, wiredTRIGGER SELECTOR
	262-0564-00	1959		1	SWITCH, wiredTRIGGER SELECTOR
				-	switch includes:
	260-0332-00	101	269	1	SWITCH, unwired
	260-0377-00	270		1	SWITCH, unwired
				-	mounting hardware: (not included w/switch)
	210-0013-00			1	LOCKWASHER, internal, 3/8 ID x 11/16 inch OD
	210-0413-00			1	NUT, hex., $3/8-32 \times 1/2$ inch
5	129-0020-00			1	ASSEMBLY, binding post
,	127-0020-00			_	
	355-0503-00			1	assembly includes: STEM
	200-0072-00			1	CAP
					mounting hardware: (not included w/assembly)
	210-0010-00			1	LOCKWASHER, internal, #10
	210-0445-00				NUT, hex., 10-32 x 3/8 inch
	120 0.115 00			-	not, next, to sa h s/o then
6	366-0032-00			1	KNOB, small redSTABILITY
				-	knob includes:
	213-0004-00			1	LOCKWASHER, internal, #4
	366-0030-00	101	3209	1	KNOB, large blackTRIGGERING LEVEL
	366-0146-00	3210		1	KNOB, large charcoalTRIGGERING LEVEL
				-	knob includes:
	213-0004-00			1	SCREW, set, $6-32 \times 3/16$ inch, HSS
				1	RESISTOR, variable
				-	mounting hardware: (not included w/resistor)
	210-0013-00			1	LOCKWASHER, internal, $3/8$ ID x $11/16$ inch OD
	210-0413-00			1	NUT, hex., $3/8-32 \times 1/2$ inch

					FRONT (cont)
Fig. &				Q	
Index	Tektronix	Sorial /A	Model No.	t	
				•	Description
No.	Part No.	Eff	Disc	<u>y</u>	1 2 3 4 5
7	366-0033-00	101	3209	1	KNOB, small blackHORIZ DISPLAY
	366-0148-00	3210		1	KNOB, small charcoalHORIZ DISPLAY
				-	knob includes:
	213-0004-00			1	SCREW, set, $6-32 \times 3/16$ inch, HSS
	358-0029-00			1	
				_	mounting hardware: (not included w/bushing)
	210-0840-00			1	WASHER, flat, 0.390 ID x 9/16 inch OD
	210-0413-00				NUT, hex., $3/8-32 \times 1/2$ inch
8	366-0038-00			1	KNOB, small redVARIABLE
U				_	knob includes:
	213-0004-00			1	SCREW, set, 6-32 x 3/16 inch, HSS
	366-0040-00	101	3209	1	KNOB, large blackTIME/DIV
	366-0160-00	3210	320)	ī	KNOB, large charcoalTIME/DIV
		3220		_	knob includes:
	213-0004-00			1	SCREW, set, $6-32 \times 3/16$ inch, HSS
				-	SEE RIGHT SIDE PAGE FOR SWITCH PARTS
8	366-0038-00			1	KNOB, small redVARIABLE
				_	knob includes:
	213-0004-00			1	SCREW, set, 6-32 x 3/16 inch, HSS
	366-0040-00	101	3209	1	KNOB, large blackTIME/DIV
	366-0160-00	3210		1	KNOB, large charcoalTIME/DIV
				_	knob includes:
	213-0004-00			1	SCREW, set, $6-32 \times 3/16$ inch, HSS
				-	SEE RIGHT SIDE PAGE FOR SWITCH PARTS
9	136-0031-00	101	4359	1	SOCKET, light, red
	136-0031-01	4360		1	SOCKET, light, green
				-	socket includes:
	210-0021-00			1	LOCKWASHER, internal, 0.480 inch ID
	210-0414-00			1	NUT, hex., $15/32-32 \times 9/16$ inch
10	129-0036-00	101	3209	3	POST, binding
	129-0063-00	3210		3	POST, binding
				-	mounting hardware for each: (not included w/post)
	358-0036-00	101	3209	1	BUSHING, binding post, black
	358-0169-00	3210	0150	1	BUSHING, binding post, charcoal
	210-0445-00	101	2159	2	
	220-0410-00	2160	21 E 037		NUT, keps, 10-32 x 3/8 inch
	210-0010-00 210-0206-00	101 101	2159X		LOCKWASHER, internal, #10
	210-0206-00	101	2159X	1	LUG, solder, SE #10
11	366-0033-00	101	3209	1	KNOB, small blackPOWER & SCALE ILLUM
	366-0148-00	3210		1	KNOB, small charcoalPOWER & SCALE ILLUM
				_	knob includes:
	213-0004-00			1	SCREW, set, 6-32 x 3/16 inch, HSS
				1	RESISTOR, variable
				-	mounting hardware: (not included w/resistor)
	210-0013-00			1	LOCKWASHER, interna, 3/8 ID x 11/16 inch OD
	210-0840-00			1	WASHER, flat, 0.390 ID $\times$ 9/16 inch OD
	210-0413-00			1	NUT, hex., $3/8-32 \times 1/2$ inch

Fig. & Index	Tektronix	Serial/	Model No.	Q t	Doggintian
No.	Part No.	Eff	Disc	у	Description 1 2 3 4 5
2	366-0028-00	101	3209	1	KNOB, large blackHORIZONTAL POSITION
	366-0145-00	3210		1	KNOB, large charcoalHORIZONTAL POSITION
	012 0004 00			-	knob includes:
	213-0004-00			1	SCREW, set, 6-32 x 3/16 inch, HSS RESISTOR, variable
				-	mounting hardware: (not included w/resistor)
	210-0013-00			1	LOCKWASHER, internal, 3/8 ID x 11/16 inch OD
	210-0413-00			1	NUT, hex., $3/8-32 \times 1/2$ inch
.3	366-0038-00			1	KNOB, small redVARIABLE
				-	knob includes:
	213-0004-00	101	320 <b>0</b>	1	SCREW, set, 6-32 x 3/16 inch, HSS
	366-0040-00 366-0160-00	101 3210	320 <b>9</b>	1 1	KNOB, large blackVOLTS/DIV KNOB, large charcoalVOLTS/DIV
		5-20		_	knob includes:
	213-0004-00			1	SCREW, set, 6-32 x 3/16 inch, HSS
ı <b>J</b> .	100 0005 00			-	SEE TURRET ATTENUATOR PAGE FOR SWITCH PARTS
4	129-0035-00			3	ASSEMBLY, binding post each assembly includes:
	355-0507-00			1	STEM
	200-0103-00			1	CAP
	210-0455-00			1	NUT, hex., $1/4-28 \times 3/8$ inch
-	210-0046-00	1.01	2200	1	LOCKWASHER, internal, 0.261 ID x 0.400 inch OD
5	366-0028-00 366-0145-00	101 3210	3209	1 1	KNOB, large blackAMPLITUDE CALIBRATOR KNOB, large charcoalAMPLITUDE CALIBRATOR
		3210		_	knob includes:
	213-0004-00			1	SCREW, set, 6-32 x 3/16 inch, HSS
	262-0332-00			1	SWITCH, wiredAMPLITUDE CALIBRATOR
	260-0098-00			- 1	switch includes:
	<b>=</b>			1	SWITCH, unwired mounting hardware: (not included w/switch)
	210-0013-00			1	LOCKWASHER, internal 3/8 ID x 11/16 inch OD
	210-0413-00			1	NUT, hex., $3/8-32 \times 1/2$ inch
.6	366-0031-00			1	KNOB, small redPOSITION
	212 000/ 00			_	knob includes:
	213-0004-00 366-0029-00	101	3209	1 1	SCREW, set, 6-32 x 3/16 inch, HSS KNOB, large blackPOLARITY
	366-0142-00	3210	3409	1	KNOB, large charcoalPOLARITY
				_	knob includes:
	213-0004-00	101	560	1	SCREW, set, 6-32 x 3/16 inch, HSS
	262-0337-00 262-0433-00	101 570	569	1 1	SWITCH, wiredB VERTICAL POLARITY SWITCH, wiredB VERTICAL POLARITY
		570		-	switch includes:
	260-0326-00			1	SWITCH, unwired
	384-0213-00			1	ROD, extension
	376-0014-00 210-0012-00			1 1	COUPLING LOCKWASHER, internal 3/8 ID x 1/2 inch OD
	210-0413-00			2	NUT, hex., 3/8-32 x 1/2 inch
				-	mounting hardware: (not included w/switch)
	210-0013-00			1	LOCKWASHER, internal, 3/8 ID x 11/16 inch OD
	210-0413-00			1	NUT, hex., $3/8-32 \times 1/2$ inch

Fig. & Index No.	Tektronix Part No.	Serial/Mode Eff Disc		Description 1 2 3 4 5
17	401-0004-00		1	CAM, plastic (under cover)
			-	
	211-0025-00		1	SCREW, $4-40 \times 3/8$ inch, FHS
18 19 20	331-0037-00 358-0054-00 366-0031-00 	101 32 3210 56 570	3 1 - 1 09 1 1 9 1 1 1 1 1 1 2	KNOB, large black*-POLARITY KNOB, large charcoalPOLARITY knob includes: SCREW, set, 6-32 x 3/16 inch, HSS SWITCH, wiredA VERTICAL POLARITY SWITCH, wiredA VERTICAL POLARITY switch includes: SWITCH, unwired ROD, extension COUPLING NUT, hex., 3/8-32 x 1/2 inch mounting hardware: (not included w/switch) LOCKWASHER, internal, 3/8 ID x 11/16 inch OD
21	366-0033-00 366-0148-00 	101 32 3210	09 1 1 - 1 1 -	
22	366-0038-00  213-0004-00 366-0160-00  213-0004-00	101 32 3210	09 1 1 - 1	KNOB, small redVARIABLE knob includes: SCREW, set, 6-32 x 3/16 inch, HSS KNOB, large blackVOLTS/DIV KNOB, large charcoalVOLTS/DIV knob includes: SCREW, set, 6-32 x 3/16 inch, HSS
23	366-0033-00 366-0148-00 	101 32 3210	09 1 1 - 1 1 1 1	WASHER, flat, 0.390 ID x 9/16 inch OD

Fig. &				Q	
	Tektronix		Model No.	t	Description
No.	Part No.	Eff	Disc	<u>y</u>	1 2 3 4 5
24	366-0033-00	101	3209	1	KNOB, small blackFOCUS
	366-0148-00	3210		1	KNOB, small charcoalFOCUS
				-	knob includes:
	213-0004-00			1	SCREW, set, $6-32 \times 3/16$ inch, HSS
				1	RESISTOR, variable
	210-0013-00			-	mounting hardware: (not included w/resistor)
	210-0013-00			1 1	,,,,,
	210-0340-00			1	
	210-0413-00			1	No1, nex., 3/0-32 x 1/2 men
25	366-0033-00	101	3209	1	KNOB, small blackASTIGMATISM
	366-0148-00	3210	320)	î	KNOB, small charcoalASTIGMATISM
				_	knob includes:
	213-0004-00			1	SCREW, set, $6-32 \times 3/16$ inch, HSS
				1	RESISTOR, variable
				-	mounting hardware: (not included w/resistor)
	210-0013-00			1	, , , , , , , , , , , , , , , , , , , ,
	210-0840-00				WASHER, flat, 0.390 ID $\times$ 9/16 inch OD
	210-0413-00			1	NUT, hex., $3/8-32 \times 1/2$ inch
26	200-0382-00			1	COVER, graticule
20				_	cover includes:
	354-0116-00			1	RING, ornamental
				-	mounting hardware: (not included w/cover)
	210-0816-00			4	WASHER, rubber
	210-0424-00			4	NUT, knurled, $3/8-24 \times 9/16$ inch
27	337-0187-00			1	CHITTED and to 1. 11 1. ( ) to a
28	124-0068-00			1	- , 6 - (,
20	124-0000-00			1	STRIP, felt (under cover)



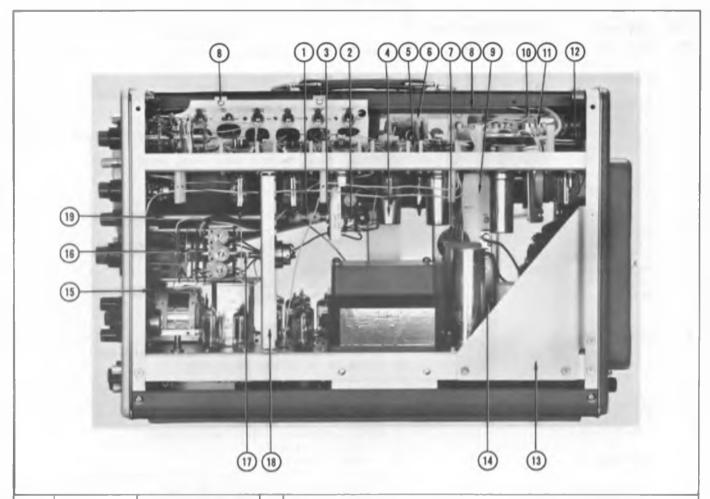
REAR (cont)

				KEAK (CONT)
Fig. &			Q	
Index	Tektronix	Serial/Model N		<b>—</b> • • •
				Description
No.	Part No.	Eff Disc	<u>y</u>	1 2 3 4 5
1	387-0220-00		1	PLATE, rear overlay
			-	mounting hardware: (not included w/plate)
	213-0104-00		4	SCREW, thread forming, #6 x 3/8 inch, THS
	387-0221-00		1	
			-	plate includes:
	354-0057-00		1	
2	260-0209-00		1	
_			-	mounting hardware: (not included w/switch)
	210-0414-00			NUT, hex., 15/32-32 x 9/16 inch
	354-0055-00			
	210-0902-00			RING, locking
				WASHER, flat, 0.470 ID x 21/32 inch OD
	210-0473-00		1	NUT, 12 sided, $15/32-32 \times 0.634$ inch
3	129-0036-00	101 3209	1	POST, binding, black
	129-0063-00	3210		POST, binding, charcoal
			_	mounting hardware: (not included w/post)
	358-0036-00	101 3209	1	BUSHING, binding post, black
	358-0169-00	3210		BUSHING, binding post, charcoal
	210-0445-00	101 2159		NUT, hex., 10-32 x 3/8 inch
	220-0410-00	2160		NUT, keps, 10-32 x 3/8 inch
	210-0010-00	101 21593		LOCKWASHER, internal, #10
	210-0206-00	101 21593		LUG, solder, SE #10
				noo, bolder, on who
6	378-0010-00			FILTER, air
7	380-0017-00		1	HOUSING, air filter
			-	mounting hardware: (not included w/housing)
	212-0031-00		2	SCREW, $8-32 \times 1 \frac{1}{4}$ inches, RHS
	210-0458-00		2	NUT, keps, 8-32 x 11/32 inch
	210-0402-00		2	NUT, hex., $8-32 \times 5/16$ inch
8	334-0649-00		1	TAG, voltage rating
U	334-0047-00		_	mounting hardware: (not included w/tag)
	213-0088-00		2	
	215-0000-00		2	Sokew, thread forming, 4-40 x 1/4 filen, rhs
9	131-0102-00	101 3919	1	CONNECTOR, motor base, 3 wire
	131-0102-01	3920 4419	1	CONNECTOR, motor base, 3 wire
	131-0102-02	4420	1	CONNECTOR, motor base, 3 wire
			-	connector includes:
	129-0041-00	101 3919	1	POST, ground
	129-0041-01	3920 4419	1	POST, ground
	200-0185-00	101 3919	1	COVER, plastic
	200-0185-01	3920 4419	1	COVER, plastic
	204-0335-00	4420	1	BODY, contact assembly
	210-0003-00	101 3919X		LOCKWASHER, external, #4
	210-0551-00	101 3919x		NUT, hex., $4-40 \times 1/4$ inch
	211-0132-00	X3920 4419	1	SCREW, sems, $4-40 \times 1/2$ inch, PHS
	211-0534-00	4420	1	SCREW, sems, $6-32 \times 5/16$ inch, PHS

# Mechanical Parts List—Type 516

# REAR (cont)

Fig. & Index	Tektronix	Serial/	Model No.	Q t	Description
No.	Part No.	Eff	Disc	у	1 2 3 4 5
	211-0015-00	101	3919	1	CODET A AO 1/2 imply DUC
			<del>-</del>		SCREW, 4-40 x 1/2 inch, PHS
	213-0088-00	3920	4419	1	SCREW, thread forming, 4-40 x 1/4 inch, PHS
	213-0146-00	4420		1	SCREW, thread forming, $\#6 \times 0.312$ inch, PHS
	214-0078-00			2	PIN, connecting
	377-0041-00	101	3919	1	INSERT, plastic
	377-0051-00	3920	4419	1	INSERT, plastic
	214-1016-00	4420		1	INSULATOR, connector
	386-0933-00	101	4419	1	PLATE, mounting
	386 <b>-13</b> 56-01	4420		1	PLATE, mounting
				_	mounting hardware: (not included w/connector)
	211-0537-00			2	SCREW, 6-32 x 3/8 inch, THS
	210-0457-00			-	NUT, keps, 6-32 x 5/16 inch
	210-0437-00			2	Not, keps, 0-32 x 3/10 likil
10	352-0002-00			1	ASSEMBLY, fuse holder
				-	assembly includes:
	352-0010-00			1	HOLDER, fuse
	200-0582-00			1	CAP, fuse, black
	210-0873-00			1	WASHER, rubber, 1/2 ID x 11/16 inch OD
				1	NUT



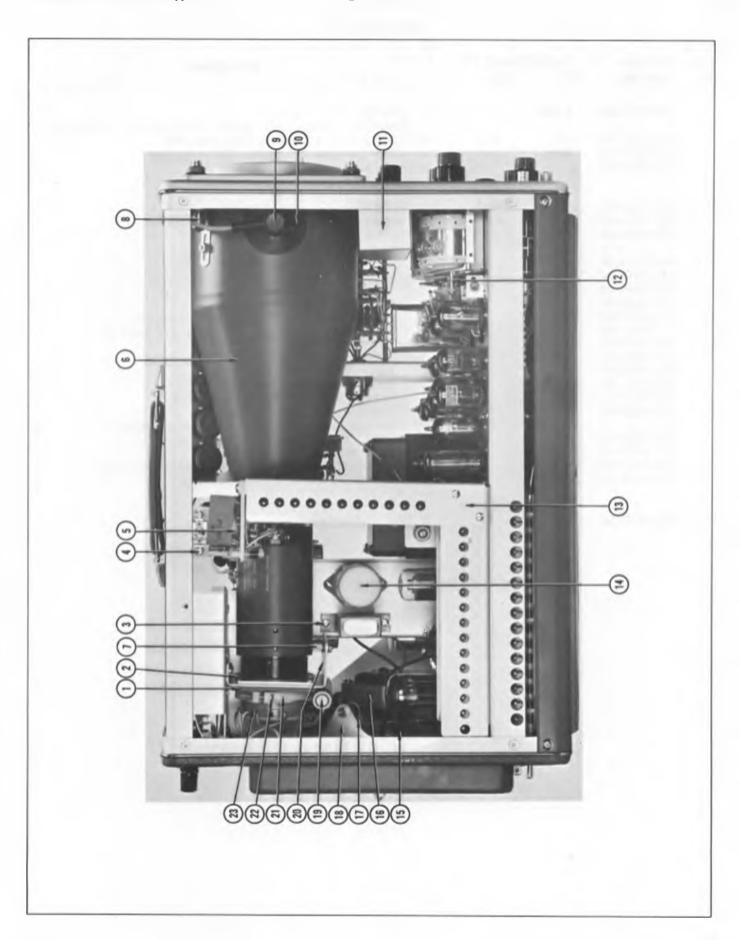
REF, NO.		SERIAL/MODEL NO.					
	PART NO.	EFF.	DISC.	Y.	DESCRIPTION		
1	376-0011-00			1	COUPLING, plastic, insulating		
				-	coupling includes:		
	213-0048-00			2	SCREW, set, 4-40 x 1/8 inch, HSS		
2	262-0125-00	101	569	1	SWITCH, wiredHORIZ. DISPLAY		
	262-0430-00	570		1	SWITCH, wiredHORIZ. DISPLAY		
				-	switch includes:		
	260-0186-00			1	SWITCH, unwired		
				-	mounting hardware: (not included w/switch)		
3	210-0413-00			1	NUT, hex., 3/8-32 x 1/2 inch		
	210-0840-00			1	WASHER, flat, 0.390 ID x 9/16 inch OD		
	210-0012-00			1	LOCKWASHER, internal, 3/8 ID x 1/2 inch OD		
4	337-0008-00			3	SHIELD, tube		
5	210-0204-00				LUG, solder, DE #6		
					mounting hardware: (not included w/lug)		
	211-0038-00			1	SCREW, 4-40 x 5/16 inch, 100° csk, FHS		
	210-0406-00			1	NUT, hex., 4-40 x 3/16 inch		

# RIGHT SIDE (cont)

Fig. & Index No.	Tektronix Part No.	Serial/Model I Eff Disc		Description
	ruii iyo.	Lii Disc	у	1 2 3 4 5
6	386-0533-00		1	PLATE, white plastic mounting hardware: (not included w/plate)
	211-0504-00			SCREW, 6-32 x 1/4 inch, PHS CAPACITOR
			-	mounting hardware: (not included w/capacitor)
7	386-0254-00	X140	1	PLATE, fiber
	211-0534-00	x140		SCREW, sems, $6-32 \times 5/16$ inch, PHS
	210-0006-00	X140		LOCKWASHER, internal, #6
	210-0407-00	X140	2	NUT, hex., 6-32 x 1/4 inch
8	166-0096-00		1	TUBE, conduit
			-	mounting hardware: (not included w/tube)
	343-0015-00	101 1315		
	343-0084-00	1316		CLAMP, strap
	211-0504-00			SCREW, 6-32 x 1/4 inch, PHS
	210-0803-00		2	WASHER, flat, 0.150 ID x 3/8 inch OD
9	387-0219-00		1	PLATE, vertical bulkhead
			-	mounting hardware: (not included w/plate)
	211-0507-00		4	SCREW, 6-32 x 5/16 inch, PHS
10	179-0435-00	101 709	1	CABLE HARNESS, focus & intensity
	179-0132-00	710	1	CABLE HARNESS, focus & intensity
11	124-0092-00		1	STRIP, ceramic, $3/4$ h, $w/3$ notches
			-	strip includes:
	355-0046-00		1	STUD, plastic
	361-0009-00	101 709	- 1	mounting hardware: (not included w/strip) SPACER, plastic, 0.406 inch long
	361-0009-00	710		SPACER, plastic, 0.281 inch long
	301 0000 00	710	1	ornold, prastic, v.201 liter long
12	348-0005-00			GROMMET, rubber, 1/2 inch diameter
13	406-0657-00	x201	1	BRACKET, chassis support
	212 0040 00	W001	-,	mounting hardware: (not included w/bracket)
	212-0040-00 210-0458-00	X201 X201		SCREW, 8-32 x 3/8 inch, 100° csk, FHS NUT, keps, 8-32 x 11/32 inch
	210-0430-00	A201	4	No1, keps, 6-32 x 11/32 Inch
14			1	SWITCH, thermal cutout
	011 0500 00		-	mounting hardware: (not included w/switch)
	211-0503-00 210-0006-00		2	
	210-0006-00			LOCKWASHER, internal, #6 NUT, hex., 6-32 x 1/4 inch
	210 0407-00		2	not, next, 0-32 x 1/4 men
15	352-0008-00	101 3619		HOLDER, neon bulb, single, black
	352-0067-00	3620	4	HOLDER, neon bulb, single, gray

# RIGHT SIDE (cont)

Fig. &				Q	
Index	Tektronix	Serial/	Model No.	t	Description
No.	Part No.	Eff	Disc	_у_	1 2 3 4 5
	378-0541-00	X3620		4	FILTER, lens, neon
				_	mounting hardware for each: (not included w/holder)
	211-0031-00	101	3619	1	
	211-0109-00	3620	30-7	$\bar{1}$	
	210-0406-00			2	
16	262-0334-00	101	3929	1	SWITCH, wiredTIME/DIV
-0	262-0334-01	3930	4769	1	
	262-0334-02	4770	47.05	1	SWITCH, wiredTIME/DIV
		,,,,		_	switch includes:
	260-0329-00			1	SWITCH, unwiredTIME/DIV
17	384-0147-00			1	ROD, extension
	376-0014-00			1	COUPLING
	210-0413-00			2	NUT, hex., $3/8-32 \times 1/2$ inch
	210-0840-00			1	WASHER, flat, 0.390 ID x 9/16 inch OD
	210-0012-00			1	LOCKWASHER, internal, 3/8 ID x 1/2 inch OD
18	406-0582-00			1	BRACK ET
	348-0002-00			3	GROMMET, rubber, 1/4 inch
	348-0003-00			2	GROMMET, rubber, 5/16 inch
	348-0004-00			1	GROMMET, rubber, 3/8 inch
	211-0029-00			2	SCREW, $5-40 \times 3/16$ inch, PHS
	210-0202-00			1	LUG, solder, SE #6
				-	mounting hardware: (not included w/switch)
	210-0803-00			4	
	211-0507-00			4	, , , , , , , , , , , , , , , , , , , ,
	210-0013-00			1	
	210-0413-00			1	NUT, hex., 3/8-32 x 1/2 inch
19	384-0133-00			1	ROD, extension



LEFT SIDE (cont)

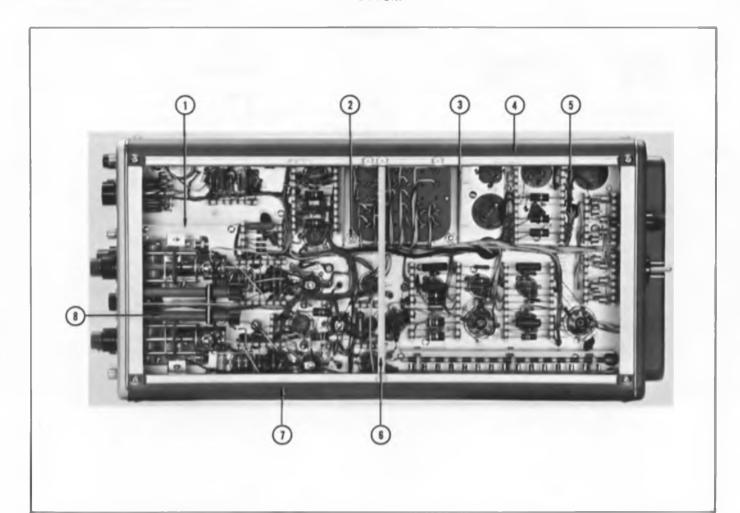
					LEFT SIDE (cont)
Fig. &				Q	
Index	Tektronix	Serial/	Model No.	t	
No.	Part No.	Eff	Disc		Description
140.	ruii 140.		Disc	<u>y</u>	1 2 3 4 5
1	432-0022-00	101	5089		BASE, CRT rotator
	432-0022-02	5090		1	BASE, CRT rotator
				-	mounting hardware: (not included w/base)
	211-0561-00			2	SCREW, 6-32 x 3/8 inch, FH cap
	210-0503-00			1	NUT, rotator securing
	004 1405 00	******		_	
	386-1485-00	X5090		1	PLATE, retaining (not shown)
	211 0022 00			-	mounting hardware: (not included w/plate)
	211-0022-00			1	SCREW, 2-56 x 3/16 inch
2	354-0078-00	101	1319	1	RING, CRT rotator
	354-0178-00	1320			RING, CRT rotator
3					CAPACITOR
				-	mounting hardware: (not included w/capacitor)
	211-0507-00			2	SCREW, 6-32 x 5/16 inch, PHS
	210-0006-00			2	LOCKWASHER, internal, #6
	210-0407-00			2	NUT, hex., $6-32 \times 1/4$ inch
4				1	CARACITOR morialia
4				1	CAPACITOR, variable
	211-0510-00			2	mounting hardware: (not included w/capacitor) SCREW, 6-32 x 3/8 inch, PHS
	210-0802-00			2	
	210-0803-00				WASHER, flat, 0.150 ID x 3/8 inch OD
	210-0202-00				LUG, solder, SE #6
	210-0407-00			2	,
				_	101, 101, 0 3
5				1	RESISTOR, variable
				-	mounting hardware: (not included w/resistor)
	210-0465-00				NUT, hex., 1/4-32 x 3/8 inch
	210-0011-00			1	LOCKWASHER, internal, $1/4$ ID x $15/32$ inch OD
6	337-0088-00			1	SHIELD, CRT
-					mounting hardware: (not included w/shield)
	211-0559-00			5	SCREW, 6-32 x 3/8 inch, 100° csk, FHS
	406-0239-00			2	
	210-0457-00			5	NUT, keps, 6-32 x 5/16 inch
7	211-0514-00			1	
	210-0006-00			1	<u>.</u>
	385-0127-00			1	ROD, hex., 9/32 inch
	210-0803-00				WASHER, flat, 0.150 ID x 3/8 inch OD
	210-0811-00			2	WASHER, fiber, #6
	210-0457-00			1	NUT, keps, $6-32 \times 5/16$ inch
8	136-0001-00			2	SOCKET, graticule light
				_	mounting hardware for each: (not included w/socket)
	211-0534-00			1	
	210-0803-00			1	WASHER, flat, 0.150 ID x 3/8 inch OD
	210-0457-00			1	

# LEFT SIDE (cont)

Fig. & Index	Tektronix	Serial/	'Model No.	Q f	Description
No.	Part No.	Eff	Disc	У	1 2 3 4 5
9	179-0435-00	101	709	1	CABLE HARNESS, focus & intensity
	179-0132-00	710		1	CABLE HARNESS, focus & intensity
				-	cable harness includes:
	131-0086-00			1	CONNECTOR, cable, anode
	131-0073-00			- 1	connector includes: CONNECTOR, CRT brush
	200-0110-00			1	CAP, CRT anode
	134-0031-00			1	
10	200-0112-00			1	COVER, CRT anode & plate
				-	cover includes:
	200-0111-00			1	COVER
11	386-0647-00 337-0371-00			1	
12	337-0371-00			1 2	SHIELD, variable resistor RESISTOR, variable
				-	mounting hardware for each: (not included w/resistor)
	211-0017-00			2	
	210-0004-00			2	LOCKWASHER, internal, #4
	166-0169-00			2	TUBE, spacer, 9/16 inch
13				-	SEE DELAY LINE PAGE
14	200-0293-00	X140		1	· ·
	635-0427-00			1	
15	369-0015-00			1	assembly includes:
16	147-0022-00			1	FAN, 5 1/2 inches, clockwise MOTOR, 115 V
				-	mounting hardware: (not included w/motor)
	212-0022-00			2	SCREW, $8-32 \times 1 \frac{1}{2}$ inches, RHS
	166-0006-00			2	TUBE, spacer, 7/16 inch
	210-0008-00			2	LOCKWASHER, internal, #8
	210-0409-00			2	NUT, hex., 8-32 x 5/16 inch
17	426-0046-00			1	MOUNT, fan motor
	348-0008-00			3	mounting hardware: (not included w/mount) SHOCKMOUNT
	210-0008-00			6	LOCKWASHER, internal, #8
	210-0409-00			6	NUT, hex., 8-32 x 5/16 inch
18	354-0051-00			1	RING, fan
	213-0104-00			<u>-</u>	mounting hardware: (not included w/assembly)
	213-0104-00			6	SCREW, thread forming, $\#6 \times 3/8$ inch, THS
19	366-0032-00			1	•
	212 0004 00			-	knob includes:
	213 <b>-</b> 0004 <b>-</b> 00 355 <b>-</b> 0049 <b>-</b> 00			1 1	SCREW, set, 6-32 x 3/16 inch, HSS STUD, CRT rotator
20	406-0368-00	101	809	1	•
	406-0729-00	810	23,	1	•
				-	mounting hardware: (not included w/bracket)
	211-0507-00			2	
	210-0803-00			2	WASHER, flat, 0.150 ID x $3/8$ inch OD

LEFT SIDE (cont)

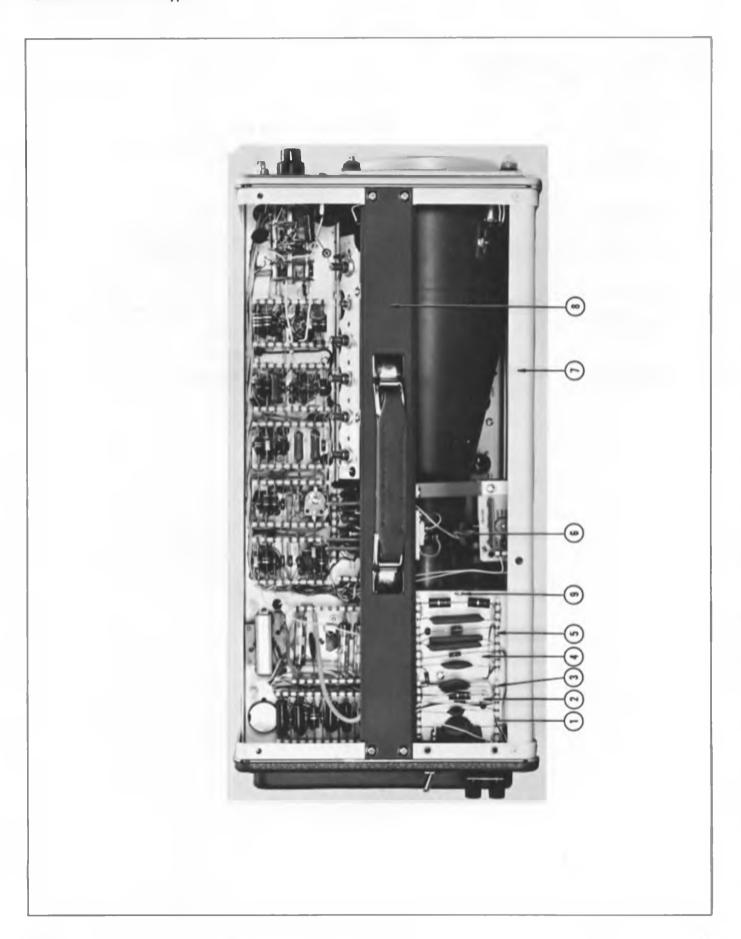
Fig. &				Q		
Index	Tektronix	Serial/Model No.		t	Description	
No.	Part No.	Eff	Disc	у	1 2 3 4 5	
21	354-0103-00			1	RING, plastic, clamping	
				_	ring includes:	
	210-0502-00			1	NUT, 10-32 x 3/8 inch	
22	211-0560-00			$\bar{1}$	SCREW, 6-32 x 1 inch, RHS	
	210-0407-00			1	NUT, hex., 6-32 x 1/4 inch	
23	136-0046-00	101	528	1	SOCKET, CRT	
	387-0344-00	101	528	1	PLATE, socket back	
	211-0038-00	101	528	2	SCREW, 4-40 x 5/16 inch, 100° csk, FHS	
	136-0046-00	529		1	ASSEMBLY, CRT socket	
				_	assembly includes:	
	136-0117-00			1	SOCKET, CRT	
	131-0178-00			7	CONNECTOR, cable end	
	387-0393-00			1	PLATE, back	
	213-0087-00			2	SCREW, thread cutting, 2-32 x 1/2 inch, RHS	



REF.	PART NO.	SERIAL/MODEL NO.		9		
		EFF.	DISC.	¥,	DESCRIPTION	
1	337-0373-00			1	SHIELD, calibrator	
				-	mounting hardware: (not included w/shield)	
	211-0507-00			2	SCREW, 6-32 x 5/16 inch, PHS	
2				1	TRANS FORMER	
				-	transformer includes:	
	212-0545-00			4	SCREW, 10-32 x 4 inch, HHS	
	210-0812-00			4	WASHER, fiber, #10	
				-	mounting hardware: (not included w/transformer)	
	210-0010-00			4	LOCKWASHER, internal, #10	
	210-0564-00			4	NUT, hex., 10-32 x 3/8 inch	
3	406-0578-00	101	200	1	BRACKET, transformer support, left	
	406-0579-00	101	200		BRACKET, transformer support, right	
	406-0656-00	201		1	BRACKET, transformer support	
				-	mounting hardware: (not included w/bracket)	
	211-0507-00			2	SCREW, 6-32 x 5/16 inch, PHS	
	211-0522-00				SCREW, 6-32 x 5/16 inch, 100° csk, FHS	
	210-0203-00				LUG, solder, SE #6 long	
	210-0457-00			2	NUT, keps, 6-32 x 5/16 inch	

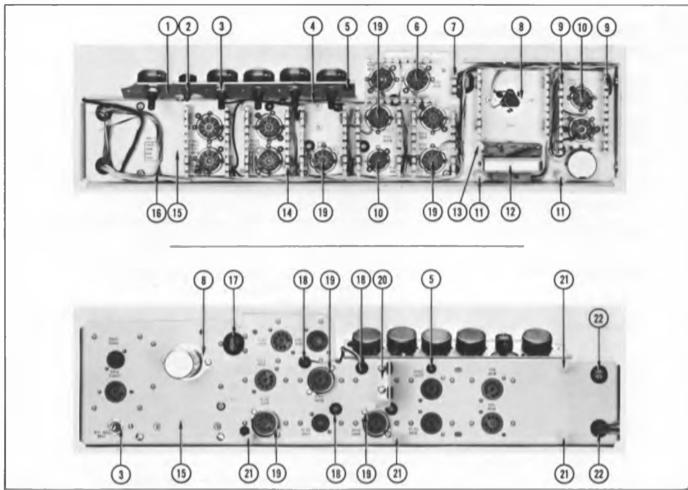
# BOTTOM (cont)

Fig. &			,,, , , , , , , , , , , , , , , , , ,	Q	
Index No.	Tektronix Part No.	Serial, Eff	/Model No. Disc	t y	Description
4	122-0085-00	101	200	1	ANGLE, frame, bottom rail, right
	122-0092-00	201		1	ANGLE, frame, bottom rail, right
				-	mounting hardware: (not included w/angle)
	212-0039-00			4	SCREW, $8-32 \times 3/8$ inch, THS
	210-0458-00			4	NUT, keps, $8-32 \times 11/32$ inch
5	179-0428-00			1	CABLE HARNESS, 110 V
6	381-0133-00			$\bar{1}$	BAR, support
				_	mounting hardware: (not included w/bar)
	212-0043-00			2	SCREW, 8-32 $\times$ 1/2 inch, 100° csk, FHS
7	122-0086-00			1	ANGLE, frame, bottom rail, left
				-	mounting hardware: (not included w/angle)
	212-0039-00			4	SCREW, $8-32 \times 3/8$ inch, THS
	210-0458-00			4	NUT, keps, 8-32 x 11/32 inch
8	384-0565-00			2	ROD, spacing, resistor mounting
	384-0194-00			2	ROD, extension (not shown)



TOP (cont)

Fig. & Index No.	Tektronix Part No.	Serial/A Eff	Model No. Disc	Q f y	Description
1 2	348-0002-00 210-0202-00			1	GROMMET, rubber, 1/4 inch diameter LUG, solder, SE #6
	211-0504-00 210-0407-00			1	mounting hardware: (not included w/lug) SCREW, 6-32 x 1/4 inch, PHS NUT, hex., 6-32 x 1/4 inch
3	124-0093-00	101	709	2	STRIP, ceramic, 3/4 inch h, w/5 notches each strip includes:
	355-0046-00			2	STUD, plastic
	361-0009-00			2	mounting hardware for each: (not included w/strip) SPACER, plastic, 0.406 inch long
	124-0095-00	710		2	STRIP, ceramic, 3/4 inch h, w/9 notches each strip includes:
	355-0046-00			2	STUD, plastic
	361-0008-00			2	mounting hardware for each: (not included w/strip) SPACER, plastic, 0.281 inch long
4	441-0318-00 441-0272-00	101 710	709		CHASSIS, focus & intensity CHASSIS, focus & intensity
				-	mounting hardware: (not included w/chassis)
	211-0559-00 211-0507-00			1 3	SCREW, 6-32 x 3/8 inch, 100° csk, FHS SCREW, 6-32 x 5/16 inch, PHS
	210-0803-00			3	
5	124-0095-00			2	STRIP, ceramic, 3/4 inch h, w/9 notches each strip includes:
	355-0046-00			2	STUD, plastic
	361-0009-00	101	709	- 2	mounting hardware for each: (not included w/strip) SPACER, plastic, 0.406 inch long
	361-0008-00		710	2	
6 7	131-0049-00 122-0060-00			5 1	CONNECTOR, cable end ANGLE, frame, top left
				-	mounting hardware: (not included w/angle)
	211-0559-00 210-0457-00				SCREW, $6-32 \times 3/8$ inch, $100^{\circ}$ csk, FHS NUT, keps, $6-32 \times 5/16$ inch
8	381-0176-00 381-0199-00	101 1316	1315		BAR, top support w/handle BAR, top support w/handle
		1310		-	mounting hardware: (not included w/bar)
	212-0039-00 381-0073-00				SCREW, 8-32 x 3/8 inch, THS BAR, retaining
9	337-0318-00				SHIELD, focus & intensity (not shown)
	211-0559-00			2	
	211-0507-00				SCREW, $6-32 \times 5/16$ inch, PHS
	210-0803-00			Ţ	WASHER, flat, 0.150 ID x $3/8$ inch OD



REF. NO.	PART NO.	SIRIAL/MC	DISC.	Q 1 Y.	DESCRIPTION
1	406-0581-00			1 -	BRACKET, variable resistor mounting hardware: (not included w/bracket)
	211-0507-00			2	SCREW, 6-32 x 5/16 inch, PHS
	210-0803-00			2	WASHER, flat, 0.150 ID x 3/8 inch OD
2				1	RESISTOR, variable
				-	mounting hardware: (not included w/resistor)
	210-0011-00			1	LOCKWASHER, internal, 1/4 ID x 15/32 inch OD
	210-0465-00			1	NUT, hex., 1/4-32 x 3/8 inch
3				6	RESISTOR, variable
				-	mounting hardware for each: (not included w/resistor)
	210-0207-00 210-0840-00			1	LUC, solder, 3/8 ID x 5/8 inch OD, SE
	210-0413-00			li	WASHER, flat, 0.390 ID x 9/16 inch OD NUT, hex., 3/8-32 x 1/2 inch
				ı.	101, 11ex., 370-32 x 1/2 11lell
4	179-0455-00			1	CABLE HARNESS
5	348-0002-00			6	GROMMET, rubber, 1/4 inch diameter

#### SWEEP CHASSIS (cont)

Fig. & Index No.	Tektronix Part No.	Serial/Model No. Eff Disc	Q t y_	Description 1 2 3 4 5
6	136-0015-00		8	SOCKET, tube, 9 pin, w/ground lugs
Ū			-	mounting hardware for each: (not included w/socket)
	213-0044-00		2	SCREW, thread forming, 5-32 x 3/16 inch, PHS
7	124-0088-00		1	STRIP, ceramic, 3/4 inch h, w/4 notches
			-	strip includes:
	355-0046-00		2	, .
	361-0009-00		2	mounting hardware: (not included w/strip) SPACER, plastic, 0.406 inch long
8			1	CAPACITOR
-				mounting hardware: (not included w/capacitor)
	386-0253-00			PLATE, metal
	211 <b>-</b> 0534 <b>-</b> 00 210 <b>-</b> 0006 <b>-</b> 00			SCREW, sems, 6-32 x 5/16 inch, PHS LOCKWASHER, internal, #6
	210-0407-00			NUT, hex., 6-32 x 1/4 inch
9	124-0091-00		2	STRIP, ceramic, 3/4 inch h, w/11 notches
			-	each strip includes:
	355 <b>-0</b> 046 <b>-</b> 00		2	SPACER, plastic mounting hardware for each: (not included w/strip)
	361-0009-00		2	SPACER, plastic, 0.406 inch long
10	136-0008-00		2	, , , , , , ,
			-	mounting hardware for each: (not included w/socket)
	213-0044-00		2	SCREW, thread forming, 5-32 x 3/16 inch, PHS
1 <b>1</b>	385-0060-00		2	ROD, plastic
	211-0507-00		1	mounting hardware for each: (not included w/rod) SCREW, 6-32 x 5/16 inch, PHS
	337-0370-00		1	SHIELD, high voltage (not shown)
			-	mounting hardware: (not included w/shield)
	211-0503-00			SCREW, 6-32 x 3/16 inch, PHS
	210-0803-00		3	WASHER, flat, 0.150 ID $\times$ 3/8 inch OD
	214-0210-00	X399	1	ASSEMBLY, solder spool
	214-0209-00		- 1	assembly includes: SPOOL, w/o solder
			-	mounting hardware: (not included w/assembly)
	361-0007-00		1	SPACER, plastic, 0.188 inch long
<b>1</b> 2			1	TRANSFORMER
	0/4 0001 00		-	mounting hardware: (not included w/transformer)
	346-0001-00 210-0004-00		1 2	STRAP, mounting LOCKWASHER, internal, #4
			_	LOCKWASHER, internal, #4
	210-0406-00		2	NUT, hex., 4-40 x 3/16 inch

#### SWEEP CHASSIS (cont)

Fig. & Index No.	Tektronix Part No.	Serial/ Eff	Model No. Disc	Q t	Description
140.	ruii ivo.		Disc	у	1 2 3 4 5
13	385-0076-00 385-0138-00	101 340	339		ROD, plastic ROD, plastic
	211-0507-00 213-0041-00	101 340	339		mounting hardware: (not included w/rod) SCREW, 6-32 x 5/16 inch, PHS SCREW, thread cutting, 6-32 x 3/8 inch, THS
14	124-0090-00  355-0046-00  361-0009-00			2	STRIP, ceramic, 3/4 inch h, w/9 notches each strip includes: STUD, plastic mounting hardware for each: (not included w/strip) SPACER, plastic, 0.406 inch long
15	441-0319-00 441-0319-01  212-0040-00 210-0458-00	101 4050	. 4049	1 -	CHASSIS, sweep CHASSIS, sweep mounting hardware: (not included w/chassis) SCREW, 8-32 x 3/8 inch, 100° csk, FHS NUT, keps, 8-32 x 11/32 inch
16 17 18 19	179-0427-00 179-0569-00 348-0012-00 348-0003-00 136-0015-00 	101 630	629		CABLE HARNESS, sweep GROMMET, rubber, 5/8 inch diameter GROMMET, rubber, 5/16 inch diameter SOCKET, tube, 9 pin, w/ground lugs mounting hardware for each: (not included w/socket) SCREW, sems, 4-40 x 5/16 inch, PHS LOCKWASHER, internal, #4 NUT, hex., 4-40 x 3/16 inch
20	406-0218-00 			2 2	BRACKET, horizontal display switch mounting hardware: (not included w/bracket) SCREW, 6-32 x 5/16 inch, PHS LOCKWASHER, internal, #6 NUT, hex., 6-32 x 1/4 inch
21	385-0096-00 385-0136-00  211-0507-00 213-0041-00	101 340 101 340	339 339		SCREW, 6-32 x 5/16 inch, PHS
22	348-0005-00			2	GROMMET, rubber, 1/2 inch diameter

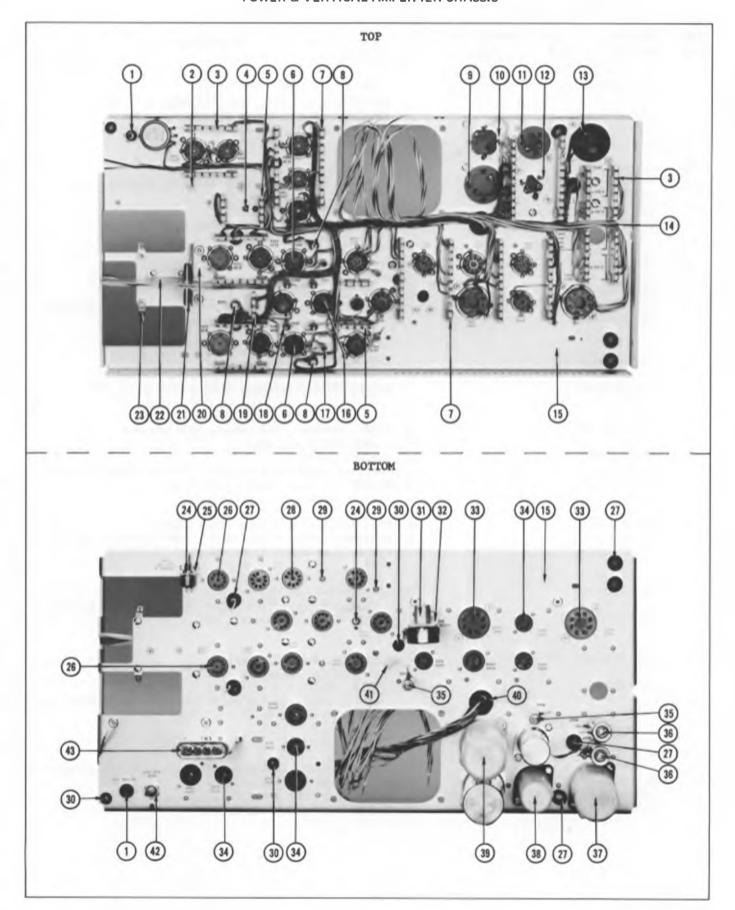


Fig. & Index	Tektronix	Serial/Model No.	Q t	
No.	Part No.	Eff Disc		Description 1 2 3 4 5
				12343
1	136-0037-00		1	SOCKET, tip jack
_				mounting hardware: (not included w/socket)
	210-0840-00			WASHER, flat, 0.390 ID x 9/16 inch OD
	210-0413-00		1	NUT, hex., $3/8-32 \times 1/2$ inch
2				CAPACITOR
	210-0407-00			mounting hardware: (not included w/capacitor) NUT, hex., 6-32 x 1/4 inch
	210-0006-00			LOCKWASHER, internal, #6
	210-0803-00			WASHER, flat, 0.150 ID $\times$ 3/8 inch OD
3	124-0089-00		10	STRIP, ceramic, 3/4 inch h, w/7 notches
			-	each strip includes:
	355-0046-00		2	STUD, plastic mounting hardware for each: (not included w/strip)
	361-0009-00		2	
4	210-0201-00		1	LUG, solder, SE #4
7				mounting hardware: (not included w/lug)
	213-0044-00			SCREW, thread forming, 5-32 x 3/16 inch, PHS
5	124-0088-00		4	STRIP, ceramic, 3/4 inch h, w/4 notches
			-	each strip includes:
	355-0046-00		2	STUD, plastic mounting hardware for each: (not included w/strip)
	361-0009-00			SPACER, plastic, 0.406 inch long
				, , ,
6	385-0107-00		2	ROD, plastic, 3/4 inch (not shown)
Ū				mounting hardware for each: (not included w/rod)
	211-0014-00		1	SCREW, $4-40 \times 1/2$ inch, PHS
	210-0201-00		2	LUG, solder, SE #4
	211-0011-00		- 1	mounting hardware for each: (not included w/lug)
	211-0011-00		1	SCREW, 4-40 x 5/16 inch, PHS
7	124-0091-00		8	STRIP, ceramic, 3/4 inch h, w/11 notches
	355-0046-00		2	each strip includes: STUD, plastic
			-	mounting hardware for each: (not included w/strip)
	361-0009-00		2	SPACER, plastic, 0.406 inch long
8			3	RESISTOR
	211-0544-00		- 1	mounting hardware for each: (not included w/resistor)
	211-0344-00			SCREW, 6-32 x 3/4 inch, THS NUT, hex., 5/16 x 21/32 inch long
	211-0507-00		1	
				SCREW, 6-32 x 5/16 inch, PHS

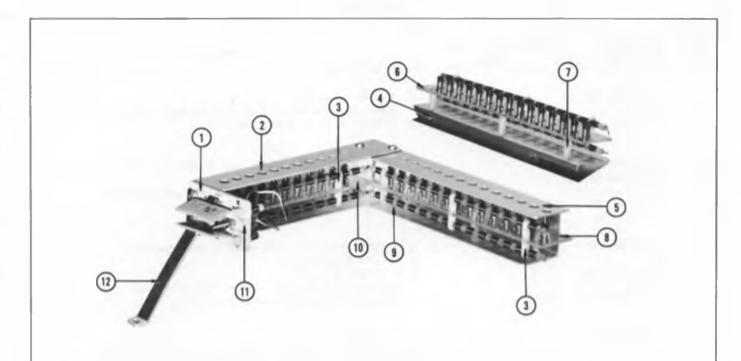
F: 0	POWER & VERTICAL AMPLIPIER CHASSIS (conf)									
Fig. &	Talles of	C: 1 / /	AA - 1   N.1.	Q						
Index	Tektronix		Model No.	t	Description					
No.	Part No.	Eff	Disc	у	1 2 3 4 5					
9				1	CADACTTOD					
,				1	CAPACITOR mounting hardware: (not included w/capacitor)					
	386-0254-00			1	PLATE, fiber, large					
	432-0044-00	x1600	3449x	1						
	211-0543-00	101	1599		SCREW, 6-32 x 5/16 inch, RHS					
	211-0514-00	1600	3449		SCREW, 6-32 x 3/4 inch, PHS					
	211-0543-00	3450			SCREW, 6-32 x 5/16 inch, RHS					
	210-0006-00			2	LOCKWASHER, internal, #6					
	210-0407-00			2	NUT, hex., $6-32 \times 1/4$ inch					
10				1	CAPACITOR					
				_						
	386-0255-00			1	PLATE, metal					
	211-0534-00			2	SCREW, sems, $6-32 \times 5/16$ inch, PHS					
	210-0006-00				LOCKWASHER, internal, #6					
	210-0407-00			2	NUT, hex., 6-32 x 1/4 inch					
11				1	CAPACITOR					
_				-	mounting hardware: (not included w/capacitor)					
	386-0252-00			1	PLATE, fiber					
	211-0534-00				SCREW, sems, 6-32 x 5/16 inch, PHS					
	210-0006-00				LOCKWASHER, internal, #6					
	210-0407-00			2	NUT, hex., 6-32 x 1/4 inch					
12				1	CAPACITOR					
				-	mounting hardware: (not included w/capacitor)					
	386-0253-00				PLATE, metal					
	211-0534-00				SCREW, sems, $6-32 \times 5/16$ inch, PHS					
	210-0006-00			2						
	210-0407-00			2	NUT, hex., $6-32 \times 1/4$ inch					
13				1	CAPACITOR					
				-	mounting hardware: (not included w/capacitor)					
	386-0254-00			1	,					
	211 <b>-</b> 0543 <b>-</b> 00 210 <b>-</b> 0006 <b>-</b> 00				SCREW, 6-32 x 5/16 inch, RHS					
	210-0407-00				LOCKWASHER, internal, #6 NUT, hex., 6-32 x 1/4 inch					
	210-0407-00			2	No1, nex., 0-32 x 1/4 Inch					
14	179-0426-00			1	CABLE HARNESS, power					
15	441-0320-00	101	4329		CHASSIS, power & vertical amplifier					
	441-0320-01	4330			CHASSIS, power & vertical amplifier					
	010 0040 00			-	mounting hardware: (not included w/chassis)					
	212-0040-00				SCREW, 8-32 x 3/8 inch, 100° csk, FHS					
	210-0458-00			8	NUT, keps, 8-32 x 11/32 inch					
16	385-0107-00			1	ROD, plastic, 3/4 inch (not shown)					
	011 001/ 00			-	mounting hardware: (not included w/rod)					
	211-0014-00			1	SCREW, $4-40 \times 1/2$ inch, PHS					

Fig. & Index	Tektronix	Seriul/Model No.	Q †	Description
No.	Part No.	Eff Disc	у	1 2 3 4 5
	210-0204-00			LUG, solder, DE #6
				mounting hardware: (not included w/lug)
	211-0011-00		1	SCREW, 4-40 x 5/16 inch, PHS
	274 0504 00		2	CORE, iron, threaded, 10-32 x 5/16 inch
17 18	276 <b>-</b> 0506 <b>-</b> 00 384 <b>-</b> 0542 <b>-</b> 00			ROD, plastic, capacitor mounting
10				mounting hardware for each: (not included w/rod)
	211-0507-00		1	SCREW, 6-32 x 5/16 inch, PHS
19	124-0087-00		1	STRIP, ceramic, 3/4 inch h, w/3 notches
			-	strip includes:
	355-0046-00		1	STUD, plastic
	0.41 0.000 0.0		- 1	
	361-0009-00		1	SPACER, prastic, 0.400 inch
20	406-0577-00		1	BRACKET, support
			-	mounting hardware: (not included w/bracket)
	211-0507-00		2	SCREW, 6-32 x 5/16 inch, PHS
21	348-0005-00		2	GROMMET, rubber, 1/2 inch
22	337-0372-00		1	SHIELD, polarity
			-	mounting hardware: (not included w/shield)
	211-0008-00			SCREW, 4-40 x 1/4 inch, PHS
	210-0004-00 210-0406-00		2	LOCKWASHER, internal, #4 NUT, hex., 4-40 x 3/16 inch
	210-0400-00		2	NOT, Hex., 4-40 X 3/10 Inch
23	210-0202-00			LUG, solder, SE #6
	211 0507 00			mounting hardware for each: (not included w/lug)
	211-0504-00 210-0407-00			SCREW, 6-32 x 1/4 inch, PHS NUT, hex., 6-32 x 1/4 inch
	210-0407-00		1	NUI, nex., 6-32 x 1/4 Inch
24			1	RESISTOR, variable
	010 0011 00		-	mounting hardware: (not included w/resistor)
	210-0011-00 210-0465-00		1	LOCKWASHER, internal, 1/4 ID x 15/32 inch OD
	210-0403-00		1	NUT, hex., 1/4-32 x 3/8 inch
25	406-0576-00		1	BRACKET, variable resistor
	011 0000 00		-	mounting hardware: (not included w/bracket)
	211-0008-00 210-0004-00			SCREW, 4-40 x 1/4 inch, PHS
	210-0004-00			LOCKWASHER, internal, #4 NUT, hex., 4-40 x $3/16$ inch
	4-4			
26	136-0015-00	101 1699		SOCKET, tube, 9 pin, w/ground lugs
	136-0145-00	1700	2	SOCKET, tube, 9 pin, w/ground lugs
	213-0044-00		2	mounting hardware for each: (not included w/socket) SCREW, thread forming, 5-32 x 3/16 inch, PHS

Fig. & Index No.	Tektronix Part No.	Serial/Model No. Eff Disc	Q t y	Description
27 28	348-0044-00 136-0015-00  213-0044-00		13 -	
29			-	COIL mounting hardware for each: (not included w/coil) SCREW, thread cutting, 4-40 x 1/4 inch, PHS
30 31	348-0003-00 		1 - 1	GROMMET, rubber, 5/16 inch diameter RESISTOR, variable mounting hardware: (not included w/resistor) WASHER, flat, 0.390 ID x 9/16 inch OD NUT, hex., 3/8-32 x 1/2 inch
32	406-0024-00  211-0507-00 210-0006-00 210-0407-00		- 2	BRACKET, variable resistor mounting hardware: (not included w/bracket) SCREW, 6-32 x 5/16 inch, PHS LOCKWASHER, internal, #6 NUT, hex., 6-32 x 1/4 inch
33	136-0011-00  211-0538-00 210-0006-00 210-0407-00		2 - 2 2 2	SOCKET, tube, 8 pin mounting hardware for each: (not included w/socket) SCREW, 6-32 x 5/16 inch, 100° csk, FHS LOCKWASHER, internal, #6 NUT, hex., 6-32 x 1/4 inch
34	136-0008-00  213-0044-00		5 - 2	mounting hardware for each: (not included w/socket)
35	211-0553-00 210-0601-00 210-0478-00 211-0507-00		1	RESISTOR mounting hardware for each: (not included w/resistor) SCREW, 6-32 x 1 1/2 inches, RHS EYELET NUT, hex., 5/16 x 21/32 inch long SCREW, 6-32 x 5/16 inch, PHS
36	212-0037-00 210-0809-00 210-0808-00 210-0008-00 210-0462-00 212-0004-00	101 5169 5170	1 1	RESISTOR mounting hardware for each: (not included w/resistor) SCREW, 8-32 x 1 3/4 inches, FIL HS WASHER, centering WASHER, centering LOCKWASHER, internal, #8 NUT, hex., 8-32 x 1/2 x 23/64 inch SCREW, 8-32 x 5/16 inch, PHS

Fig. & Index No.	Tektronix Part No.	Serial// Eff	Model No. Disc	Q t y	Description
37	200-0258-00			1	COVER, capacitor
38	200-0257-00			1	COVER, capacitor
39	200-0261-00	101	1599	1	COVER, capacitor, 4 1/16 inches
	200-0259-00	1600	3199	1	COVER, capacitor, 3 9/16 inches
	200-0293-00	3200	3449	1	COVER, capacitor, 2 9/16 inches
	200-0258-00	3450		1	COVER, capacitor, 3 1/32 inches
40	348-0006-00	101	5069	1	GROMMET, rubber, 3/4 inch diameter
	348-0050-00	5070		1	GROMMET, plastic, 3/4 inch diameter
41	385-0092-00	101	339	1	ROD, plastic
	385-0137-00	340		1	ROD, plastic
				-	mounting hardware: (not included w/rod)
	211-0507-00	101	339	1	SCREW, $6-32 \times 5/16$ inch, PHS
	213-0041-00	340		1	SCREW, thread cutting, 6-32 x 3/8 inch, THS
42				1	RESISTOR, variable
				-	mounting hardware: (not included w/resistor)
	210-0840-00			1	WASHER, flat, 0.390 ID x 9/16 inch OD
	210-0413-00				NUT, hex., 3/8-32 x 1/2 inch
43				1	CAPACITOR
45				-	capacitor includes:
	407-0277-00	X3930		1	BRACKET, capacitor
	124-0187-00	X3930		1	STRIP, ceramic, 7/16 inch h, w/5 notches
	124-0107-00	N3)30		_	strip includes:
	355-0046-00			2	STUD, plastic
	124-0187-01	X3930		1	STRIP, ceramic, 7/16 inch h, w/5 notches & silver band
				_	strip includes:
	355-0046-00			2	STUD, plastic
	361-0007-00	X3930		4	SPACER, plastic, 0.188 inch long
					· · · · · · · · · · · · · · · · · · ·

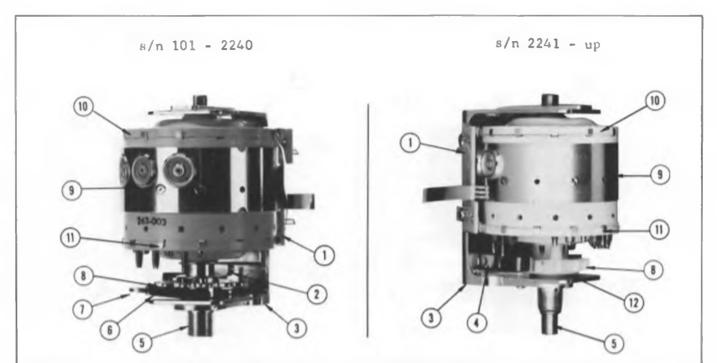
## **DELAY LINE**



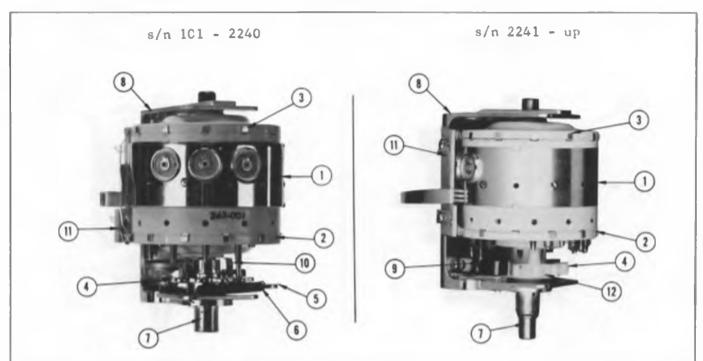
REF. NO.	PART NO.	SERIAL/MODEL NO.					
		CFF.	DISC.	٧,	DESCRIPTION		
1	386-0488-00			1	PLATE, white plastic		
					mounting hardware: (not included w/plate)		
	211-0510-00				SCREW, 6-32 x 3/8 inch, PHS		
	210-0803-00			2	WASHER, flat, 0.150 ID x 3/8 inch OD		
2	441-0131-00			1	CHASSIS		
				-	mounting hardware: (not included w/chassis)		
	211-0504-00			4	SCREW, 6-32 x 1/4 inch, PHS		
3	352-0016-00			8	HOLDER, coil form, w/o pin		
	352-0017-00			8	HOLDER, coil form, w/pin		
4	337-0368-00			1	SHIELD		
				-	mounting hardware: (not included w/shield)		
	211-0507-00			2	SCREW, 6-32 x 5/16 inch, PHS		
5	441-0129-00			1	CHASSIS		
				-	mounting hardware: (not included w/chassis)		
	211-0507-00			3	SCREW, 6-32 x 5/16 inch, PHS		
6	386-0487-00			1	PLATE, plastic		
					mounting hardware: (not included w/plate)		
	211-0011-00			3	SCREW, 4-40 x 5/16, PHS		
7	384-0531-00			8	ROD, spacing, plastic		
					mounting hardware for each: (not included w/rod)		
	211-0011-00			1	SCREW, 4-40 x 5/16 inch, PHS		

#### DELAY LINE

Fig. & Index	Tektronix	Serial/	Model No.	Q †	
No.	Part No.	Eff	Disc	у_	Description
8	386-0486-00				DIAMP. 1
0	300-0400-00			T	PLATE, plastic
	011 0011 00			_	mounting hardware: (not included w/plate)
	211-0011-00			2	SCREW, 4-40 x 5/16, PHS
9	348-0002-00			1	GROMMET, rubber, 1/4 inch diameter (not shown)
10	386-0485-00				PLATE, plastic
LU	300-0403-00			7	* •
	211-0011-00			2	mounting hardware: (not included w/plate)
	211-0011-00			3	SCREW, 4-40 x 5/16 inch, PHS
11	124-0086-00	X140		1	STRIP, ceramic, 3/4 inch h, w/2 notches
				_	strip includes:
	355-0046-00			1	STUD, plastic
				_	mounting hardware: (not included w/strip)
	361-0009-00	X140		1	SPACER, plastic, 0.406 inch long
10	106 0000 00				DD LOVER
12	406-0220-00			1	BRACKET, support
	011 0504 00			.=	
	211-0504-00			1	SCREW, 6-32 x 1/4 inch, PHS
	210-0803-00			1	WASHER, flat, 0.150 ID $\times$ 3/8 inch OD

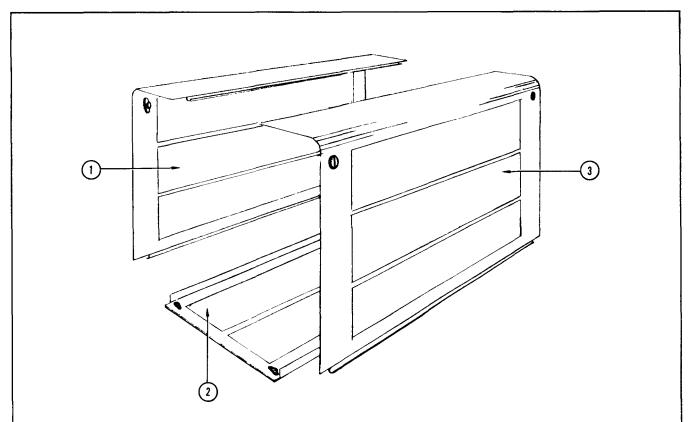


REF.		SERIAL/MODEL NO.		0	
NO.	PART NO.	EFF.	DISC.	Y.	DESCRIPTION
	263-0003-00			1	SWITCH, wiredVOLTS/DIV CHANNEL "B"
				-	switch includes:
1	387-0421-00	101	2059	1	PLATE, turret contact
	387-0692-00	2060		1	PLATE, turret contact
				-	mounting hardware: (not included w/plate)
	211-0007-00			2	SCREW, 4-40 x 3/16 inch, PHS
	210-0003-00			2	LOCKWASHER, external, #4
2	214-0134-00	101	2240X	1	SPRING, thrust
3	426-0076-00	101	2059	1	FRAME, turret
	426-0189-00	2060	2240	1	FRAME, turret
	426-0195-00	2241		1	FRAME, turret
				-	frame includes:
4	214-0324-00	X2241		1	SPRING, detent
	211-0097-00	X2241		1	SCREW, 4-40 x 5/16 inch, PHS
	210-0589-00	X2241		1	NUT, locking, 4-40 x 1/4 inch
5	384-0235-00			1	ROD, shaft
5	214-0088-00	101	2240X	1	SPRING, extension
7	214-0189-00	101	2240X	1	CARRIER, detent
				-	mounting hardware: (not included w/carrier)
	211-0563-00			1	SCREW, 6-32 x 9/32 inch, CRS shoulder
В	401-0017-00	101	2059	1	WHEEL, detent
	401-0025-00	2241		1	WHEEL, detent
	213-0022-00			1	SCREW, set, 4-40 x 3/16 inch, HSS
9	204-0068-00	101	2059	1	BODY, wired turret assembly
	204-0129-00	2060		1	BODY, wired turret assembly
				-	body includes:
)	200-0191-00			2	CAP, end
L	204-0024-00	101	2059	18	BODY, capacitor barrel
	204-0127-00	2060		18	BODY
2	354-0206-00	X2241		1	RING, stop, detent
				-	mounting hardware: (not included w/switch)
	210-0419-00			1	NUT, hex., 3/8-32 x 1/2 inch



REF.		SERIAL/MODEL NO.					
	PART NO.	EFF.	DISC.	Y	DESCRIPTION		
	263-0001-00			1	SWITCH, wired VOLTS/DIV, CHANNEL "A"		
				-	switch includes:		
1	204-0068-00	101	2059	1	BODY, wired turret assembly		
_	204-0129-00	2060		1	BODY, wired turret assembly		
				-	body includes:		
2	200-0191-00			2	CAP, end		
3	204-0024-00	101	2059	18	BODY, capacitor barrel		
	204-0127-00	2060		18	BODY, capacitor barrel		
4	401-0017-00	101	2240	1	WHEEL, detent		
	401-0025-00	2241		1	Wileel, detent		
	213-0022-00			1	SCREW, set, 4-40 x 3/16 inch, HSS		
5	214-0189-00	101	2240X	1	CARRIER, detent assembly		
				-	mounting hardware: (not included w/carrier)		
	211-0563-00			1	SCREW, 6-32 x 9/32 inch, CRS shoulder		
6	214-0088-00	101	2240X	1	SPRING, extension		
7	384-0235-00			1	ROD, shaft		
8	426-0076-00	101	2059	1	FRAME, turret		
	426-0189-00	2060	2240	1	FRAME, turret		
	426-0195-00	2241		1	FRAME, turret		
				-	frame includes:		
9	214-0324-00	X2241		1	SPRING, detent		
	211-0097-00	X2241		1	SCREW, 4-40 x 5/16 inch, PHS		
	210-0589-00	X2241		1	NUT, locking, 4-40 x 1/4 inch		
10	214-0134-00	101	2240X	1	SPRING, thrust		
11	387-0420-00	101	2059	1	PLATE, turret contact		
	387-0691-00	2060		1	PLATE, turret contact		
				-	mounting hardware: (not included w/plate)		
	211-0007-00			2	SCREW, 4-40 x 3/16 inch, PHS		
	210-0003-00			2	LOCKWASHER, external, #4		
12	354-0206-00	X2241		1	RING, stop, detent		
				-	mounting hardware: (not included w/switch)		
	210-0419-00			1	NUT, hex., 3/8-32 x 1/2 inch		

## CABINET



REF.	SERIAL/MODE	L NO. Q			
NO. PART NO	O. EFF. D	DISC. Y.	DESCRIPTION		
1 387-0222:	-00 	1 -2 -1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	PLATE, cabinet side, left plate includes: FASTENER, cabinet latch assembly fastener includes: STOP NUT, latch WASHER, plastic, 0.164 ID x 0.500 inch OD SCREW, 8-32 x 1/2 inch PLATE, cabinet bottom plate includes: FASTENER, cabinet latch assembly fastener includes: STOP NUT, latch WASHER, plastic, 0.164 ID x 0.500 inch OD SCREW, 8-32 x 1/2 inch PLATE, cabinet side, right plate includes: FASTENER, cabinet latch assembly fastener includes: STOP NUT, latch WASHER, plastic, 0.164 ID x 0.500 inch OD SCREW, 8-32 x 1/2 inch		

## STANDARD ACCESSORIES

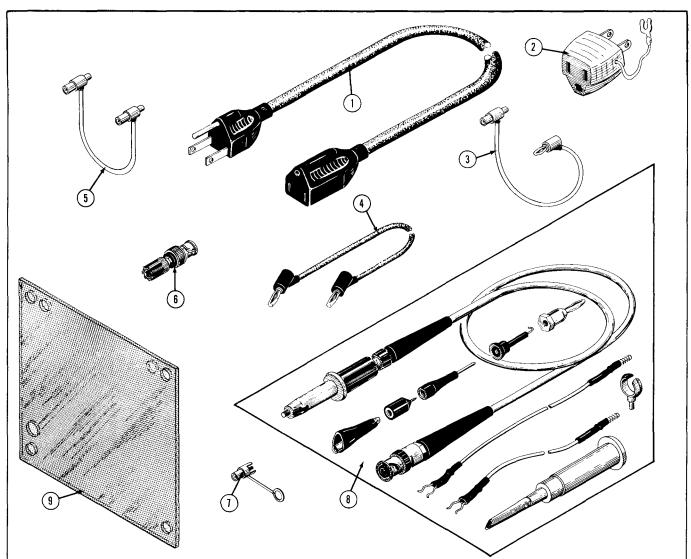


Fig. &			Q				
Index	Tektronix		Nodel No.	_	Description		
No.	Part No.	Eff	Disc	У	1 2 3 4 5		
1 2 3 4 5 6 7 8 9	161-0010-00 161-0010-03 103-0013-00 012-0091-00 012-0087-00 013-0004-00 103-0033-00 012-0092-00 010-0127-00 378-0514-00 378-0567-00 070-0225-01	101 2030 23050 22270	3049 2029 3049x 3519	1 1 1 1 1 2 2 1 1 2 2 1 1 2 2	CORD, power, 16 guage, 8 foot, 3 wire CORD, power, 16 guage, 8 foot, 3 wire ADAPTER, power cord, 3 to 2 wire CORD, patch, BNC to banana, red, 18 inches long CORD, patch-banana, 18 inches, red CORD, patch, BNC to BNC, red, 18 inches long ADAPTER, binding post ADAPTER, binding post JACK, BNC PROBE, P6006, 10 MEG, 10X, 42 inches, BNC FILTER, light, plastic, 5 inches, green, w/cam hole FILTER, light, smoke gray MANUAL, instruction (not shown)		

## PARTS LIST ABBREVIATIONS

внв	binding head brass	int	internal
BHS	binding head steel	lg	length or long
cap.	capacitor	met.	metal
cer	ceramic	mtg hdw	mounting hardware
comp	composition	OD	outside diameter
conn	connector	OHB	oval head brass
CRT	cathode-ray tube	OHS	oval head steel
csk	countersunk	РНВ	pan head brass
DE	double end	PHS	pan head steel
dia	diameter	plstc	plastic
div	division	PMC	paper, metal cased
elect.	electrolytic	poly	polystyrene
EMC	electrolytic, metal cased	prec	precision
EMT	electrolytic, metal tubular	PT	paper, tubular
ext	external	PTM	paper or plastic, tubular, molded
F & 1	focus and intensity	RHB	round head brass
FHB	flat head brass	RHS	round head steel
FHS	flat head steel	SE	single end
Fil HB	fillister head brass	SN or S/N	serial number
Fil HS	fillister head steel	SW	switch
h	height or high	TC	temperature compensated
hex.	hexagonal	THB	truss head brass
ННВ	hex head brass	thk	thick
HHS	hex head steel	THS	truss head steel
HSB	hex socket brass	tub.	tubular
HSS	hex socket steel	var	variable
ID	inside diameter	w	wide or width
incd	incandescent	WW	wire-wound

#### PARTS ORDERING INFORMATION

Replacement parts are available from or through your local Tektronix, Inc. Field Office or representative.

Changes to Tektronix instruments are sometimes made to accommodate improved components as they become available, and to give you the benefit of the latest circuit improvements developed in our engineering department. It is therefore important, when ordering parts, to include the following information in your order: Part number, instrument type or number, serial or model number, and modification number if applicable.

If a part you have ordered has been replaced with a new or improved part, your local Tektronix, Inc. Field Office or representative will contact you concerning any change in part number.

#### SPECIAL NOTES AND SYMBOLS

×000	Part first added at this serial number
00×	Part removed after this serial number
*000-0000-00	Asterisk preceding Tektronix Part Number indicates manufactured by or for Tektronix, Inc., or reworked or checked components.
Use 000-0000-00	Part number indicated is direct replacement.
0	Screwdriver adjustment.
	Control, adjustment or connector.

# SECTION 8 ELECTRICAL PARTS LIST

Values fixed unless marked variable.

Q.,	lhe
DU	105

Ckt. No.	S/N Range	Descripition	Tektronix Part Number
B160W	101-3619	NE23	Use 150-027
B160W	3620-up	NE-2V	150-0030-00
B167	3020-up	NE-2 <b>V</b> NE23	Use 150-027
B171		NE23 NE23	Use 150-027
B346	101-3619	NE23	Use 150-027
B346	3620-up	NE-2V	150-0030-00
B364		NE23	Use 150-027
B384		NE23	Use 150-027
B442	101-361 <b>9</b>	NE23	Use 150-027
B442	3620-up	NE-2V	150-0030-00
B493	•	NE23	Use 150-027
B494		NE23	Use 150-027
B542	101-3619	NE23	Use 150-027
B542	3620-up	NE-2V	150-0030-00
B590	X3150-3340X	NE23	150-0027-00
B594	X3150- <b>3</b> 340X	NE23	150-0027-00
B601		Graticule Light	150-001
B602		Graticule Light	150-001
B603		Pilot Light Assembly	136-031
		<b>~</b> · · · · · · · · · · · · · · · · · · ·	

## Capacitors

Tolerance	<b>±20%</b>	uniess	otherwise	indicated.
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C1 C2 C10 C11 C15 C21	X1959-up X27 <b>0</b> -up	4.7 pf .1 μf .01 μf 100 pf .001 μf .01 μf	Cer. Cer. MT Cer. Discap MT		500 v 500 v 400 v 350 v 500 v	±1 pf	281-501 283-008 285-510 281-523 283-000 283-002
C24 C25 C31 C37 C44 C47		100 pf .001 μf .01 μf 18 pf 27 pf .005 μf	Cer. Discap Hi-Kap Cer. Cer. Discap		350 v 500 v 150 v 500 v 500 v	10% ±5.4 pf	281-523 283-000 283-003 281-542 281-513 283-001
C134 C138 C141 C149 C150		8 pf .01 μf 5.6 pf .005 μf 56 pf	Cer. Discap Cer. Discap Cer.		500 v 500 v 500 v 500 v 500 v	±0.5 pf ±10% ±5.6 pf	281-503 283-002 281-544 283-001 281-521
C160A C160B C160C C160D C160E		3-12 pf 82 pf 4.5-25 pf 82 pf 4.5-25 pf	Cer. Mica Cer. Mica Cer.	Var Var. Var.	500 v 500 v	5% 5%	281-007 283-534 281-010 283-534 281-010

®2 8-1

## Capacitors (cont)

Ckt. No.	S/N Range		Descripitio	n			Tektronix Part Number
C160F	101-3929	.001 μf	Mylar			±1/2%	*291-008
C160G C160H C160J	101-3929	$\begin{array}{c} 0.01 \ \mu \mathrm{f} \\ 0.1 \ \mu \mathrm{f} \\ 0.1 \ \mu \mathrm{f} \end{array}$	-	ning Series		±1/2%	*291-007
C160F C160G C160H C160J C165	3930-up	0.001 µf 0.01 µf 0.1 µf 1 µf 470 pf	Timing Co	apacitor Assemb	oly 500 v	±94 pf	*295-0095-00 281-525
C167 C180A C180B C180C C180D		.001 μf 180 μf .0022 μf .022 μf .1 μf	Discap Mica MT MT MT		500 v 500 v 400 v 400 v 400 v	10% 5%	283-000 283-509 285-543 285-515 285-526
C180E C181 C181 C190 C196 C224	101-476 <b>9</b> 4770-ир	.1 μf 39 pf 22 pf 12 pf .001 μf 3-12 pf	MT Cer. Cer. Cer. Discap Cer.	Var.	400 v 500 v 500 v 500 v 500 v	±3.9 pf 10% ±1.2 pf	285-526 281-516 281-0511-00 281-506 283-000 281-009
C260 C330 C340 C355 C364		9-180 pf 4.5-25 pf 15 pf 1.5 pf 6.8 pf	Mica Cer. Cer. Cer. Cer.	Var. Var.	500 v 500 v 500 v	±1.5 pf ±0.5 pf 10%	281-023 281-010 281-509 281-526 281-541
C381 C384 C400 C403 C404	101-569X	.01 μf 3-12 pf .1 μf .7-3 pf .005 μf	Discap Cer. MT Tub. Discap	Var. Var.	500 v 600 v 500 v		283-002 281-009 285-587 281-027 283-001
C420 C421 C422 C430 C431	X570-up X570-up X124-up X124-up	.7-3 pf .7-3 pf .1 μf 1.8 pf 1.8 pf	Tub. Tub. PTM Cer. Cer.	Var. Var.	200 v 500 v 500 v		281-027 281-027 285-572 281-557 281-557
C450 C451 C454 C457 C462		.7-3 pf .7-3 pf .005 μf .01 μf .001 μf	Tub. Tub. Discap Discap Discap	Var. Var.	500 v 500 v 500 v		281-027 281-027 283-001 283-002 283-000
C472 C480A-C484L C485 C486 C488A,B	Х140-ир	.001 μf .7-3 pf .7-3 pf 3-12 pf 2 x 40 μf	Discap Tub. Tub. Cer. EMC	Var. Var. Var.	500 v 350 v		283-000 (40) 281-037 281-037 281-031 Use 290-027
C491 C500 C503 C504	101-569X	.01 μf .01 μf .1 μf .7-3 pf .005 μf	D'scap Discap MT Tub. Discap	Var.	500 v 600 v		283-002 283-002 285-587 281-037 283-001



## Capacitors (conf)

Ckt. No.	S/N Range		Descripitio	n			Tektronix Part Number
C520 C521 C522 C530 C531	X570-up X570-up X124-up X124-up	.7-3 pf .7-3 pf .1 μf 1.8 pf 1.8 pf	Tub. Tub. PTM Cer. Cer.	Var. Var.	200 v 500 v 500 v		281-027 281-027 285-572 281-557 281-557
C550 C551 C571 C580		.7-3 pf .7-3 pf .01 μf 10 pf	Tub. Tub. Discap Cer.	Var. Var.	500 v 500 v	±1 pf	281-027 281-027 283-002 281-504
C581 C584 C585 C586 C588		47 pf .005 μf 10 pf 47 pf .005 μf	Cer. Discap Cer. Cer. Discap		500 v 500 v 500 v 500 v 500 v	±4.7 pf	281-519 283-001 281-504 281-519 283-001
C594 C597 C601 C610 C617		12 pf .01 μf 2 μf .01 μf .01 μf	Cer. Discap PMC PTM PTM		500 v 500 v 236 v 400 v 400 v	±1.2 pf	281-506 283-002 285-588 285-510 285-510
C627 C640 C642 C644 C650		2 × 20 μf 125 μf 2 × 125 μf 2 × 125 μf .01 μf	EMC EMC EMC EMC PTM		450 v 350 v 350 v 350 v 400 v		Use 290-010 Use 290-016 Use 290-130 Use 290-130 285-510
C666 <b>A,B</b> C670 C680 C681 C684		2 x 20 μf .01 μf .02 μf .02 μf .01 μf	EMC PTM Discap Discap Discap		450 v 400 v 600 v 600 v 500 v		Use 290-010 285-510 283-006 283-006 283-002
C801A,B C803 C807 C808 C821	101-709	2 × 20 μf .001 μf .001 μf .001 μf .015 μf	EMC MT MT MT Discap		450 v 600 v 1,000 v 1,000 v 2,500 v		Use 290-010 285-501 285-502 285-502 283-030
C821 C827A,B C827A,B C831 C831	710-up 101-709 710-up 101-709 710-up	.015 µf .01 µf .015 µf .01 µf .0068 µf	Discap Discap Discap Discap Discap		3,000 v 2,000 v 3,000 v 2,000 v 3,000 v		283-042 283-011 283-042 283-011 283-043
C832 C832 C841 C842 C842	101-169 170-up 101-709 710-up	.0068 μf .005 μf .047 μf .015 μf .0068 μf	PTM Cer. MT Discap Discap		3,000 v 4,000 v 400 v 2,500 v 3,000 v		Use 283-034 283-034 285-519 283-030 283-043
C844 C844 C846 C848 C848	101-269 270-up X710-up 101-709 <b>710</b> -up	.01 μf .001 μf .0068 μf .015 μf .015 μf	Discap Discap Discap Discap Discap		2,000 v 3,000 v 3,000 v 2500 v 3,000 v		Use 283-044 283-044 283-043 283-030 283-042

## Capacitors (cont)

Ckt. No.	S/N Range		Descripi	tion			Tektronix Part Number
C863 C864 C871 C874 C885		.01 μf .01 μf 330 pf 330 pf 27 pf	Discap Discap Mica Mica Cer.		500 v 500 v 500 v 500 v 500 v	10% 10%	283-002 283-002 283-518 283-518 281-513
			Fu	Ises			
F601			G Slo-Blo G Slo-Blo	117 V Operation 234 V Operation			159-026 159-003
			Indu	uctors			
LR149 LR149 L455 L460 L460	101-4709 4710-∪p 101-123 124-up	850 μh 1.2 mh 6.8-14.6 μh 1.4 μh .5 μh			Fixed Fixed Var. Fixed Fixed	core 27 <i>6</i>	*108-058 *108-0164-00 6-506 *114-080 108-095 *108-170
L464 L470 L470 L474 L480 <b>A,B</b>	1 <b>01-123</b> 124-up	7 μh 1.4 μh .5 μh 7 μh 15 Secti <b>on Del</b> ay	Line		Fixed Fixed Fixed Fixed		*108-137 108-095 *108-170 *108-138 *108-108
L481 A, <b>B</b> L482A,B L484A,B L485 L486		2.2 μh 14 Section Delay 11 Section Delay 7.3-16 μh 7.3-16 μh			Fixed Yar. Var.		*108-147 *108-107 *108-106 5-506 *114-054 5-506 *114-054
L555		6.8-14 <b>.6</b> μh			Var.	core 276	5-506 *114-080
			Pos	istors			
Resistors are	fixed Composition	, ±10% unless other					
R1 R2 R3 R4 R5	X1959-up	1 meg 560 k 50 k 100 k 10 meg	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w	Var.	Comp. Comp. DC Comp. Comp. Comp.	5% 5% Level Adj. 10% 10%	301-105 301-564 311-023 302-104 302-106
R13 R14 R15 R16 R17 <sup>1</sup>	X270-up	1 meg 100 k 470 k 100 k 100 k	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w	Var.	Comp. Comp. Comp. Comp. Trig.	10% 10% 10% 10% Level	302-105 302-104 302-474 302-104 311-099
R19 R20 R21 R22 R23		1 meg 3.9 meg 820 k 100 Ω 100 Ω	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	302-105 302-395 302-824 302-101 302-101

<sup>&</sup>lt;sup>1</sup>R17 concentric with R110, R349 & SW110, furnished as a unit.

Ckt. No.	S/N Range		Descripition				Tektronix Part Number
R24 R25 R26 R28 R29		3.9 k 27 Ω 3.9 k 39 k 39 k	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 2 w 2 w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	302-392 302-270 302-392 306-393 306-393
R30 R31 R32 R34 R35	101-399	100 k 100 Ω 220 k 1 k 2.2 k	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	302-104 302-101 302-224 302-102 Use 301-222
R35 R37 R38 R39 R40	400-up	2.2 k 150 k 120 k 100 k 2.2 meg	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w	Var.	Comp. Comp. Comp. Comp. Trig Comp.	5% 10% 10% Level Centering 10%	301-222 302-154 302-124 311-026 302-225
R41 R43 R44 R46 R47		$100~\Omega$ $820~\Omega$ $1~meg$ $22~k$ $500~\Omega$	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1 w .2 w	Var.	Comp. Comp. Comp. Comp. Comp. Trig	10% 10% 10% 10% Sensitivity	302-101 302-821 302-105 304-223 311-066
R48 R110 <sup>2</sup> R111 R114 R131		22 k 500 k 100 k 100 k 4.7 k	1 w 1/2 w 1/2 w	Var. Var.	Comp. Comp. Stal Comp. Pres Comp. Comp.		304-223 311-099 311-173 302-104 302-472
R132 R133 R134 R135 R138		100 Ω 3.6 k 3.6 k 100 Ω 100 Ω	1/2 w 1 w 1 w 1/2 w 1/2 w		Comp. Comp. Comp. Comp. Comp.	10% 5% 5% 10% 10%	302-101 303-362 303-362 302-101 302-101
R140 R141 R143 R144 R146		47 Ω 43 k 33 k 10 k 47 Ω	1/2 w 1 w 1 w 5 w 1/2 w		Comp. Comp. Comp. WW Comp.	10% 5% 5% 5% 10%	302-470 303-433 303-333 308-054 302-470
R147 R148 R150 R152 R160A	<b>X38</b> 70-up	1 k 150 k 680 Ω 6.8 meg 100 k	1/2 w 1 w 1/2 w 1/2 w 1/2 w		Comp. Comp. Comp. Comp. Prec.	10% 10% 10% 10% 1%	302-102 304-154 302-681 302-0685-00 309-045
R160B R160C R160D R160E R160F		200 k 500 k 1 meg 2 meg 5 meg	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w		Prec. Prec. Prec. Prec. Prec.	1% 1% 1% 1%	309-051 309-003 309-014 309-023 309-087

<sup>&</sup>lt;sup>2</sup>R110 concentric with R17, R349 & SW110, furnished as a unit.

## Electrical Parts List—Type 516

Ckt. No.	S/N Range		Description				Tektronix Part Number
R160G R160H R160J R160W R160X		10 meg 10 meg 10 meg 100 k 8.2 k	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w		Prec. Prec. Prec. Comp. Comp.	1% 1% 1% 10% 10%	309-095 309-095 309-095 302-104 302-822
R160Y R161 R165 R166 R167		20 k 100 Ω 47 k 47 k 1.5 meg	1/2 w 1 w 1 w 1/2 w	Var.	WW Time/I Comp. Comp. Comp. Comp.	Div. 10% 10% 10% 10%	311-108 302-101 304-473 304-473 302-155
R168 R171 R173 R174 R176		470 k 100 Ω 10 k 2.2 k 2 k	1/2 w 1/2 w 5 w 1 w	Var.	Comp. Comp. WW Comp. Comp. Sweep	10% 10% 5% 10% Length	302-474 302-101 308-054 304-222 311-008
R178 R180A R180B R181 R183		6 k 470 k 4.7 meg 4.7 meg 100 Ω	5 w 1/2 w 1/2 w 1/2 w 1/2 w 1/2 w		WW Comp. Comp. Comp. Comp.	5% 10% 10% 10% 10%	308-052 302-474 302-475 302-475 302-101
R184 R185 R190 R191 R192		39 k 27 k 12 k 22 k 47 Ω	1/2 w 1/2 w 1 k 2 w 1/2 w		Comp. Comp. Comp. Comp. Comp.	5% 5% 10% 10%	301-393 301-273 304-123 306-223 302-470
R194 <b>R196</b> R197 R198 R199		10 k 100 k 47 Ω 47 Ω 100 k	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10%	302-103 302-104 302-470 302-470 302-104
R225 R259 R330 R332 R333A,B		50 k 2.5 k 1.84 meg 1.5 meg 2 x 20 k	.1 w .5 w ½ w ½ w	Var. Var.	Comp. Sweep Comp. Mag. ( Prec. Prec. Comp. Horiz.	Gain Adjust 1% 1%	311-078 311-086 309-021 309-017 311-090
R340 R341 R345 R346 R347		560 Ω 47 Ω 47 k 100 k 120 k	Y <sub>2</sub> w Y <sub>2</sub> w 2 w Y <sub>2</sub> w Y <sub>2</sub> w		Comp. Comp. Comp. Comp. Prec.	10% 10% 10% 10%	302-561 302-470 306-473 302-104 309-091
R349 <sup>3</sup> R351 R353 R355 R356		100 k 100 Ω 68 k 400 k 250 k	1/2 w 2 w 1/2 w 1/2 w	Var.	Comp. Horiz. Comp. Comp. Prec. Prec.	Input Atten. 10% 10% 1% 1%	311-099 302-101 306-683 309-126 309-109

 $<sup>^3\</sup>mbox{R349}$  concentric with R17, R110 & SW110, furnished as a unit.

Ckt. No.	S/N Range		Description				Tektronix Part Number
R358 R361 R364 R366 R368		50 k 47 Ω 630 k 47 Ω 41.5 k	1/2 w 7 w 1/2 w 8 w	Var.	Comp. Comp. Mica Comp. WW	1%	311-023 302-470 *310-507 302-470 Use *310-614
R373 R376 R377 R380 R380	101-5 <b>28</b> 9 5290-up	15 k 15 k 10 k 22 k 24 k	1 w 1 w 5 w 1 w 1 w		Comp. Comp. WW Comp. Comp.	10% 5% 10%	304-153 304-153 308-054 304-223 303-0243-00
R381 R381 R382 R382 R383	101-5289 5290-up 101-5289 5290-up	5.6 k 5.6 k 100 k 100 k 47 Ω	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w 1/2 w		Comp. Comp. Comp. Comp. Comp.	5% 10 <b>%</b> 5%	302-562 301-0562-00 302-104 301-0104-00 302-470
R384 R386 R388 R403 R404		7—35 k 47 Ω 270 k 1 meg 100 k	7 w 1/2 w 1 w 1/2 w 1/2 w		Mica Comp. Comp. Prec. Comp.	1% 10% 10% 1% 1%	*310-524 302-470 304-274 309-014 302-104
R420 R421 R422 R423 R424		47 Ω 47 Ω 220 Ω 100 Ω 39 k	1/2 w 1/2 w 1/2 w 1/2 w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	302-470 302-470 302-221 302-101 304-393
R426 R427 R430 R432 R433		39 k 200 Ω 47 Ω 1 k 600 Ω	1 w '½ w 5 w 2 w	Var.	Comp. Comp. Comp. WW Mica Plate	10% D.C. Balance 10% 5% 1%	304-393 311-158 302-470 308-106 *310-567
R434 R437 R438 R439 R440		8.2 k 8.2 k 600 Ω 3.5 k 100 Ω	2 w 2 w 2 w 5 w 1/2 w	<b>V</b> ar.	Comp. Comp. Mica Plate WW Comp.	5% 5% 1% 5% Gain Set, Channel A	305-822 305-822 *310-567 308-080 311-169
R441 R442 R443 R444 R446		550 Ω 100 k 180 k 180 k 27 k	½ w ½ w ½ w ½ w	Var.	WW Comp. Comp. Comp.	Volts/Div 10% 10% 10% 10%	Use *311-287 302-104 302-184 302-184 302-273
R447 R448 R450 R451 R454		2 x 50 k 27 k 47 Ω 47 Ω 47 Ω	1/2 w 1/2 w 1/2 w 1/2 w	Var.	Comp. Comp. Comp. Comp.	Position 10% 10% 10% 10%	311-111 302-273 302-470 302-470 302-470

Ckt. No.	S/N Range		Descripition				Tektronix Part Number
R455 R456 R457 R458 R460		670 Ω 47 Ω 4.7 k 12 k 47 Ω	1/2 w 1/2 w 2 w 2 w 1/2 w		Prec. Comp. Comp. Comp. Comp.	1% 10% 10% 10% 10%	309-082 302-470 306-472 306-123 302-470
R462 R463 R468 R468 R470	101-106 107-up	100 k 3 k 3.3 k 2.2 k 47 Ω	1/2 w 5 w 2 w 1 w 1/2 w		Comp. WW Comp. Comp. Comp.	10% 5% 10% 5% 10%	302-104 308-062 Use 303-222 303-222 302-470
R472 R473 R477 R478 R485		100 k 3 k 27 Ω 100 Ω 1 k	½ w 5 w ½ w ½ w 7 w	Var. Mica Plate	Comp. WW Comp. Comp.	10% 5% 10% Gain Adjust 1%	302-104 308-062 302-270 311-169 *310-523
R486 R487 R488A, <b>B</b> R489 <b>R</b> 490	X140 up	1 k 470 Ω 33 k 750 Ω 680 Ω	7 w 1/2 w 1/2 w 10 w 1/2 w	Mica Plate	Comp. Comp. WW Comp.	1% 10% 10% 5% 10%	*310-523 302-471 (2) 302-333 308-016 302-681
R491 R494 R496 R497 R498	101- <b>2</b> 1 <i>6</i> 9	100 Ω 39 k 1 k 100 Ω 39 k	1/ <sub>2</sub> w 2 w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 2 w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	302-101 306-393 302-102 302-101 306-393
R498 R503 R504 R520 R521	2170-up	15 k 1 meg 100 k 47 Ω 47 Ω	5 w 1/2 w 1/2 w 1/2 w 1/2 w		WW Prec. Comp. Comp. Comp.	5% 1% 10% 10% 10%	308-108 309-014 302-104 302-470 302-470
R522 R523 R524 R526 R527		220 Ω 100 Ω 39 k 39 k 200 Ω	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1 k 1 k	Var.	Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% D.C. Bal	302-221 302-101 304-393 304-393 311-158
R530 R533 R534 R537 R538		47 Ω 600 Ω 8.2 k 8.2 k 600 Ω	1/2 w 2 w 2 w 2 w 2 w		Comp. ica Plate Comp. Comp. ica Plate	10% 1% 5% 5% 1%	302-470 *310-567 305-822 305-822 *310-567
R539 R540 R540 R541 R542	101-4851 4852-⊍p	3.5 k 27 Ω 47 Ω 550 Ω 100 k	5 w 1/2 w 1/2 w 1/2 w	Var.	WW Comp. Comp. WW Comp.	5% 10% 5% Volts/Div. 10%	308-080 302-270 301-0470-00 Use *311-287 302-104

Ckt. No.	S/N Range		Descripition				Tektronix Part Number
R543 R544 R546 R547 R548		180 k 180 k 27 k 2 × 50 k <b>2</b> 7 <b>k</b>	1/2 w 1/2 w 1/2 w 1/2 w	Var.	Comp. Comp. Comp. Comp. Comp.	10% 10% 10% Position 10%	302-184 302-184 302-273 311-111 302-273
R550 R551 R555 R556 R558		47 Ω 47 Ω 670 Ω 47 Ω 12 k	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w 2 w		Comp. Comp. Prec. Comp. Comp.	10% 10% 1% 10% 10%	302-470 302-470 309-082 302-470 306-123
R571 R572 R573 R576 R577		47 Ω 1.2 k 220 k 56 k 1.5 meg	1/ <sub>2</sub> w 1 w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	302-470 304-122 302-224 302-563 302-155
R578 R580 R581 R583 R584		10 k 18 k 180 k 360 k 120 k	1/ <sub>2</sub> w 2 w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w		Comp. Comp. Comp. Comp. Comp.	10% 10% 5% 5% 10%	302-103 306-183 301-184 301-364 302-124
R585 R586 R588 R589 R590		18 k 180 k 120 k 360 k 100 Ω	2 w ½ w ½ w ½ w ½ w		Comp. Comp. Comp. Comp. Comp.	10% 5% 10% 5% 10%	306-183 301-184 302-124 301-364 302-101
R591 R593 R594 R594 R595	101-299 300-ир	220 k 220 k 10 k 8 k 1.5 k	У <sub>2</sub> w У <sub>2</sub> w 5 w 5 w У <sub>2</sub> w		Comp. Comp. WW WW Comp.	10% 10% <b>5%</b> 5% 10%	302-224 302-224 Use 308-007 308-007 302-152
R596 R597 R601 R603 R609	X630-up	27 k 100 Ω 50 Ω 33 Ω 33 k	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1 w 1/ <sub>2</sub> w	Var.	Comp. Comp. WW Sca Comp. Comp.	10% 10% le Illum. 10% 10%	302-273 302-101 311-057 304-330 302-333
R615 R616 R617 R618 R623		68 k 10 k 50 k 1 meg 1 k	1/2 w 1/2 w 1/2 w 1/2 w	Var.	Prec. WW —1. Prec. Comp. Comp.	1 % 50 Adjust 1 % 10 % 10 %	309-042 311-015 309-090 302-105 302-102
R627 R633 R635 R638 R639		750 Ω 470 k 18 k 120 k 10 k	25 w ½ w 1 w ½ w ½ w		WW Comp. Comp. Comp.	5% 10% 10% 10% 10%	308-147 302-474 304-183 302-124 302-103

Ckt. No.	S/N Range		Descripitio	n			Tektronix Part Number
R640 R642 R644 R650 R651		10 Ω 10 Ω 10 Ω 333 k 490 k	2 w 2 w 2 w ½ w ½ w		Comp. Comp. Comp. Prec. Prec.	10% 10% 10% 1%	306-100 306-100 306-100 309-053 309-002
R653 R657 R658 R659 R663		1 meg 330 k 47 k 33 k 1 k	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	302-105 302-334 302-473 302-333 302-102
R664 R666 R667 R670 R671		1 k 1 k 1.25 k 610 k 300 k	1/ <sub>2</sub> w 10 w 25 w 1/ <sub>2</sub> w 1/ <sub>2</sub> w		Comp. WW WW Prec. Prec.	10% 5% 5% 1% 1%	302-102 308-089 308-102 309-006 309-125
R673 R677 R678 R678 R679	X4852-up 101-4851 4852-up 101-4851	1 meg 560 k 270 k 560 k 47 k	1/2 w 1/2 w 1 w 1/2 w 1/2 w		Comp. Comp. Comp. Comp. Comp.	10% 5% 10% 5% 10%	302-105 301-0564-00 Use 304-274 301-0564-00 302-473
R679 R680 R684 R685 R801	4852-up	56 k 100 k 180 k 47 k 1 k	1/2 W 1/2 W 1/2 W 1/2 W 1/2 W		Comp. Comp. Comp. Comp. Comp.	5% 10% 10% 10% 10%	301-0563-00 302-104 302-184 302-473 302-102
R802 R803 R806 R807 R814		390 Ω 68 k 47 k 1.5 k 470 k	1 w 2 w ½ w ½ w ½ w		Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	304-391 306-683 302-473 302-152 302-474
R815 R824 R824 R825 R825	101-709 710-up 101-709 710-up	47 Ω 6.8 meg 6.8 meg 6.8 meg 6.8 meg	1/2 w 1 w 2 w 1 w 2 w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	302-470 304-685 306-685 304-685
R826 R826 R827 R828 R828	101-4449 4450-up 101-709 710-up	1 meg 1 meg 33 k 1 meg 1 meg	1/ <sub>2</sub> w 1/ <sub>4</sub> w 1/ <sub>2</sub> w	Var. Var.	Comp. Comp. Comp. Comp. Comp.	Intensity Intensity 10% 10% 10%	311-041 311-0041-02 302-333 316-105 302-105
R840 R841 R842 R842 R844	101-709 710-up 101-4449	2.2 meg 2 meg 3.3 meg 3.3 meg 2 meg	1 w 2 w	Var. Var.	Comp. Comp. Comp. Comp.	10% HV Adjust 10% 10% Focus	302-225 311-042 304-335 306-335 311-043

Ckt. No.	S/N Range		Descripition				Tektronix Part Number
R844 R845 R846 R846 R847	4450-ир 101-709 710-ир	2 meg 1 meg 10 k 10 k 27 k	1/ <sub>2</sub> w 1/ <sub>4</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w	Var.	Comp. F Comp. Comp. Comp. Comp.	Focus 10% 10% 10% 10%	311-0043-02 302-105 316-103 302-103 302-273
R848 R861 R862 R863 R864		1 meg 100 k 100 k 120 k 100 k	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w	Var. Var.	Comp. Comp. Geo Comp. Comp. Comp. Astig	10% 10%	302-105 311-026 302-104 302-124 311-026
R870 R871 R872 R874 R874	101-339 340-ир	150 k 2.7 meg 1 k 3.3 meg 3.9 meg	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w		Comp. Comp. Comp. Comp. Comp.	10% 10% 10% 10% 10%	302-154 302-275 302-102 302-335 302-395
R875 R876 R878 R879 <b>R8</b> 80	101-339	68 k 1 k 33 k 10 k 100 k	1/2 w 1/2 w 1/2 w 1/2 w	Var.	Comp. Comp. Comp. Comp. Cal. Comp.	10% 10% 10% Adjust 10%	302-683 302-102 302-333 311-076 302-104
R880 R883 R885 R886 R887	340-ир	68 k 100 Ω 10 k 6 k 2 k	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w		Comp. Comp. Prec. Prec. Prec.	10% 10% 1% 1% 1%	302-683 302-101 309-100 309-099 309-098
R888 R889 R890 R891 R892		1 k 600 Ω 200 Ω 100 Ω 60 Ω	1/2 w 1/2 w 1/2 w 1/2 w 1/2 w		Prec. Prec. Prec. Prec. Prec.	1 % 1 % 1 % 1 %	309-115 309-097 309-073 309-112 309-067
R893 R894 R895		<b>20</b> Ω 10 Ω 10 Ω	1/ <sub>2</sub> w 1/ <sub>2</sub> w 1/ <sub>2</sub> w		Prec. Prec. Prec.	1% 1% 1%	309-064 309-096 309-096
			Switche	<b>)</b> \$			Wired Unwired
SW10A,B, SW10A,B SW10A,B SW110 SW160	101-269 270-1958 1959-up 101-3929	Trigger Selector Trigger Selector Trigger Selector Preset W/R17, R11 Time/Division	0 & R349			Use *	262-333 *260-332 050-107 *260-377 262-564 *260-377 311-099 262-334 *260-329
SW160 SW160 SW348 SW348 SW400	3930-4769 4770-up 101-569 570-up 101-569	Time/Division Time/Division Horizontal Display Horizontal Display Polarity "A"				* <b>262-03</b> 34 *( **	4-01 260-0329-00 4-02 260-0329-00 050-027 *260-186 262-430 *260-186 050-046 *260-331

## Switches (cont)

Ckt. No.	S/N Range	Descripition	Tektronix Part Number
SW400 SW410 SW500 SW500 SW510	570-ир 101-569 570-ир	Polarity "A" Volts/Division "A" turret atten complete Polarity "B" Polarity "B" Volts/Division "B" turret atten complete	*262-432 *260-331 *263-003 Use *050-047 *260-326 *262-433 *260-326 *263-001
SW585 SW601 TK601 SW848 SW870		Mode Power W/R601 Thermo Cutout 137° CRT Cathode Selector Square Wave Calibrator	*260-325 311-057 260-120 *260-209 *262-332 *260-098
		Diodes	
D131 D152 D454 D477 D478	Х3800-ир	Germanium, T12G Silicon, Low Leakage 0.25 w 40 V Germanium, T12G Germanium, Tek Spec Germanium, Tek Spec	152-008 152-0246-00 152-008 Use *152-0075-00 Use *152-0075-00
D554 D594 D640A,B,C,D D640A,B,C,D D642A,B,C.D	3800-up	Germanium, T12G Silicon Replaceable by 1N4152 1N2862 Silicon, 1N3194 1N2862	152-008 *152-0185-00 Use 152-047 152-0066-00 Use 152-047
D642A,B,C,D D644A,B,C,D D644A,B,C,D	101-3799	Silicon, 1N3194 1N2862 Silicon, 1N3194	152-0066-00 Use 152-047 152-0066-00
		Transformers	
T601 T801		Power (All Voltages) CRT Supply	*120-142 *120-079
		Electron Tubes	
V24 V45 V135 V145 V145	101-4049 4050-up	6DJ8 6DJ8 6DJ8 6AN8 EFC-80/6BL8	154-187 154-187 154-187 154-078 154-0278-00
V152 V152 V161 V183 V193	101-3799 3800-up	6AL5 Selected 6AL5 6AN8 12AT7 6DJ8	Use *157-0104-01 154-0016-00 154-078 154-039 154-187
VILI V364 V384 V423 V434		6DJ8 6DJ8 6DJ8 6DJ8	154-187 154-187 154-187 154-187 154-187

## Electron Tubes (cont)

Ckt. No.	S/N Range	Descripition	Tektronix Part Number
V454		6GM8 (ECC86)	Use *157-068
V463		6DJ8	154-187
V464		6197	154-146
V474		6197	154-146
V493		6DJ8	154-187
V523		6DJ8	154-187
V534		6DJ8	154-187
V554		6GM8 (ECC86)	Use *157-068
V572		6AL5	154-016
V574		6DJ8	154-187
V585 V594 V594 V609 V627	101-4329 4330-∪p	12AT7 6DJ8 7119 5651 6AU5	154-039 154-187 154-0340-00 154-052 154-021
V634		6AN8	154-078
V654		6AU6	154-022
V667		6080	154-056
V674		6AU6	154-022
V800		6AQ5	154-017
V814		12AU7	154-041
V822		5642	154-051
V832		5642	154-051
V842		5642	154-051
V859		T0550-31 CRT Standard Phosphor	Use *154-344
V875 V885 V885	101-339 340-up	6AU6 12AT7 12AU7	154-022 154-039 154-041

# **Type 516 Turret Attenuator**

Values fixed unless marked variable.

Capo	acitors
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Tolerance ±20% unless oth	nerwise indicated.				
C407C	82 pf	Cer.	500 v	10%	281-528
C408B C408C				Adjusting Slug Adjusting Slug	use 214-142 use 214-142
C408D C409B	5.6 pf	Cer.	500 v	10%	281-544
C407B				Adjusting Slug	use 214-142
C409 <b>C</b>				Adjusting Slug	use 214-142
C410B				Adjusting Slug	use 214-142
C410C C410E	00 t	<b>C</b>	500	Adjusting Slug	use 214-142
C410E	22 pf	Cer.	500 v	10%	281-511
CTITE				Adjusting Slug	use 214-142

#### **Turret Attenuator (cont)**

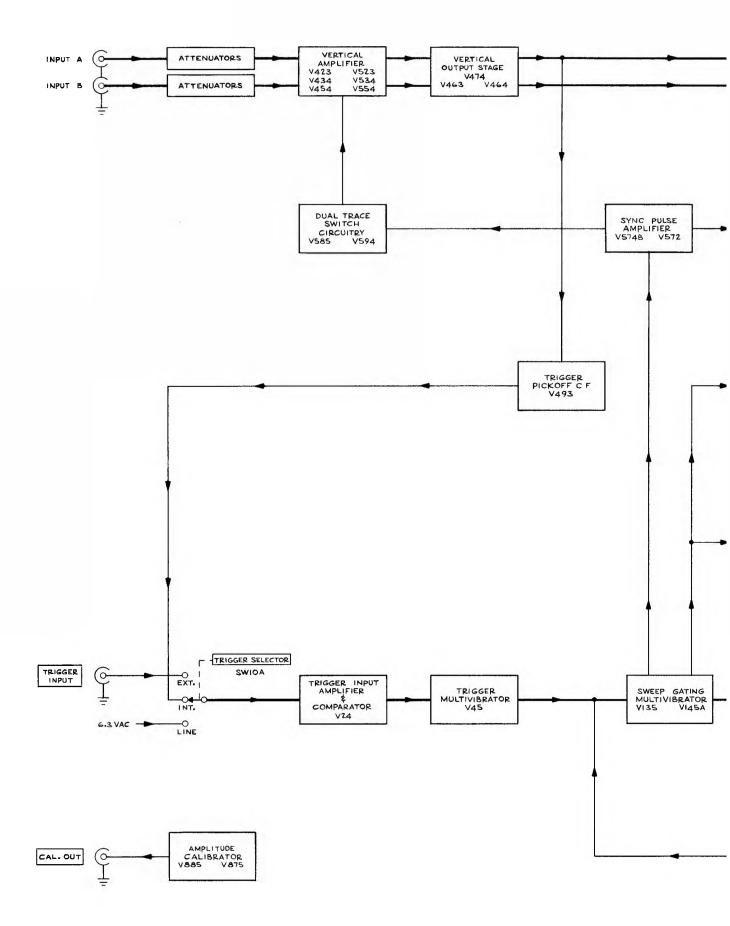
#### Capacitors (cont)

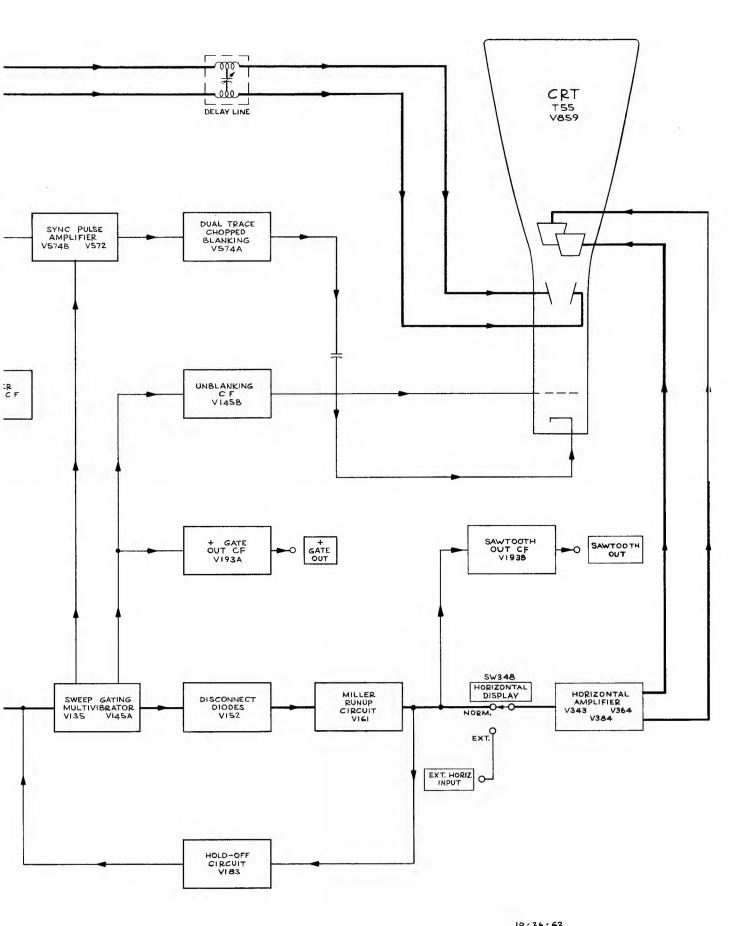
Ckt. No.	S/N Range		Description			Tektronix Part Number
C411C C411E C412A C412B C412C	X3300-up	47 pf Selected	Cer.	500 v Nominal value 2.2 pf	Adjusting Slug 10% Adjusting Slug Adjusting Slug	use 214-142 281-519 281-0500-00 use 214-142 use 214-142
C412E C413A C413B C413C C413E <sup>4</sup>	Х3300-ир	100 pf Selected 250 pf	Cer. Mica	500 v Nominal value 2.2 pf 500 v	10% Adjusting Slug Adjusting Slug 10%	281-530 281-0500-00 use 214-142 use 214-142 283-0539-00
C414A C414B C414C C414E <sup>4</sup> C415A	X3300-up	Selected 500 pf Selected	Mica	Nominal value 2.2 pf  500 v  Nominal Value 2.2 pf	Adjusting Slug Adjusting Slug 10%	281-0500-00 use 214-142 use 214-142 283-0541-00 281-500
C415E <sup>4</sup> C415B C415C		625 pf	Mica	500 v	10% Adjusting Slug Adjusting Slug	283-0547-00 use 214-142 use 214-142
			Res	istors		
Resistors are	fixed, composition,	$\pm 10\%$ unless of	herwise indica	ited.		
R407C R407E R408C R408C R408E	101-5235 5236-up 101-5235	82 Ω 47 Ω 500 k 500 k 1 meg	1/4 w 1/4 w 1/2 w 1/4 w 1/8 w	Prec. Prec. Prec.	1 % ½ % 1 %	316-820 316-470 309-140 322-0610-01 318-004
R408E R409C R409C R409E R409E	5236-up 101-5235 5236-up 101-5235 5236-up	1 meg 750 k 750 k 333 k 333 k	1/4 W 1/2 W 1/2 W 1/8 W 1/8 W	Prec. Prec. Prec. Prec. Prec.	1/2% 1% 1/2% 1% 1%	322-0481-01 309-141 323-0655-00 318-005 321-0628-01
R410C R410C R410E R410E R411C	101-5235 5236-up 101-5235 5236-up 101-5235	900 k 900 k 111 k 111 k 950 k	1/2 W 1/2 W 1/8 W 1/8 W 1/2 W	Prec. Prec. Prec. Prec. Prec.	1 % ½% 1 % ½% 1 %	309-142 323-0611-01 318-006 321-1389-01 309-143
R411C R411E R411E R412C	5236-up 101-5235 5236-up 101•5735 5236-up	950 k 52.6 k 52.6 k 975 k 975 k	1/2 w 1/8 w 1/8 w 1/8 W 1/2 W	Prec. Prec. Prec. Prec. Prec.	1/2 % 1 % 1/2 % ] % 1/2 %	323-0612-01 318-007 321-0616-01 309•144 323-0757-01

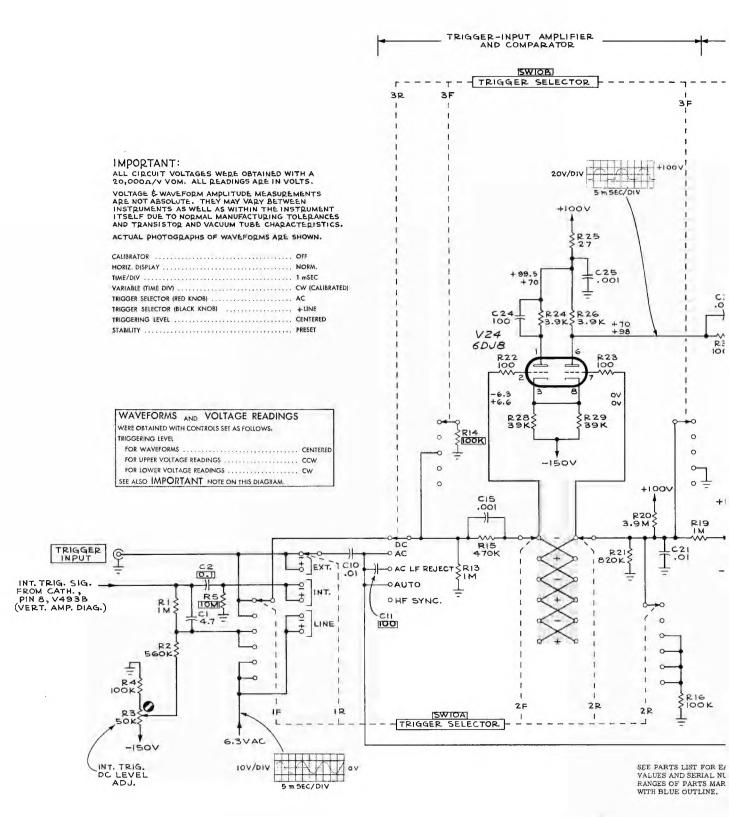
<sup>\*</sup>These capacitors are installed of the factory by a special process. If replacement is necessary, order a wired turnet body, part number \*204-0129-00 (S/N 2060-up). Below S/N 2060 order \*263-0001-00 or \*263-0003-00 for wired turnet body.

## **Turret Attenuator (cont)**

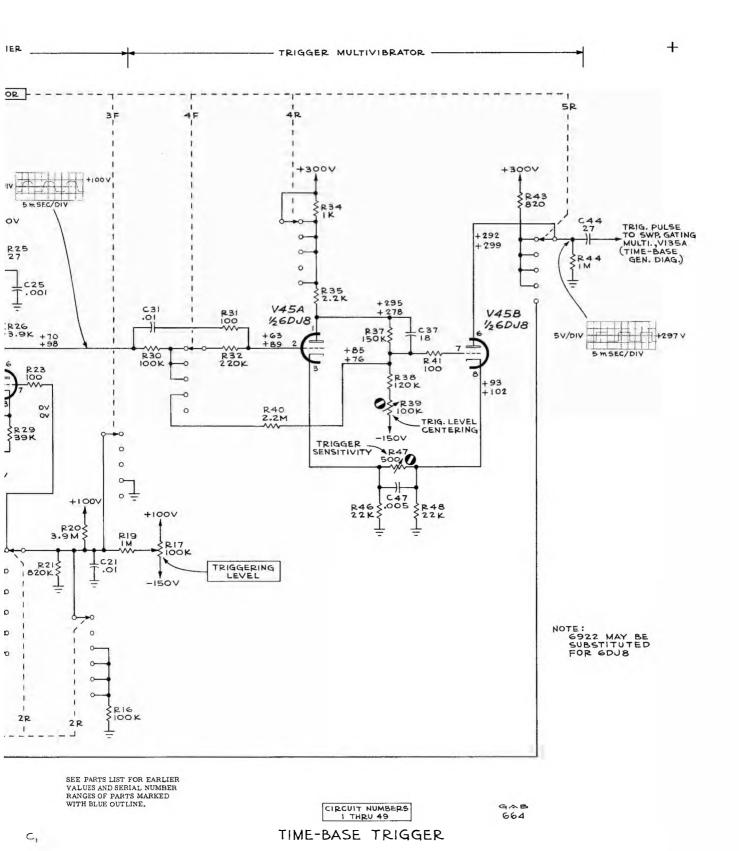
Ckt. No.	S/N Range		Description			Tektronix Part Number
R412E	101-5235	25.6 k	1/8 W	Prec.	1%	318-008
R412E	5236-up	25.6 k	1/8 W	Prec.	½%	321-0627-01
R413C	101-5235	990 k	1/2 W	Prec.	1%	309-145
R413C	5236-up	990 k	1/2 W	Prec.	½%	323-0614-01
R413E	101-5235	10.1 k	1/8 W	Prec.	1%	318-009
R413E	5236-up	10.1 k	1/8 W 1/2 W 1/2 W 1/8 W 1/8 W	Prec.	Y2 %	321-1289-01
R414C	101-5235	995 k		Prec.	1 %	309-146
R414C	5236-up	995 k		Prec.	Y2 %	323-0615-01
R414E	101-5235	5.03 k		Prec.	1 %	318-010
R414E	5236-up	5.03 k		Prec.	Y2 %	321-0613-01
R415C	101-5235	997.5 k	1/ <sub>2</sub> w	Prec.	1%	309-147
R415C	5236-up	997.5 k	1/ <sub>2</sub> w	Prec.	½%	323-0616-01
R415E	101-5235	2.51 k	1/ <sub>8</sub> w	Prec.	1%	318-011
R415E	5236-∪p	2.51 k	1/ <sub>8</sub> w	Prec.	½%	321-0626-01

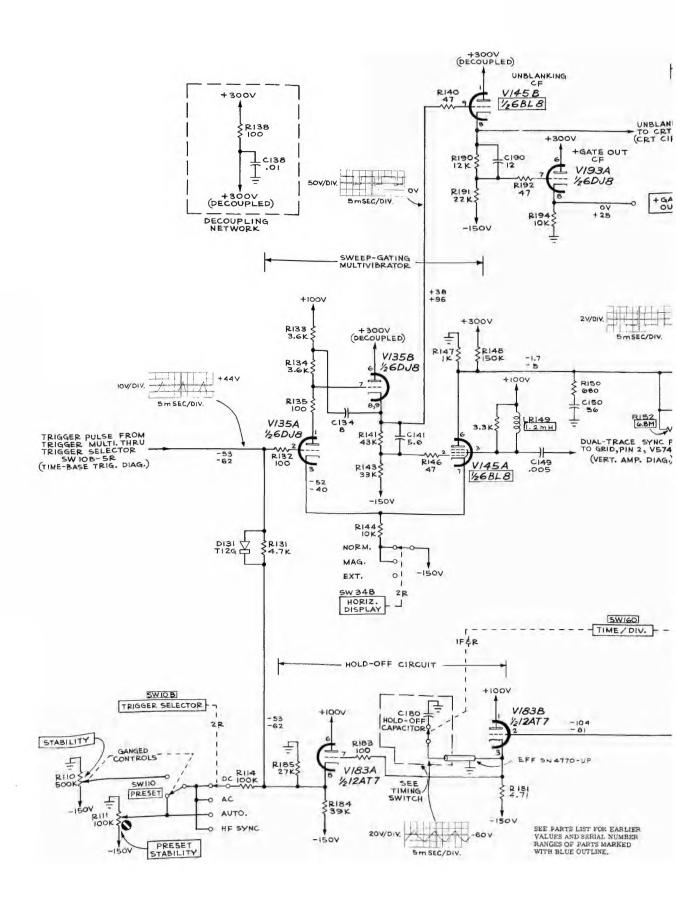




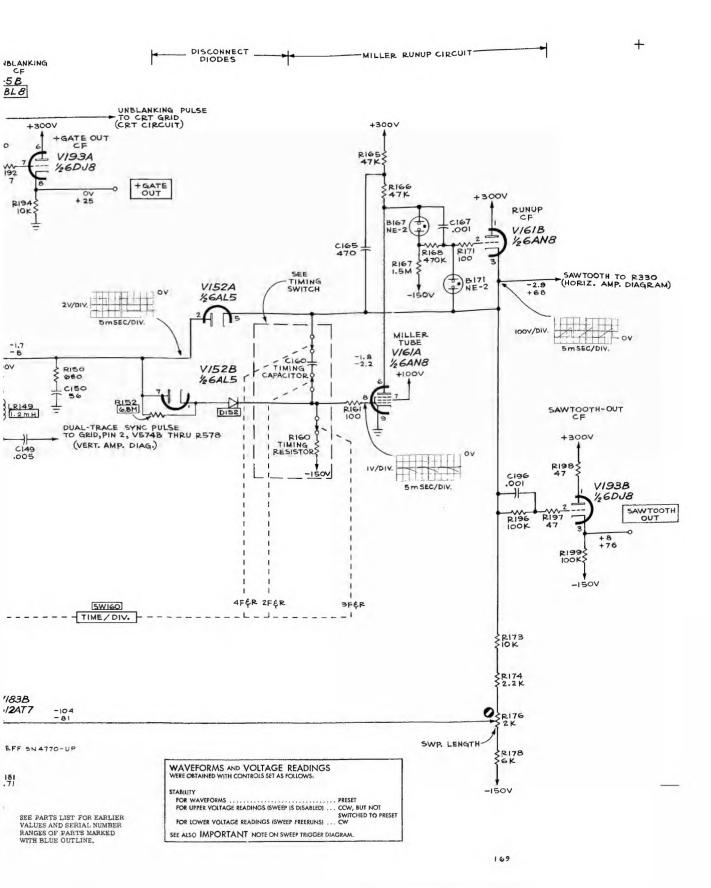


TYPE 516 OSCILLOSCOPE



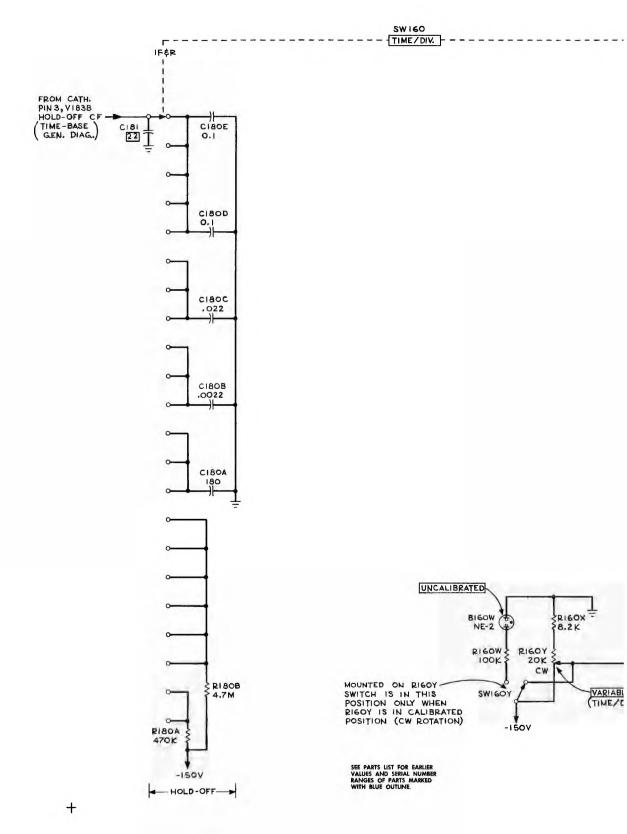


TYPE 516 OSCILLOSCOPE

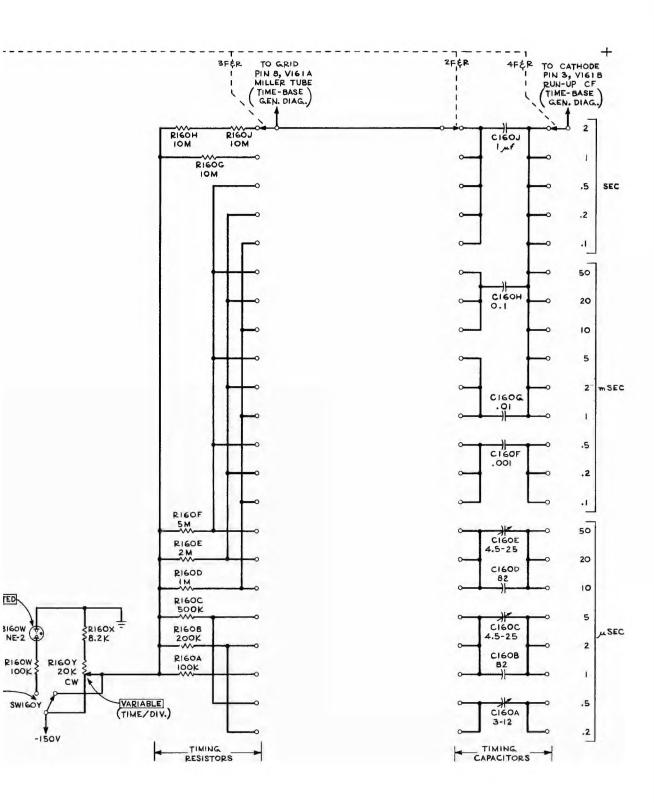


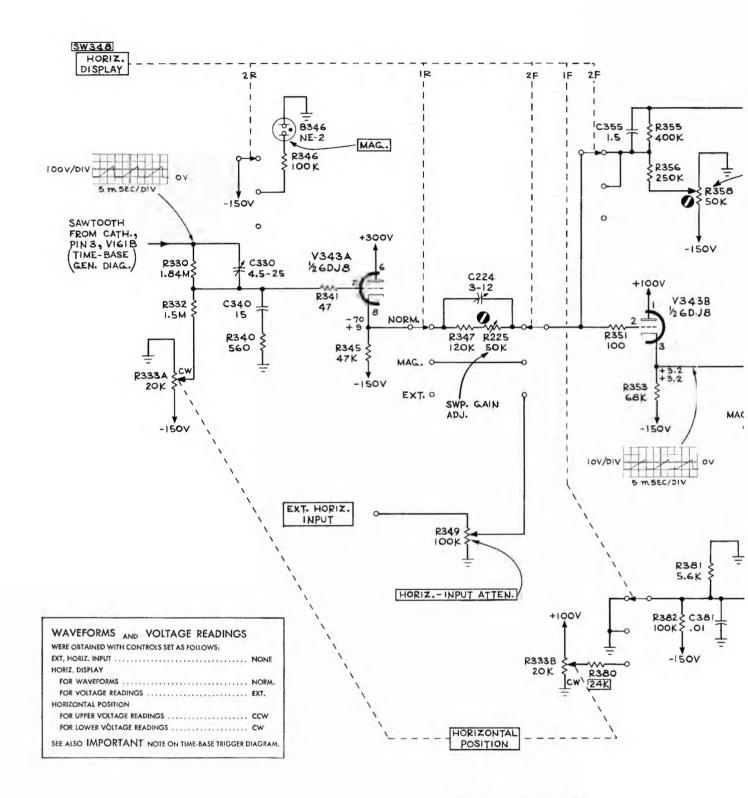
TIME-BASE GENERATOR

CIRCUIT NUMBERS

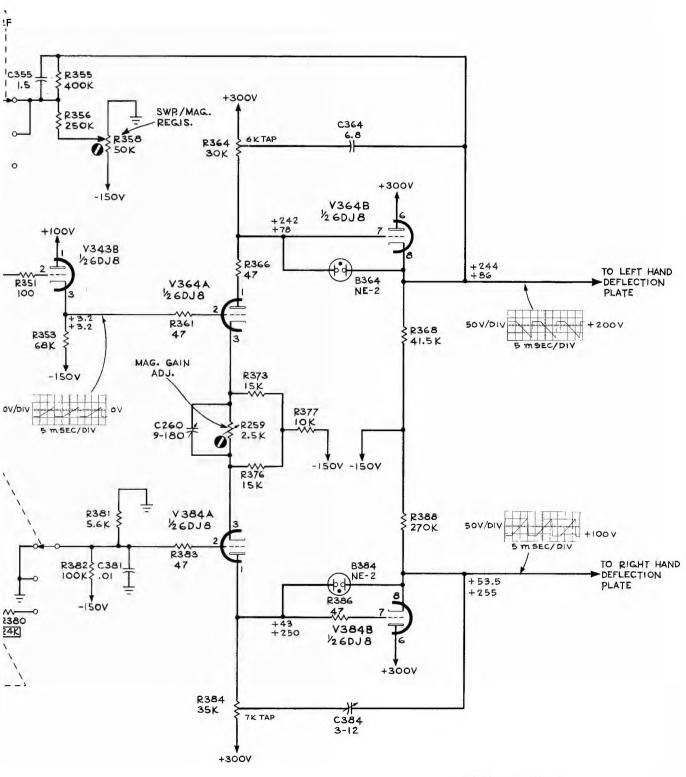


TYPE 516 OSCILLOSCOPE



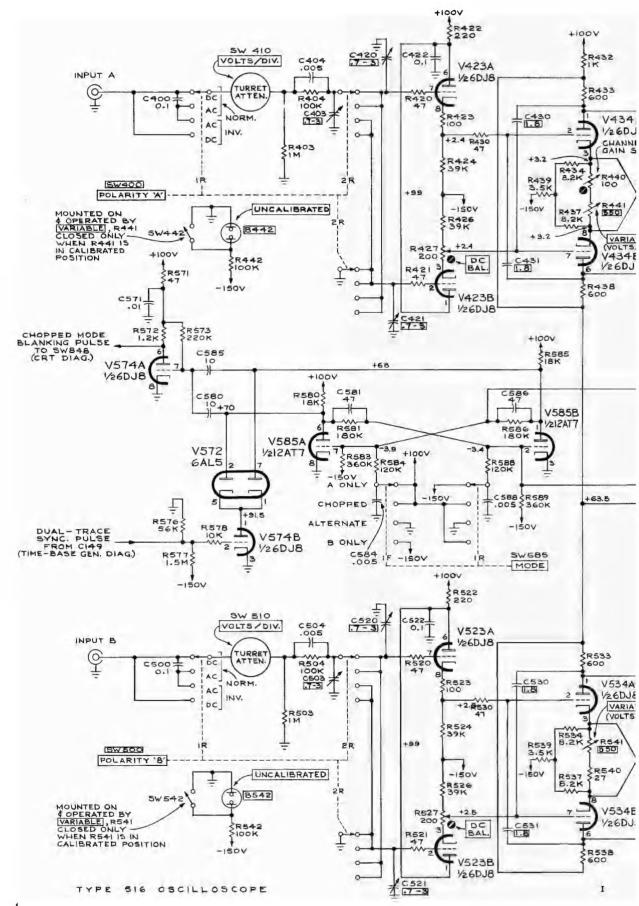


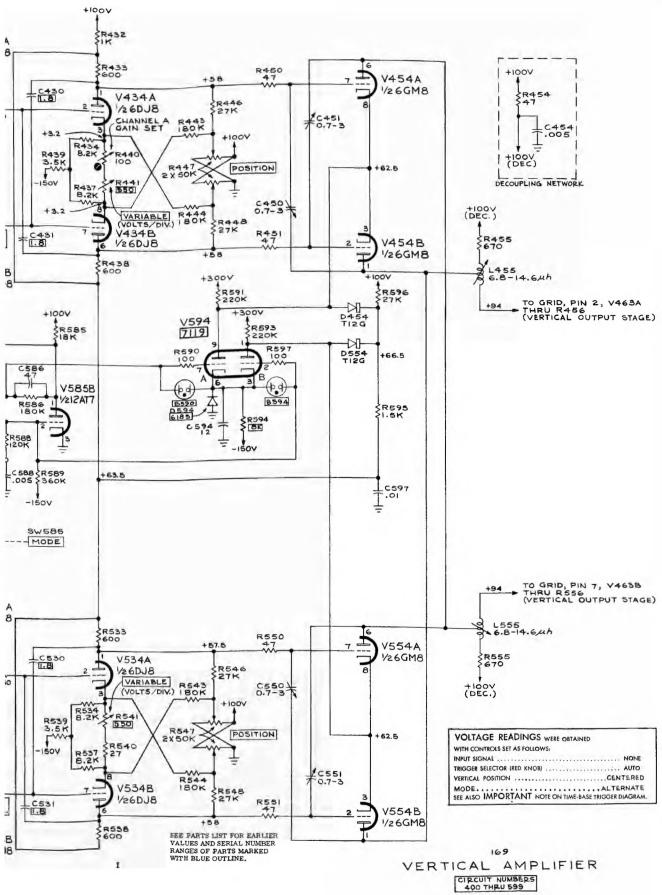
SEE PARTS LIST FOR EARLIER VALUES AND SERIAL NUMBER RANGES OF PARTS MARKED WITH BLUE OUTLINE.

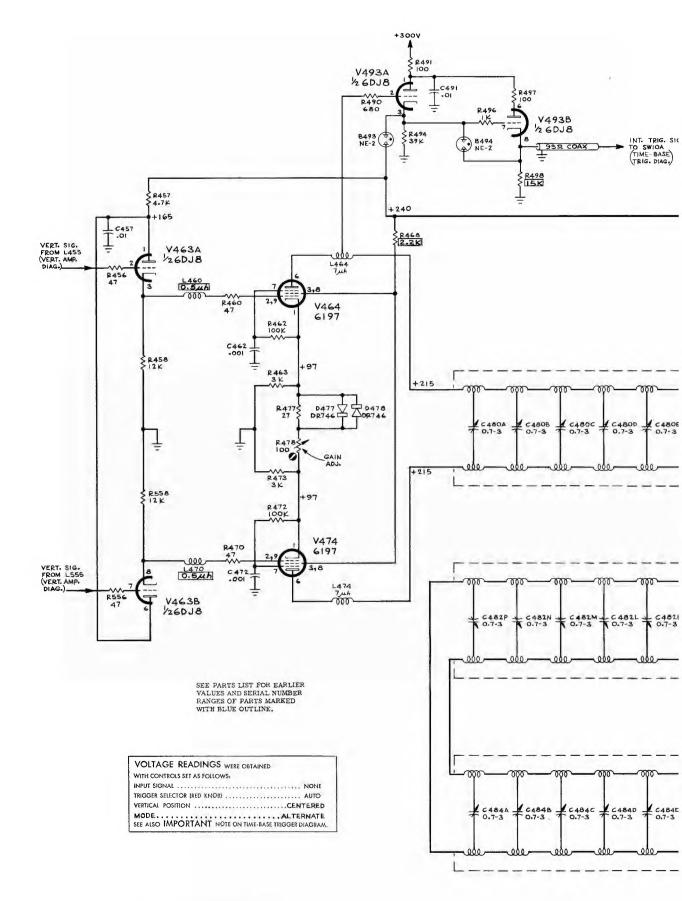


CIRCUIT NUMBERS 225 THRU 389

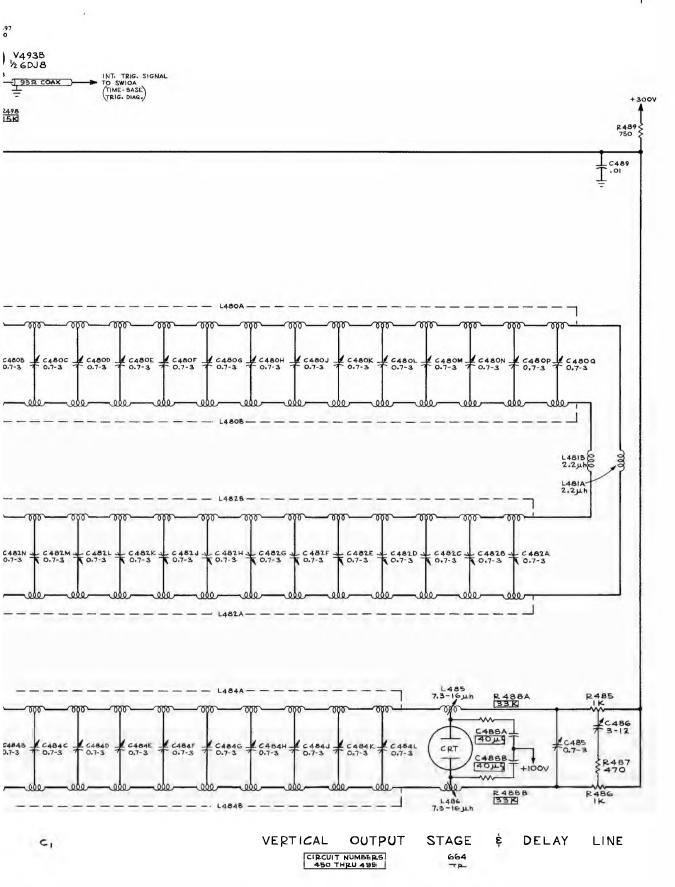
HORIZONTAL AMPLIFIER

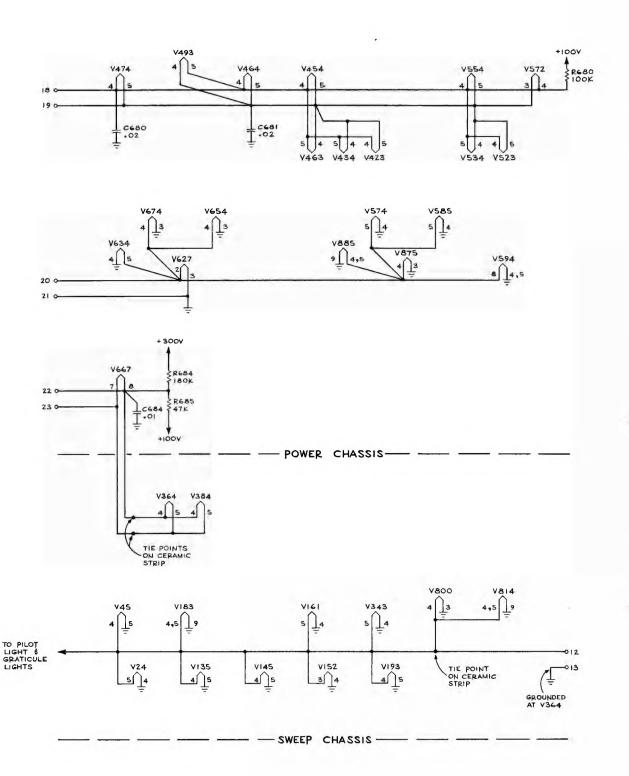






TYPE 516 OSCILLOSCOPE





SWGOI

FGOI

1.6A

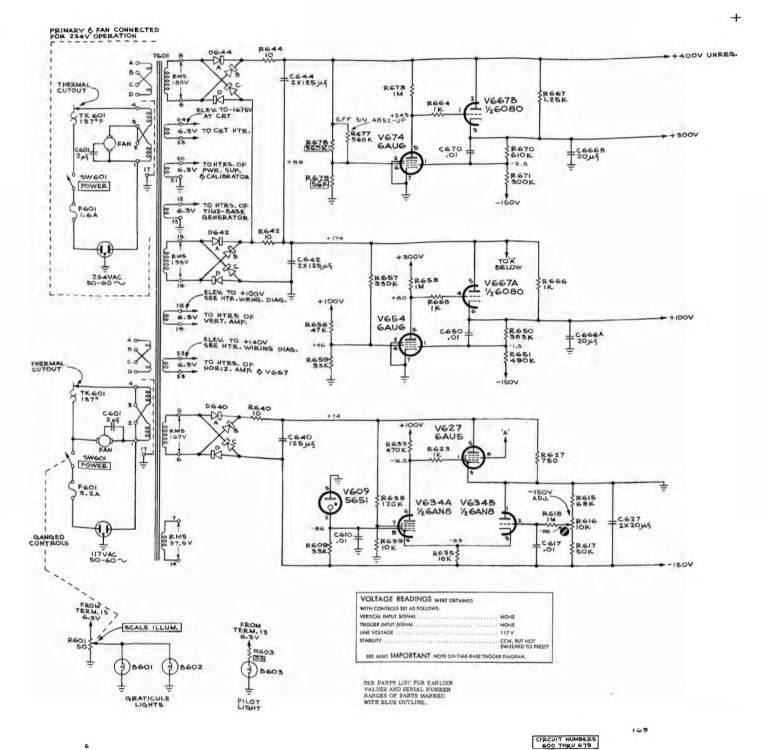
TKGOI

1379

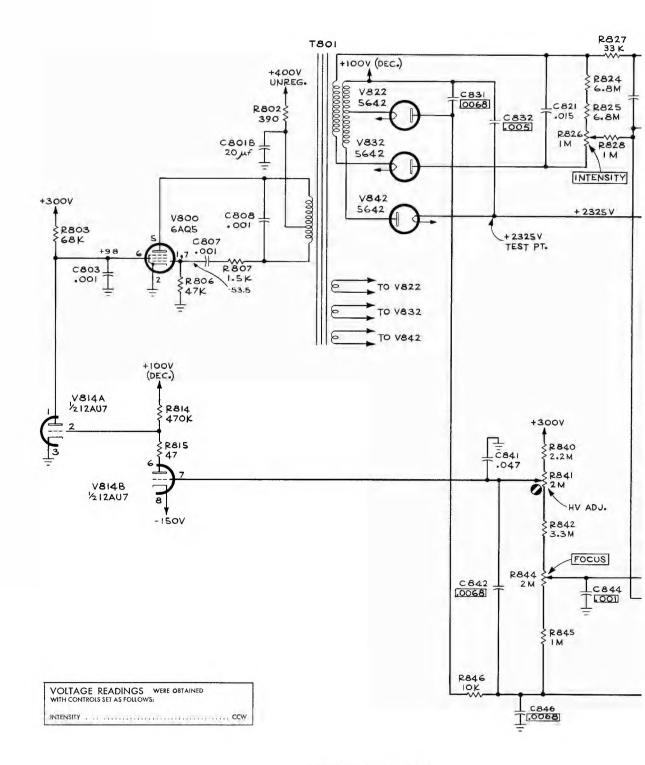
CGOI

1379

+

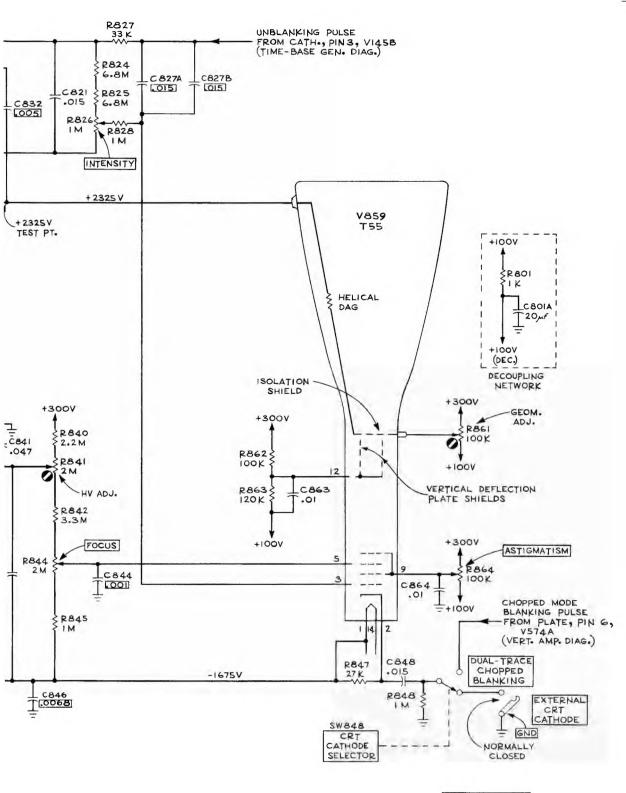


POWER SUPPLY



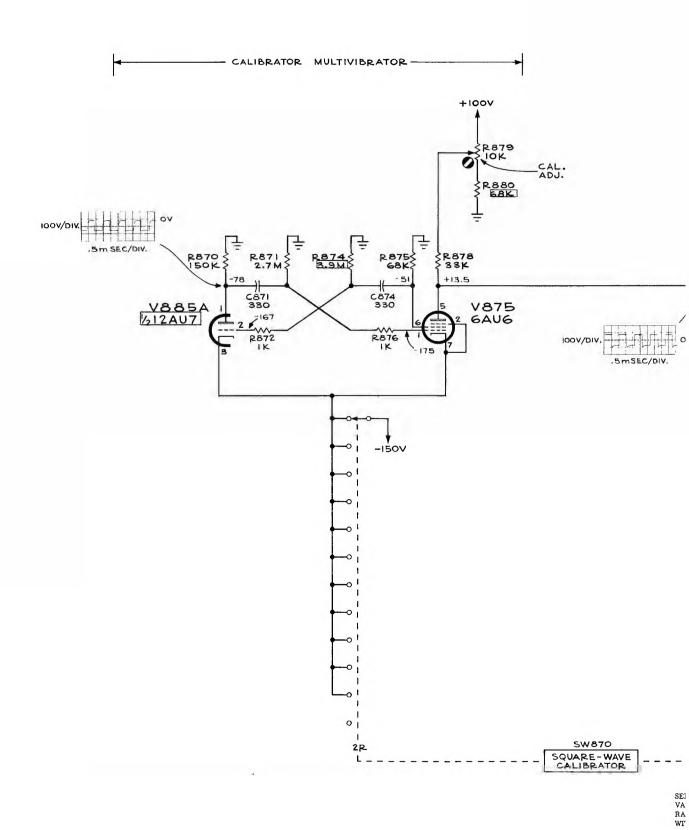
SEE PARTS LIST FOR EARLIER VALUES AND SERIAL NUMBER RANGES OF PARTS MARKED WITH BLUE OUTLINE.

TYPE 516 OSCILLOSCOPE



SOO THEU 869

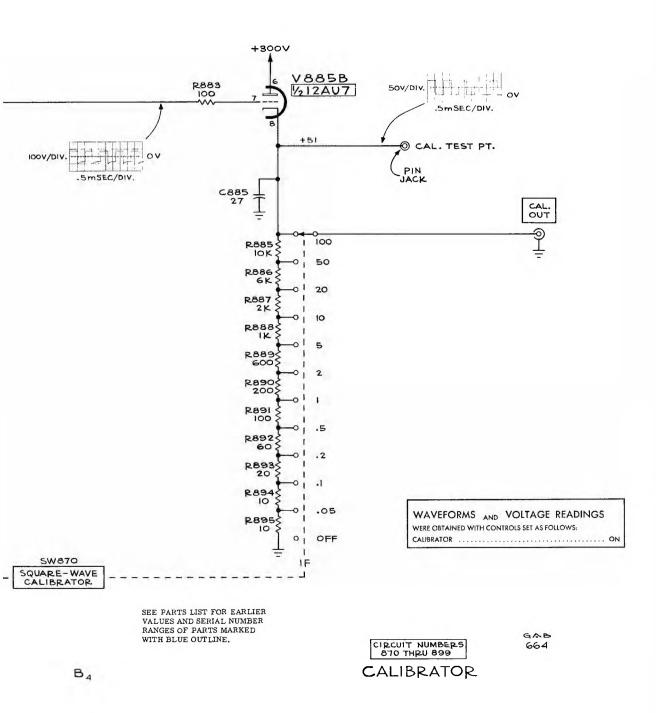
CRT CIRCUIT

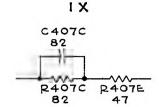


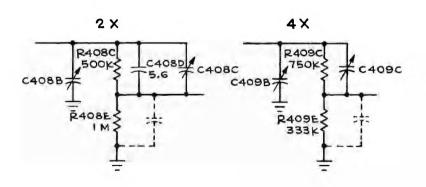
TYPE 516 OSCILLOSCOPE

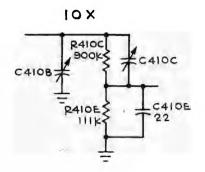
 $B_4$ 

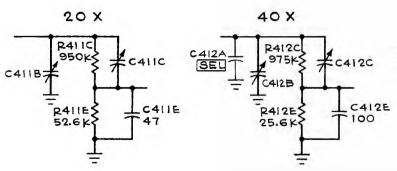
CAL.

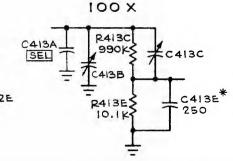


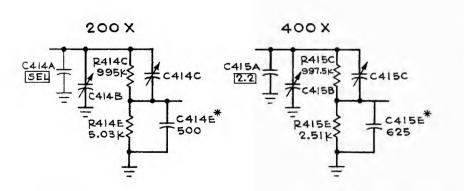












SEE PARTS LIST FOR EARLIER VALUES AND SERIAL NUMBER RANGES OF PARTS MARKED WITH BLUE OUTLINE.

# NOTE: CAPACITORS MARKED WITH \* ARE SILVERED MICA BUTTONS ALL VARIABLE CAPACITORS ARE APPROX. 0.3-8 pf

**TYPE 516** 

TURRET ATTENUATOR

C

#### MANUAL CHANGE INFORMATION

At Tektronix, we continually strive to keep up with latest electronic developments by adding circuit and component improvements to our instruments as soon as they are developed and tested.

Sometimes, due to printing and shipping requirements, we can't get these changes immediately into printed manuals. Hence, your manual may contain new change information on following pages.

A single change may affect several sections. Sections of the manual are often printed at different times, so some of the information on the change pages may already be in your manual. Since the change information sheets are carried in the manual until ALL changes are permanently entered, some duplication may occur. If no such change pages appear in this section, your manual is correct as printed.

#### TEXT CORRECTION

Section 6

Calibration

Page 6-8 Step 22, top of left column

CHANGE: "two cycles" to read "four cycles" in line five.

## ELECTRICAL PARTS LIST CORRECTION

CHANGE TO:

C181

281-0511-00

22 pF

Cer

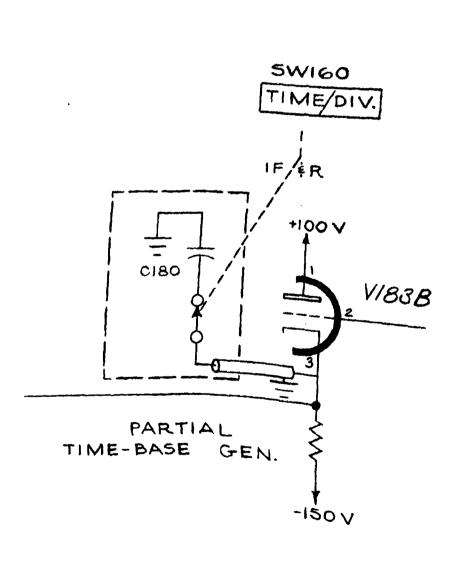
500 V

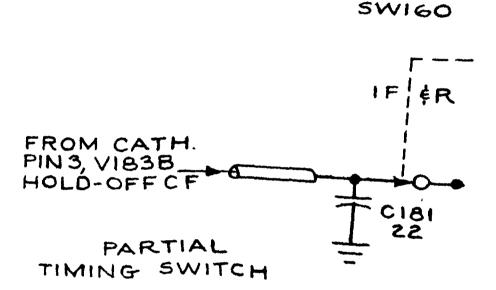
SW160

262-0334-02

Time/Division

## SCHEMATIC CORRECTION





#### ELECTRICAL PARTS LIST CORRECTION

#### CHANGE TO:

L460	108-0170-01	<b>0.</b> 5 μH
L470	108-0170-01	<b>0.</b> 5 μH

### ELECTRICAL PARTS LIST AND SCHEMATIC CORRECTION

CHANGE TO:

R478

311-0097-00

200  $\Omega$ , Var